#### **ABSOLUTE MAXIMUM RATINGS**

(Referred to GND)

(	(Releited to GND)		Continuous Power Dissipation
١	Vcc	0.3V to +6V	µMAX (derate 4.1mW/°C a
-	TXD, SHDN, LEDC	0.3V to +6V	SO (derate 5.88mW/°C ab
ł	RXD	0.3V to (V <sub>CC</sub> + 0.3V)	Operating Temperature Range
F	PGND	0.1V to +0.1V	MAX3120C_A
F	PINC	10mA	MAX3120E_A
(	Continuous LEDC Current		Junction Temperature
ł	Repetitive Pulsed LEDC Current		Storage Temperature Range
			Lead Temperature (soldering,

Continuous Power Dissipation ( $T_A = +70^\circ$	
µMAX (derate 4.1mW/°C above +70°C	C)
SO (derate 5.88mW/°C above +70°C)	471mW
Operating Temperature Ranges	
MAX3120C_A	0°C to +70°C
MAX3120E_A	40°C to +85°C
Junction Temperature	+150°C
Storage Temperature Range	65°C to +160°C
Lead Temperature (soldering, 10sec)	+300°C

Stresses beyond those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

#### ELECTRICAL CHARACTERISTICS

(V<sub>CC</sub> = +3.0V to +5.5V, T<sub>A</sub> = T<sub>MIN</sub> to T<sub>MAX</sub>, unless otherwise noted. Typical values are at T<sub>A</sub> = +25°C and V<sub>CC</sub> = +3.3V.)

		5.					
PARAMETER	SYMBOL	CONDITIONS	MIN	TYP	MAX	UNITS	
DC CHARACTERISTICS							
Supply Current	Icc	$T_A = +25^{\circ}C, \overline{SHDN} = V_{CC}$ (Note 1)		120	200	μA	
Shutdown Supply Current	ICC(SHDN)	$T_A = +25^{\circ}C$ , $\overline{SHDN} = GND$ (Note 1)		0.01	1.0	μA	
LOGIC INPUTS (TXD, SHDN)		-					
Input Logic Threshold Low	VIL				0.8	V	
Input Logic Throphold Lligh		$V_{CC} = 3.3V$	2.0			- V	
Input Logic Threshold High	VIH	$V_{CC} = 5.0V$	2.4				
Input Leakage Current	ILEAK		-1		1	μA	
Input Capacitance	CIN			2		рF	
LOGIC OUTPUT (RXD)	1						
	Vol	I <sub>SINK</sub> = 200μA		0.1	0.4		
Output Voltage	V <sub>OH</sub>	I <sub>SOURCE</sub> = 100μA	V <sub>CC</sub> - 0.5	V <sub>CC</sub> - 0.05		V	
Output Rise and Fall Time	t <sub>r</sub> , t <sub>f</sub>	$C_{LOAD} = 50 pF$		50		ns	
IR RECEIVER							
Supported Data Rates			2.4		115.2	kbps	
Equivalent Input Noise Current	INOISE	(Note 2)		10		nA <sub>RMS</sub>	
Input Current Sensitivity		(Note 3)	0.0002		6	mA	
Ambient DC Current Rejection		$V_{CC} = 3.3V$		100		μA	
-		$V_{CC} = 5.0V$		375		μπ	
Shutdown Time		Delay until I <sub>CC</sub> < 1µA		10		μs	
Shutdown Disable Time		Delay until maximum IR receiver data rate is valid		300		μs	
IR Receiver Output Pulse Width		Data rate = 2.4kbps	1		90		
in neceiver Output Fuise Width		Data rate = 115.2kbps	1		8	μs	

#### **ELECTRICAL CHARACTERISTICS (continued)**

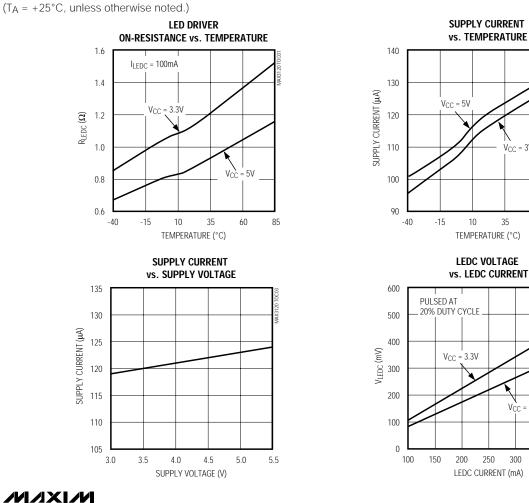
 $(V_{CC} = +3.0V \text{ to } +5.5V, T_A = T_{MIN} \text{ to } T_{MAX}, \text{ unless otherwise noted. Typical values are at } T_A = +25^{\circ}C \text{ and } V_{CC} = +3.3V.)$ 

			51			-	
PARAMETER	SYMBOL	CON	DITIONS	MIN	TYP	MAX	UNITS
IR TRANSMITTER							
Transmitter Rise Time	tr	10% to 90% of 200mA drive current			20	600	ns
Transmitter Fall Time	tf	90% to 10% of 200mA drive current			20	600	ns
Transmitter Output Resistance		IOUT = 200mA	$V_{CC} = 3.3V$		1.15	2.0	Ω
		1001 = 20011A	$V_{CC} = 5.0V$		0.9	1.6	52
Off-Leakage Current					0.01	10	μΑ

Note 1: All supply current measurements are made under the following conditions: no load at all outputs, input voltages at GND or V<sub>CC</sub>, no PIN diode input current.

Note 2: Equivalent input current noise is calculated by dividing the output noise of the transimpedance amplifier by the midband transimpedance gain.

Note 3: Sensitivity is measured with an IrDA-compliant input signal, where the data rate is within the Supported Data Rate, rise/fall times are less than 600ns, and pulse widths are between 1.41µs and 3/16 of the baud rate.



Typical Operating Characteristics

 $V_{CC} = 3V$ 

35

60

 $V_{CC} = 5V$ 

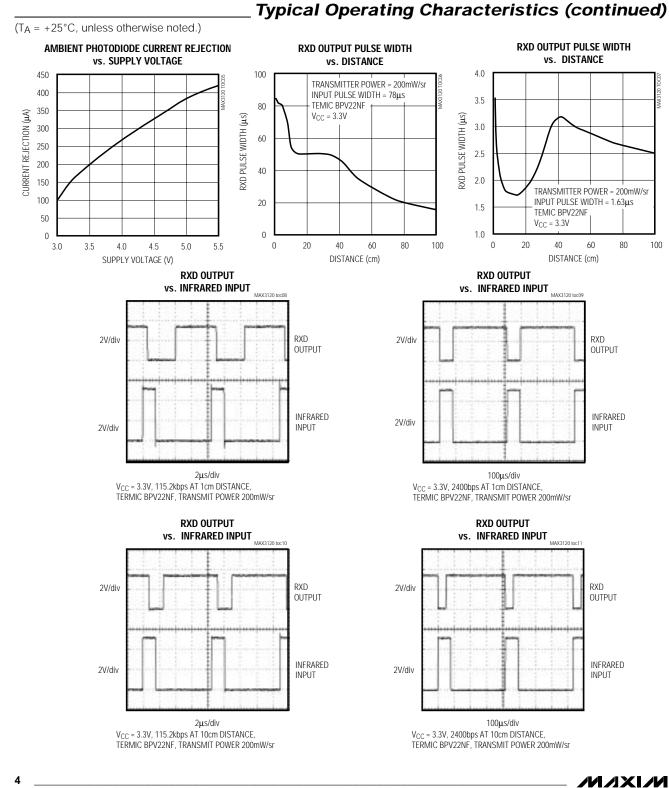
300

85

400

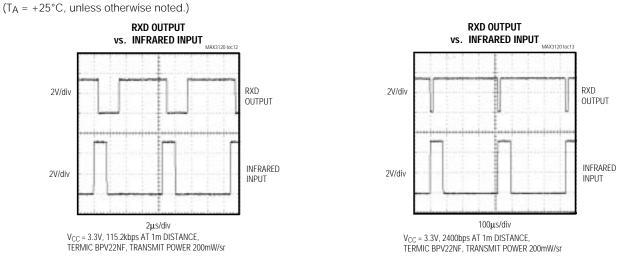
350

MAX3120



# MAX3120

#### Typical Operating Characteristics (continued)



#### \_Pin Description

PIN	NAME	FUNCTION					
1	TXD	IR Transmitter TTL/CMOS Data Input. High = LED on.					
2	V <sub>CC</sub>	Supply Voltage					
3	GND	Ground. Connect anode of PIN diode to GND. Connect GND to PGND.					
4	PINC	PIN Diode Cathode Input. Connect cathode of PIN diode to PINC.					
5	SHDN	DN Shutdown Input. Active low.					
6	PGND	Power Ground. Ground for IR LED driver. Connect PGND to GND.					
7	LEDC	LED Driver Output. Connect cathode of IR-emitting LED to LEDC.					
8	RXD	IR Receiver TTL/CMOS Data Output. Pulses low for IR input pulse.					

#### **Detailed Description**

The MAX3120 is an IrDA 1.2-compatible infrared (IR) transceiver. By selecting appropriate external optical components (see *IR LED and PIN Photodiode Selection* section), the MAX3120 will operate at data rates of 2.4kbps to 115kbps at distances from 1cm to 1m. Because of its low-noise design, the MAX3120 achieves a bit error rate (BER) below 10<sup>-8</sup> at maximum data rates when used with the appropriate external components. On-chip filtering rejects out-of-band ambient light signals that would otherwise interfere with IR communication. Also included in the MAX3120 is a high-power LED driver capable of sinking 200mA. It can drive most available IR LEDs at IrDA speeds of 2.4kbps to 115kbps.

#### Receiver

The MAX3120's IR receiver amplifier reverse biases the PIN diode by approximately 1.2V, and the PIN diode converts pulses of IR light into pulses of current. The input transimpedance (current-to-voltage) amplifier then converts these current pulses into voltage pulses of a useful magnitude. The MAX3120 filters the resulting output voltage pulses to remove low-frequency ambient light interference and high-frequency circuit noise. Finally, a high-speed comparator translates these voltage pulses into usable CMOS output levels (Figure 1).



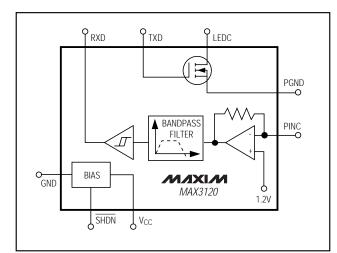


Figure 1. Functional Diagram

MAX3120

#### Transmitter

The MAX3120's IR transmitter consists of a high-power MOS switch, capable of quickly switching 200mA with less than 2 $\Omega$  of on-resistance. Internal buffering keeps the input capacitance of the TXD pin extremely low to ease the input drive requirement. Connect an IR LED in series with a current-setting resistor to select the appropriate IR output power (see the *Powering the IR LED* section). Note that the transmitter does not have an automatic shutoff circuit, so pay special attention to component power dissipation in high-duty-cycle transmit schemes.

#### \_Applications Information

#### **IR LED and PIN Photodiode Selection**

The IrDA specification calls for an IR transmitter with a peak wavelength between 850nm and 900nm. Within a  $\pm 15^{\circ}$  half-cone-angle, the output intensity of the IR LED must be between 40mW/sr and 500mW/sr. Outside a  $\pm 30^{\circ}$  half-cone-angle, the output intensity of the IR LED must fall below 40mW/sr. The optical rise and fall times of the IR LED must be less than 600ns. Based on these system requirements, the Hewlett Packard HSDL-4220 or the Temic TSHF5400 IR LEDs are two appropriate choices.

Appropriate PIN photodiode selection is extremely important to system performance. The PIN diode must generate at least 200nA (minimum sensitivity of the MAX3120) of current when aimed  $\pm 15^{\circ}$  off-axis with an incident irradiance of  $4\mu$ W/cm<sup>2</sup>. Use the following equation to determine if the Temic BPV22NF meets these requirements:

#### $I_{\text{PIN}} = (4\mu \text{W/cm}^2)(0.075 \text{cm}^2)(1.8)(0.95)^2(0.6\text{A/W})$ = 291nA

The first term  $(4\mu W/cm^2)$  is the minimum guaranteed irradiance in the ±15° angular range. The second term  $(0.075cm^2)$  is the effective sensitive area of the PIN diode. The factor of 1.8 accounts for the efficiency increase due to the spherical lens. The first 0.95 factor normalizes the sensitivity to the 875nm wavelength, while the second 0.95 factor adjusts for decreased receiver efficiency at ±15° off-axis. The last term, 0.6A/W, is the sensitivity of the PIN diode. In this example, the Temic BPV22NF is an appropriate selection.

The final important factor in selecting a PIN diode is effective diode capacitance. It is important to keep this capacitance below 70pF at 1.2V reverse bias. Higher input capacitance can compromise system noise performance by increasing the noise gain of the input transimpedance amplifier.

#### Powering the IR LED

Set the current in the IR LED using an external resistor. Consult the IR LED manufacturer's data sheet to select a forward current that will meet IrDA specifications discussed in the *IR LED and PIN Photodiode Selection* section. Look up the drop across the LED (V<sub>LED</sub>) and the drop across the MAX3120 LED driver (see *Typical Operating Characteristics* - V<sub>LEDC</sub>) and choose the current-setting resistor based on the following equation:

$$R_{SET} = \frac{V_{CC} - V_{LED} - V_{LEDC}}{I_{SET}}$$

Using the Hewlett Packard HSDL-4220 IR LED as an example, V<sub>CC</sub> = 5V, I<sub>SET</sub> = 100mA, and V<sub>LED</sub> = 1.67V, therefore:

$$V_{LEDC} = 0.08V$$
  
RSET =  $32.5\Omega$ 

# MAX3120

## Low-Profile, 3V, 120µA, IrDA Infrared Transceiver

Power-dissipation requirements of the MAX3120, IR LED, and  $R_{SET}$  must be met based on maximum duty cycle and output current requirements.

MAX3120 Power Dissipation = ISET • VLEDC • Duty Cycle

IR LED Power Dissipation = ISET • VLED • Duty Cycle

RSET Power Dissipation =  $ISET^2 \cdot RSET \cdot Duty Cycle$ 

#### **Power-Supply Noise Rejection**

Because of the extremely sensitive nature of photodiode amplifiers, it is important to maintain a quiet supply voltage. Use a separate analog supply voltage where possible. Place a 1µF ceramic bypass capacitor as close to the V<sub>CC</sub> pin as possible. In especially noisy systems, connect a small (10 $\Omega$ ) resistor in series with V<sub>CC</sub>, in addition to the normal bypass capacitor.

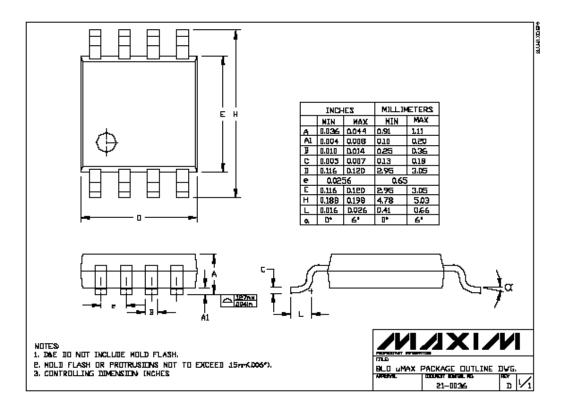
#### Layout Considerations

The MAX3120 requires careful layout techniques to minimize parasitic signal coupling to the PINC input. Keep the lead length between the photodiode and PINC as short as possible. Be sure to keep PC board traces to the PIN diode separate from other noisy traces. To minimize coupling, run the AGND trace adjacent to the PINC trace on both sides. To prevent oscillation, avoid routing the RXD signal near the PINC signal. Connect the anode of the PIN diode, the GND pin, and the supply bypass capacitor pin in a star-ground connection. Connect PGND and GND together. Reduce the output trace length from RXD as much as possible to minimize coupling back to the input via parasitic capacitance.

#### Chip Information

TRANSISTOR COUNT: 256

#### Package Information



Package Information (continued) Π П П ΕН ŧ. 0°-8° INCHES MILLIMETERS INCHES MILLIMETERS MIN MAX MIN MAX MIN MAX MIN | MAX | N | MS012 5.00 8 1.35 1.75 0.10 0.25 А 0.053 0.069 D 0.189 0.197 4.80 А D 0.337 0.344 8.55 8.75 14 A1 0.004 0.010 В B 0.014 0.019 0.35 0.49 D 0.386 0.394 9.80 10.00 16 С С 0.007 0.010 0.19 0.25 0.050 1.27 е NDTES: NDTES: 1. D&E DD NDT INCLUDE MDLD FLASH 2. MDLD FLASH DR PRDITRUSIDNS NDT TD EXCEED .15mm (.006°) 3. LEADS TD BE COPLANAR WITHIN .102mm (.004°) 4. CONTRDLLING DIMENSIDN: MILLIMETER 5. MEETS JEDEC MS012-XX AS SHDWN IN ABDVE TABLE 6. N = NUMBER DF PINS E 0.150 0.157 3.80 4.00 H 0.228 0.244 5.80 6.20 0.25 0.50 h 0.010 0.020 L 0.016 0.050 0.40 1.27 ////XI/// PACKAGE FAMILY DUTLINE: SDIC .150" 21-0041 A

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