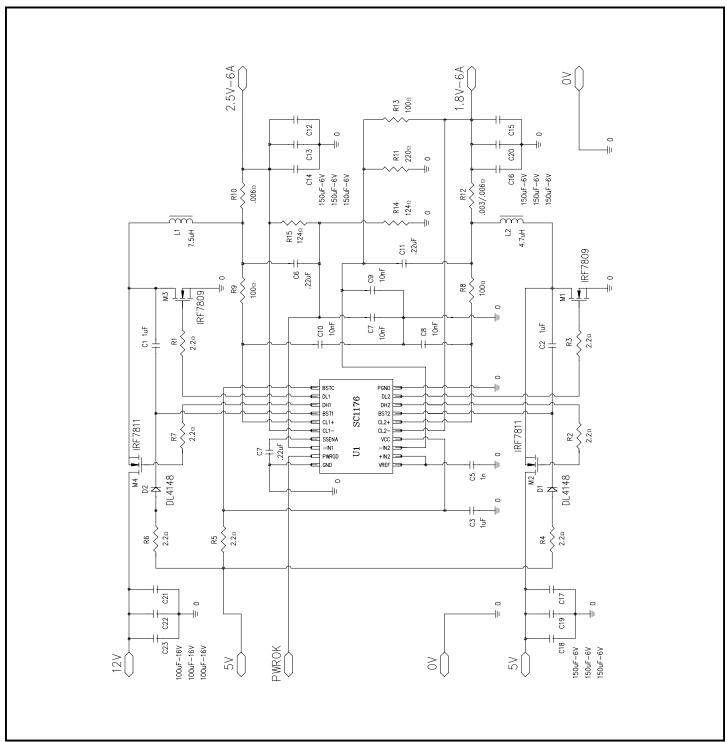


### Typical Application Circuit

### 2 Channels with Current Sharing





## **Absolute Maximum Rating**

Exceeding the specifications below may result in permanent damage to the device, or device malfunction. Operation outside of the parameters specified in the Electrical Characteristics section is not implied.

Parameter	Symbol	Limits	Units
V <sub>cc</sub> to GND	V <sub>IN</sub>	-0.3 to 15	V
PGND to GND		± 1	V
BST to GND		-0.3 to 26	V
Thermal Resistance Junction to Case SOIC TSSOP	$\theta_{\sf JC}$	24 27	°C/W
Thermal Resistance Junction to Ambient SOIC TSSOP	$\theta_{JA}$	65 85	°C/W
Operating Ambient Temperature Range	T <sub>A</sub>	0 to 70	°C
Operating Junction Temperature Range	T <sub>J</sub>	0 to 125	°C
Storage Temperature Range	T <sub>STG</sub>	-65 to +150	°C
Lead Temperature (Soldering) 10 sec	T <sub>LEAD</sub>	300	°C

### **Electrical Characteristics**

Unless Specified:  $V_{CC}$  = 4.75 to 5.25V, GND = PGND = 0V, FB =  $V_{CC}$ , 0mV < (CS(+) - (CS(-)) < 60mV ,  $T_{CC}$  = 25°C

Parameter	Conditions	Min	Тур	Max	Units
Output Voltage	I <sub>O</sub> = 2A <sup>(1)</sup> , V <sub>OUT</sub> set to 2.75V	2.65	2.75	2.85	V
Supply Voltage	V <sub>cc</sub>	4.2		15	V
Supply Current	V <sub>CC</sub> = 5.0		15		mA
Reference Voltage	$I_{REF} = 0\mu A$	0.89	0.9	0.91	V
Reference Voltage Line Regulation	5V < V <sub>cc</sub> < 15V			0.5	%
Reference Voltage Load Regulation	$I_{REF} = 0\mu A \text{ to } 30\mu A$			-12	mV
Output Line Regulation	5V < V <sub>IN</sub> < 15V			0.5	%
Output Load Regulation	$I_{\odot} = 0.3A \text{ to } 15A^{(1)}$		1		%
Gain (A <sub>OL</sub> )	V <sub>OSENSE</sub> to V <sub>O</sub>		35		dB
Current Limit Voltage		60	70	80	mV
Oscillator Frequency		270	300	330	kHz
Oscillator Max Duty Cycle		90	95		%
DH Sink Current	DH - PGND = 3.0V	1.3			Α
DH Sink Current	DH - PGND = 2.0V	0.85			А
DH Source Current	BSTH - DH = 3.5V	1.3			Α
DH Source Current	BSTH - DH = 2.5V	0.85			Α



### Electrical Characteristics (Cont.)

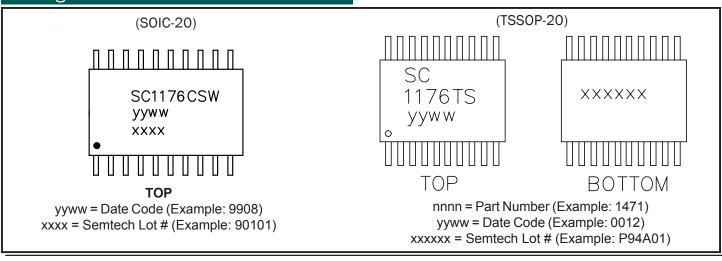
Unless Specified:  $V_{CC}$  = 4.75 to 5.25V, GND = PGND = 0V, FB =  $V_{O}$ , 0mV < (CS(+) - (CS(-)) < 60mV ,  $T_{J}$  = 25°C

PARAMETER	CONDITIONS	MIN	TYP	MAX	UNITS
DL Sink Current	DL - PGNG = 3.0V	1.3			Α
DL Sink Current	DL - PGND = 2.0V	0.85			Α
DL Source Current	BSTL - DL = 3.5V	1.3			Α
DL Source Current	BSTL - DL = 2.5V	0.85			Α
Dead Time	Note 5	50	100	150	ns
Soft Start Charge Current (2)			25		μA
Soft Start Enable	0% duty cycle		1.4		V
Soft Start End	100% duty cycle		2.5		V
Soft Start Transition <sup>(2)</sup>	Synchronous mode		3.3		V
Power Good Window <sup>(3)</sup>			+10		%V <sub>out</sub>
Fold Back Current	V <sub>OUT</sub> = 0V		50%		I <sub>LIM</sub>
Fold Back Voltage Knee	I = <sub>ILIM</sub>	0.9		V <sub>out</sub>	V
Input Bias Current	-IN1, +IN2, -IN2			1	μA
BSTH to DH Leakage	BSTH - DH = 5V			10	μA
DH to PGND Leakage	DH - PGND = 5V			10	μA
BSTL to DL Leakage	BSTL - DL = 5V			10	μA
DL to PGND	DL - PGND = 5V			10	μΑ

#### Notes:

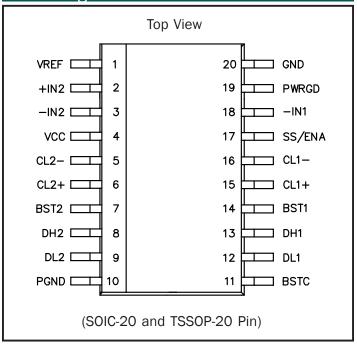
- (1) Specification refers to application circuit.
- (2) The soft start pin sources 25µA to an external capacitor. The converter operates in synchronous mode above the soft start transition threshold and in asynchronous mode below it.
- (3) Power good is an open collector pulled low when the output voltage is outside the ±10% window.
- (4) This device is ESD sensitive. Use of standard ESD handling precautions is required.
- (5) 200ns maximum at 75°.

### Marking Information





# Pin Configuration



Ordering Information					
Device <sup>(1)</sup>	Package <sup>(2)</sup>				
SC1176CSWTR	0010.00				
SC1176CSWTRT	SOIC-20				
SC1176ASWTR					
SC1176ASWTRT					
SC1176TSTR	T000D 00				
SC1176TSTRT	TSSOP-20				
SC1176ATSTR					
SC1176EVB-1	Current Share Version Evaluation Board				
SC1176-2EVB-2	Dual Channel Version Evaluation				

#### Notes:

- (1) Only available in tape and reel packaging. A reel contains 1000 (SOIC) or 2500 (TSSOP) devices.
- (2) Lead free product. This product is fully WEEE and RoHS compliant.

### Pin Descriptions

#### **EXPANDED PIN DESCRIPTION**

Pin 1: (VREF)

Internal .9V reference

Connected to the + input of the master channel error amplifier.

Pin 2: (+IN2)

+ Input of slave channel error amplifier.

Connected to .9V reference (Pin 1) for

2 independent channel configuration.

**Pin 3, 18:** (-IN2, -IN1)

- Inputs of close loop error amplifiers.

Works as a feedback inputs (For both modes).

Pin 4: (VCC)

VCC chip supply voltage.

15V maximum, 15mA typical.

Needs a 1µF ceramic multilayer decoupling capacitor to GND (Pin 20).

Pin 5, 6,15, 16: (CL2-, CL2+, CL1+, CL1-)

Pins (-) and (+) of the current limit amplifiers for both channels.

Connected to output current sense resistors. Compares that sense voltage to internal 75mV reference. Needs RC filter for noise rejection.

Pin 7, 14: (BST2, BST1)

BST signal. Supply for high side driver.

Can be connected to a high enough voltage source.

Usually connected to bootstrap circuit.

**Pin 8, 13:** (DH2, DH1)

DH signal (Drive High).

Gate drive for top MOSFETs.

Requires a small serie resistor.

Pin 9, 12: (DL2, DL1)

DL signal (Drive Low).

Gate drive for bottom MOSFETs.

Requires a small serie resistor.

**Pin 10:** (PGND)

Power GND. Return of gate drive currents.

**Pin 11:** (BSTC)

Supply for bottom MOSFETs gate drive.

**Pin 17:** (SS/ENA)

Soft start pin. Internal current source connected to

external capacitor.

Inhibits he chip if pulled down.

Pin 19: (PWRGD)

Power good signal.

Open collector signal.

Turns to 0 if output voltage is outside the power good window.

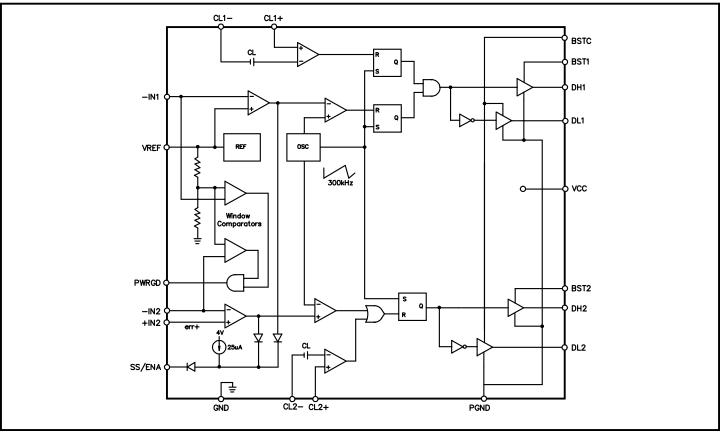
Pin 20: (GND)

Analog GND.

Return of analog signals and bias of chip.



### **Block Diagram**



### **NOTES**

- (1) Block 1 (top) is the Master and Block 2 (bottom) is the Slave in current sharing configuration.
- (2) For independent operation there is no Master or Slave.

### Applications Information - Theory of Operation

### Main Loop(s)

The SC1176 is a dual, voltage mode synchronous Buck controller, the two separate channels are identical and share only IC supply pins (Vcc and GND), output driver ground (PGND) and pre-driver supply voltage (BSTC). They also share a common oscillator generating a sawtooth waveform for channel 1 and an inverted sawtooth for channel 2. Each channel has its own current limit comparator. Channel 1 has the positive input of the error amplifier internally connected to Vref. Channel 2 has both inputs of the error amplifier uncommitted and available externally. This allows the SC1176 to operate in two distinct modes.

a) Two independent channels with either common or different input voltages and different output voltages. The two channels each have

their own voltage feedback path from their own output. In this mode, the positive input of error amplifier 2 is connected externally to Vref. If the application uses a common input voltage, the sawtooth phase shift between the channels provides some measure of input ripple current cancellation.

b) Two channels operating in current sharing mode with common output voltage and either common input voltage or different input voltages. In this mode, channel 1 operates as a voltage mode Buck controller, as before, but error amp 2 monitors and amplifies the difference in voltage across the output current sense resistors of channel 1 and channel 2 (Master and Slave) and adjusts the Slave duty cycle to match output currents. Because of finite gain and offsets in the loop, the resistor ratio for perfect current match-



### Applications Information - Theory of Operation

ing is not 1:1. The Master and Slave channels still have their own current limits, identical to the independent channel case.

#### **Power Good**

The controller provides a power good signal. This is an open collector output, which is pulled low if the output voltage is outside of the power good window.

### Soft Start/Enable

The Soft Start/Enable (SS/ENA) pin serves several functions. If held below the Soft Start Enable threshold, both channels are inhibited. DH1 and DH2 will be low, turning off the top FETs. Between the Soft Start Enable threshold and the Soft Start End threshold, the duty cycle is allowed to increase. At the Soft Start End threshold, maximum duty cycle is reached. In practical applications the error amplifier will be controlling the duty cycle before the Soft Start End threshold is reached. To avoid boost problems during startup in current share mode, both channels start up in asynchronous mode, and the bottom FET body diode is used for recirculating current during the FET off time. When the SS/ENA pin reaches the Soft Start Transition threshold, the channels begin operating in synchronous mode for improved efficiency. The soft start pin sources approximately 25uA and soft start timing can be set by selection of an appropriate soft start capacitor value.

#### SENSE RESISTOR SELECTION

#### **Current Sharing Mode**

Calculation of the three programming resistors to achieve sharing.

Three resistors will determine the current sharing load line.

First the offset resistor will ensure that the load line crosses the origin (O Amp on each channel) for sharing at light current. A pull up resistor from the 5V bias ( $V_{\rm cc}$  of the chip) will be used. For low duty cycle on the slave channel (below 50%), the pull up will be on pin 3. For high duty cycle on the slave channel (above 50%), the pull up will be on pin 2.

The formula is:

$$R_{\text{pull-up}} (K\Omega) = 2.1 \text{ X} \frac{5 - V_{\text{OUT}}}{.5 - \frac{V_{\text{OUT}} + .1}{V_{\text{SLAVE}}}}$$

 $100\Omega$  being the value of the resistors connecting the pins 2 and 3 to the two output sense resistors.

.1 V is an estimated voltage drop across the MOSFETs.

Positive values go to pin 3, negative to pin 2.

A +20K will be a 20K on pin 3.

A -20K will be a 20K on pin 2.

Now that the offset resistor has been fixed, we need to set up the maximum current for each channel.

Selection of  $R_{SENSE}$  1 for the master channel: (in m ohm)

 $R_{SENSE} 1 = 72 \text{mV} / I \text{ max master}$ 

Selection of  $R_{SENSE}$  2 for the slave channel: (in m ohm)

 $R_{SENSE} 2 = 72 \text{mV} / I \text{ max slave}$ 

The errors will be minimized if the power components have been sized proportionately to the maximum currents.

### **Independent Channels**

Calculation of the two current limiting resistors. There is no need for an offset resistor in the independent channels mode, only the two sense resistors are used:

Selection of R<sub>SENSE</sub> 1 for the channel 1: (in m ohm)

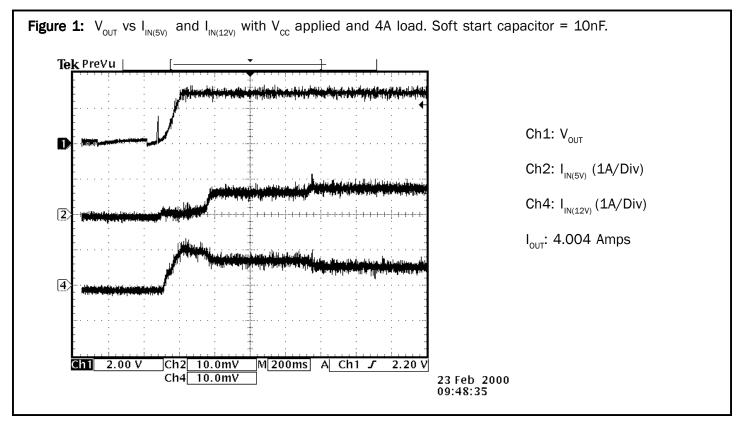
$$R_{SENSE} 1 = 72mV / I max ch 1$$

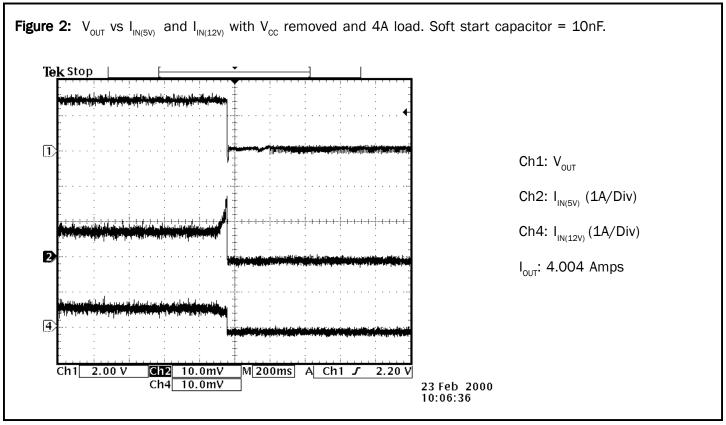
Selection of R<sub>SENSE</sub> 2 for the channel 2: (in m ohm)

$$R_{SENSE} 1 = 72 \text{mV} / 1 \text{ max ch } 2$$



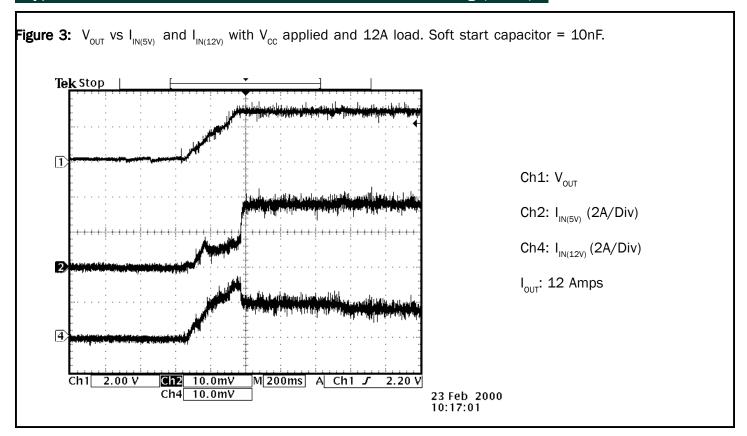
## Typical Characteristics - 2 Channels with Current Sharing

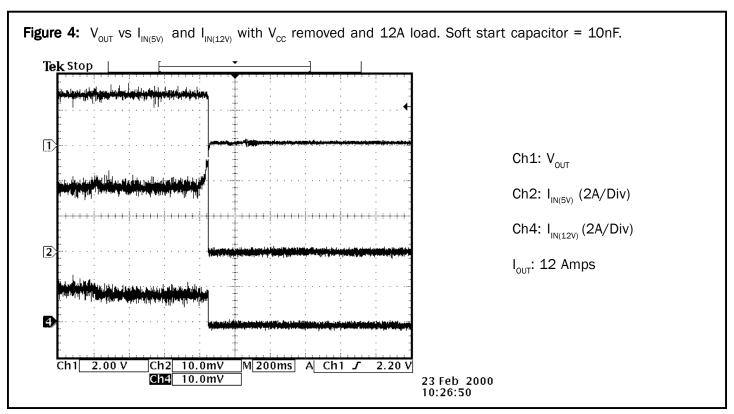






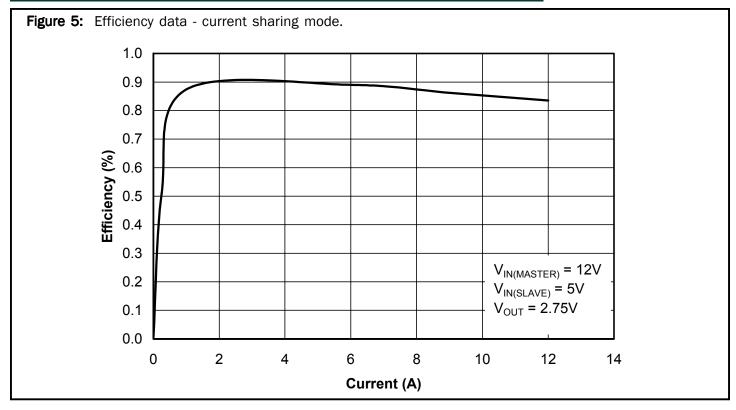
# Typical Characteristics - 2 Channels with Current Sharing (Cont.)







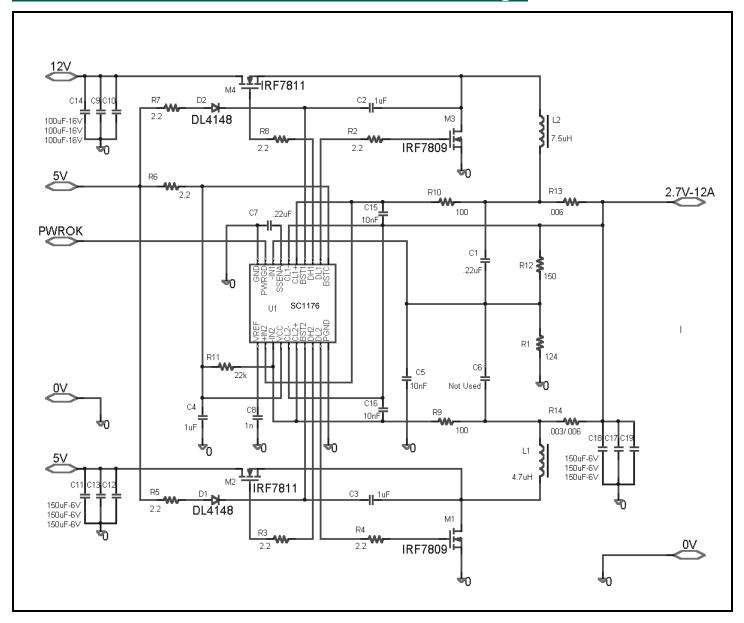
# Typical Characteristics - 2 Channels with Current Sharing (Cont.)



The Current Sharing Evaluation Board is not intended for a specific application. The power components are not optimized for minimum cost and size. This evaluation board should be used to understand the operation of the SC1176. To design with SC1176 for specific current sharing applications. Please refer to: Application note ANO0-3.



# Evaluation Board Schematic - 2 Channel with Current Sharing



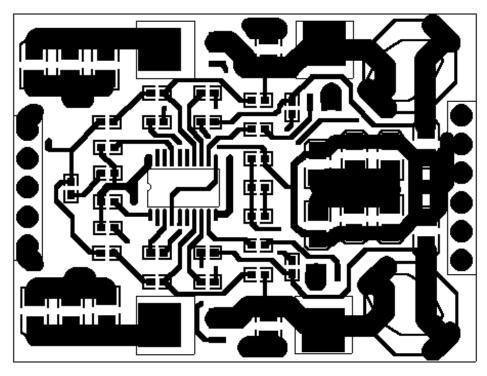


# Evaluation Board Bill of Materials - 2 Channels with Current Sharing

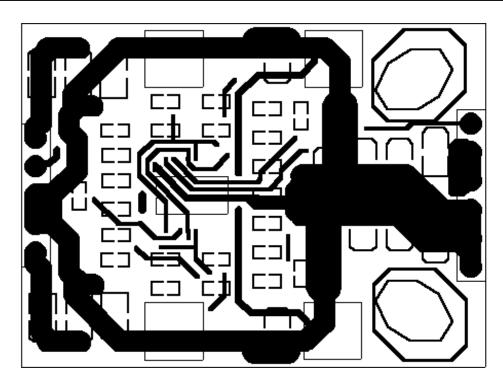
Item	Quantity	Reference	Part
1	2	C1,C7	.22uF, 50V
2	3	C2,C3,C4	1uF, 50V
3	3	C5,C15,C16	10nF, 50V
4	1	C8	1nF, 50V
5	3	C9,C10,C14	100uF, 6V
6	6	C11,C12,C13,C17,C18,C19	150uF, 16V
7	2	D1,D2	DL4148
8	1	L1	7.5uH, 8A
9	1	L2	4.7uH, 8A
10	2	M1,M3	IRF7809 or FDB7030
11	2	M2,M4	IRF7811 or FDB7030
12	1	R1	124
13	7	R2,R3,R4,R5,R6,R7,R8	2.2
14	2	R9,R10	100
15	1	R12	150
16	1	R13	.006
17	1	R14	.003
18	1	U1	SC1176



# Evaluation Board Gerber Plots - 2 Channels with Current Sharing



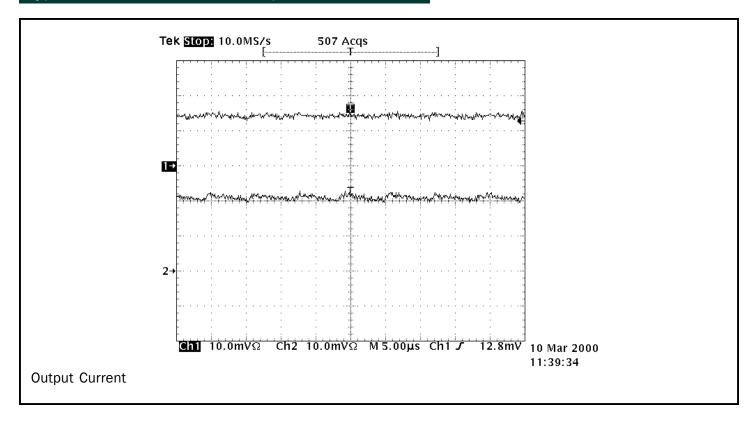
Top Side Traces

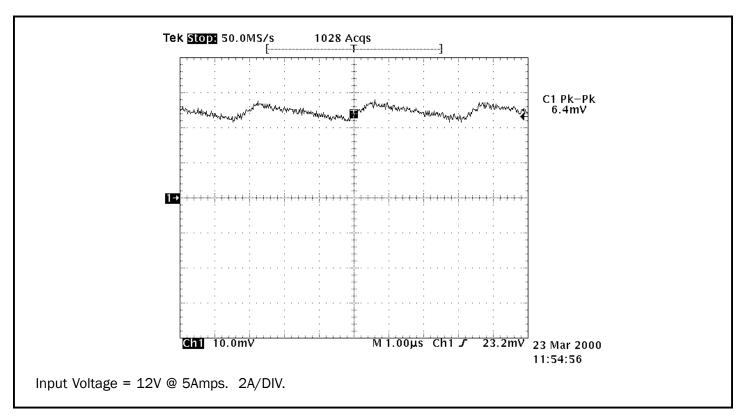


**Bottom Side Traces** 



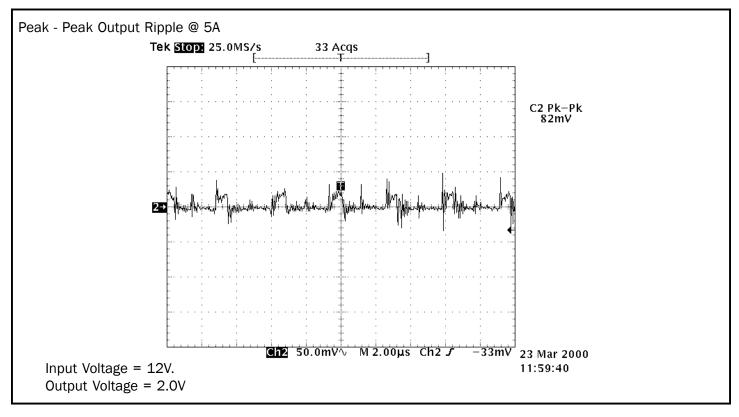
## Typical Characteristics - 2 Independent Channels

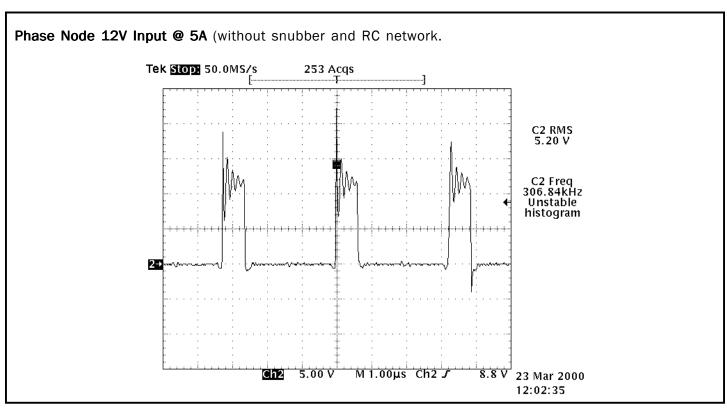






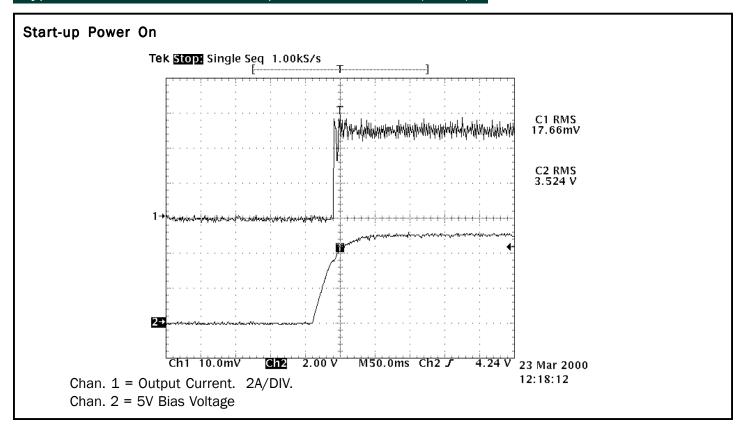
## Typical Characteristics - 2 Independent Channels (Cont.)

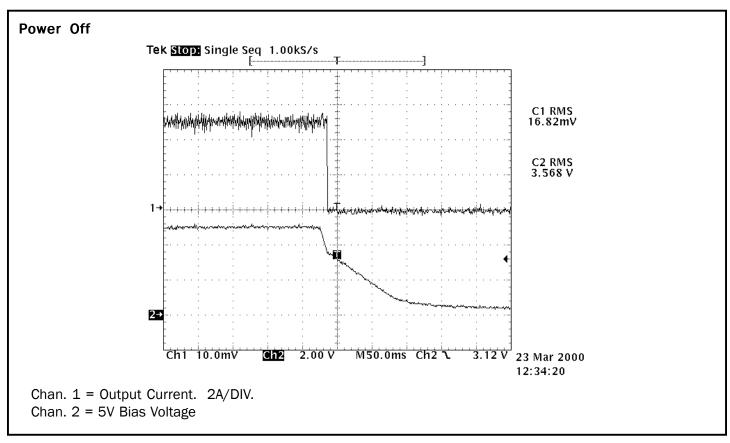






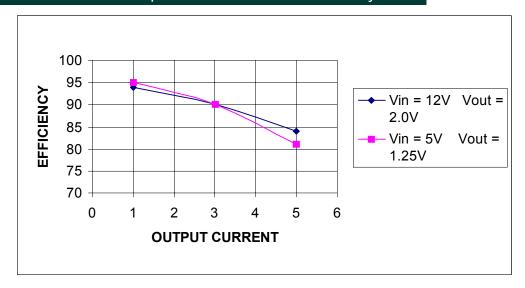
## Typical Characteristics - 2 Independent Channels (Cont.)







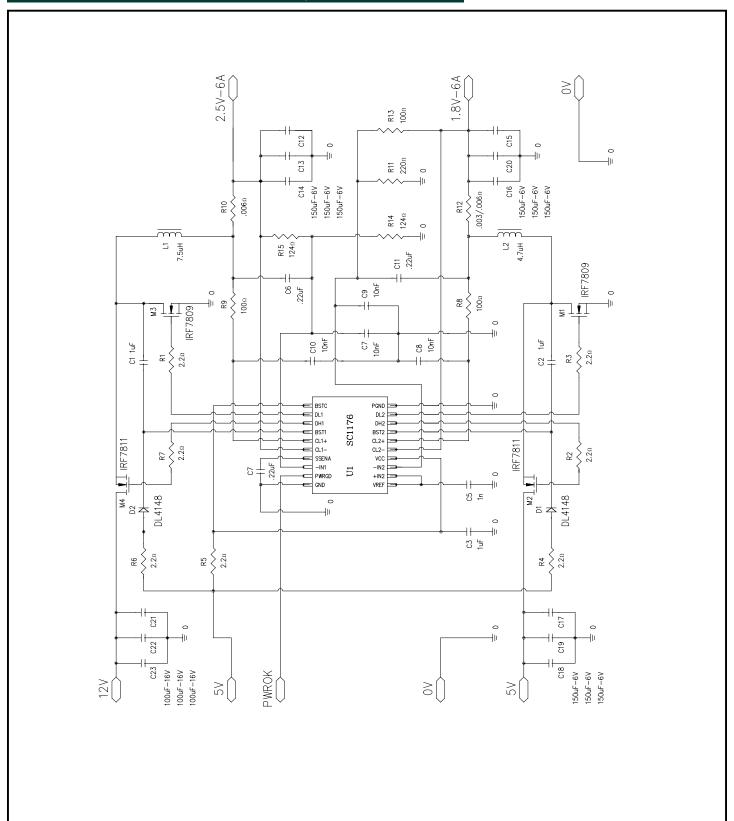
# Typical Characteristics - 2 Independent Channels Efficiency Test



The Independent Channels Evaluation Board is not intended for a specific application. The power components are not optimized for minimum cost and size. This evaluation board should be used to understand the operation of the SC1176. To design with the SC1176 for specific independent channels applications. Please refer to: Application note ANO0-4.



# Evaluation Board Schematic - 2 Independent Channels



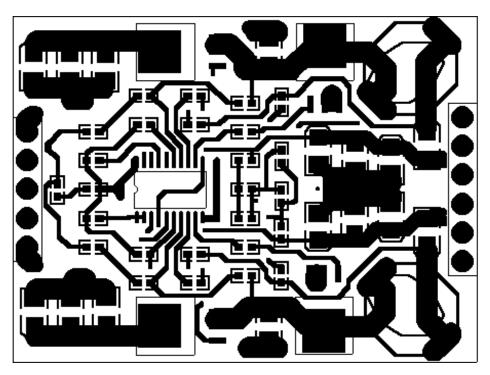


# Evaluation Board Bill of Materials - 2 Independent Channels

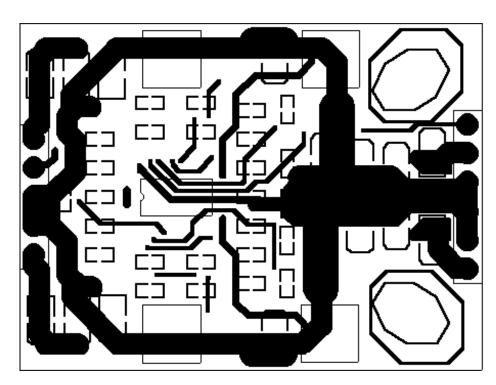
Item	Quantity	Reference	Part
1	3	C1,C2,C3	1uF, 50V
2	3	C4,C6,C11	.22uF, 50V
3	1	C5	1nF, 50V
4	4	C7,C8,C9,C10	10nF, 50V
5	9	C12,C13,C14,C15,C16,C17,C18,C19,C20	150uF, 6V
6	3	C21,C22,C23	100uF, 16V
7	2	D1,D2	DL4148
8	1	L1	7.5uH, 8A
9	1	L2	4.7uH, 8A
10	2	M1,M3	IRF7809 or FDB7030
11	2	M2,M4	IRF7811 or FDB7030
12	7	R1,R2,R3,R4,R5,R6,R7	2.2
13	3	R8,R9,R13	100
14	1	R10	.006
15	1	R11	220
16	1	R12	.003
17	2	R14,R15	124
18	1	U1	SC1176



# Evaluation Board Gerber Plots - 2 Independent Channels



Top Side Traces



**Bottom Side Traces** 



### Layout Guidelines

Power and signal traces must be kept separated for noise considerations. Feedback, current sense traces and analog ground should not cross any traces or planes carrying high switching currents, such as the input loop or the phase node.

The input loop, consisting of the input capacitors and both MOSFETs must be kept as small as possible. All of the high switching currents occur in this loop. The enclosed loop area must be kept small to minimize inductance and radiated and conducted emissions. Designing for minimum trace length is not always the best approach, often a more optimum layout can be achieved by keeping loop area constraints in mind.

It is important to keep gate lengths short, the IC must be close to the power switches. This is more difficult in a dual channel device than a single and requires that the two power paths run on either side of a centrally located controller.

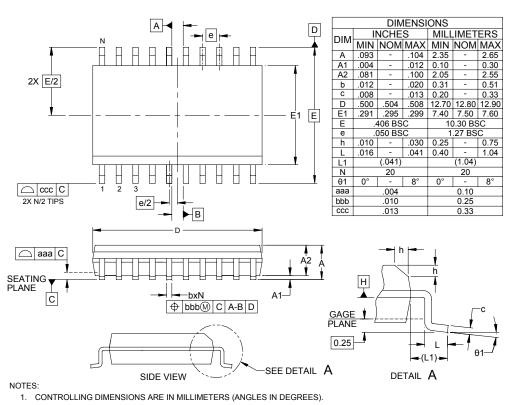
Grounding requirements are always conflicting in a buck converter, especially at high power, and the trick is to achieve the best compromise. Power ground (PGND) should be returned to the bottom MOSFET source to provide the best gate current return path. Analog ground (GND) should be returned to the ground side of the output capacitors so that the analog circuitry in the controller has an electrically quiet reference and to provide the greatest feedback accuracy. The problem is that the differential voltage capability of the two IC grounds is limited to about 1V for proper operation and so the physical separation between the two grounds must also be minimized. If the grounds are too far apart, fast current transitions in the connection can generate voltage spikes exceeding the 1V capability, resulting in unstable and erratic behavior.

The feedback divider must be close to the IC and be returned to analog ground. Current sense traces must be run parallel and close to each other and to analog ground.

The IC must have a ceramic decoupling capacitor across its supply pins, mounted as close to the device as possible. The small ceramic, noise-filtering capacitors on the current sense lines should also be placed as close to the IC as possible.

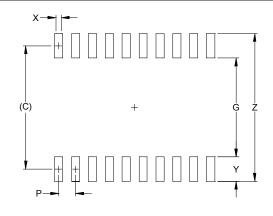


### Outline Drawing - SOIC-20



- 2. DATUMS -A- AND -B- TO BE DETERMINED AT DATUM PLANEH-
- 3. DIMENSIONS "E1" AND "D" DO NOT INCLUDE MOLD FLASH, PROTRUSIONS OR GATE BURRS.
- 4. REFERENCE JEDEC STD MS-013, VARIATION AC.

### Land Pattern - SOIC-20



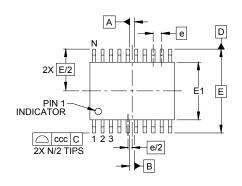
DIMENSIONS				
DIM	INCHES	MILLIMETERS		
С	(.362)	(9.20)		
G .276		7.00		
P .050		1.27		
X .024		0.60		
Υ	.087	2.20		
Z .449		11.40		

### NOTES:

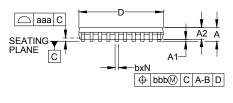
- THIS LAND PATTERN IS FOR REFERENCE PURPOSES ONLY.
  CONSULT YOUR MANUFACTURING GROUP TO ENSURE YOUR
  COMPANY'S MANUFACTURING GUIDELINES ARE MET.
- 2. REFERENCE IPC-SM-782A, RLP NO. 307A.

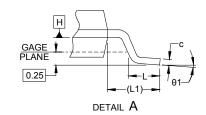


## Outline Drawing - TSSOP-20



	DIMENSIONS					
DIM	11	NCHE	S	MILL	IMET	ERS
ווווטן	MIN	NOM	MAX	MIN	NOM	MAX
Α	-	-	.047	-	-	1.20
A1	.002	-	.006	0.05	-	0.15
A2	.031	-	.042	0.80	-	1.05
b	.007	-	.012	0.19	-	0.30
С	.003	-	.007	0.09	-	0.20
D	.251	.255	.259	6.40	6.50	6.60
E1	.169	.173	.177	4.30	4.40	4.50
Е		252 BS		6.40 BSC		
е	.(	)26 BS	С	0	.65 BS	
L	.018	.024	.030	0.45	0.60	0.75
L1	(.039)				(1.0)	
N	20				20	
θ1	0	-	8	0	-	8
aaa	.004				0.10	
bbb	.004			0.10		
ccc	.008				0.20	



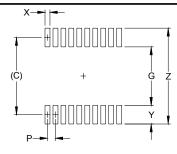




#### NOTES:

- 1. CONTROLLING DIMENSIONS ARE IN MILLIMETERS (ANGLES IN DEGREES).
- 2. DATUMS -A- AND -B- TO BE DETERMINED AT DATUM PLANEH-
- 3. DIMENSIONS "E1" AND "D" DO NOT INCLUDE MOLD FLASH, PROTRUSIONS OR GATE BURRS.
- 4. REFERENCE JEDEC STD MO-153, VARIATION AC.

## Land Pattern - TSSOP-20



	DIMENSIONS				
DIM	INCHES	MILLIMETERS			
С	(.222)	(5.65)			
G	.161	4.10			
Р	.026	0.65			
Х	.016	0.40			
Υ	.061	1.55			
Ζ	.283	7.20			

#### NOTES:

THIS LAND PATTERN IS FOR REFERENCE PURPOSES ONLY.
CONSULT YOUR MANUFACTURING GROUP TO ENSURE YOUR
COMPANY'S MANUFACTURING GUIDELINES ARE MET.

### **Contact Information**

Semtech Corporation Power Management Products Division 200 Flynn Road, Camarillo, CA 93012 Phone: (805)498-2111 FAX (805)498-3804

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