ABSOLUTE MAXIMUM RATINGS

(All voltages referenced to GND, unless otherwise noted.)

Supply Voltage (V _{CC})+7V Cable Ground (ISOCOM)±75V Isolated Supply ISOVCC Relative to Cable Ground
(ISOCOM)+7V
Digital Input, Output Voltage
(DI, DE, RE, TXP, RXP, RO)0.3V to (V _{CC} + 0.3V)
Digital Inputs (H/F) Relative to Cable Ground
(ISOCOM)0.3V to (ISOVCC + 0.3V)
Driver Output Voltage (Y, Z) Relative to Cable Ground
(ISOCOM)8V to +12.5V
Receiver Input Voltage (A, B) Relative to Cable Ground
(ISOCOM)8V to +12.5V
Termination Connections (YR, ZR, AR, BR) Relative to Cable
Ground (ISOCOM)8V to +12.5V

Charge-Pump Capacitance Low
(C1LO, C2LO)0.3V to (V _{CC} + 0.3V) Charge-Pump Capacitance High (C1HI, C2HI) Relative
to Cable Ground (ISOCOM)
Continuous Power Dissipation ($T_A = +70^{\circ}C$)
28-Pin SSOP (derate 10.8mW/°C above +70°C)860mW
28-Pin PDIP (derate 14.3mW/°C above +70°C)1143mW
Operating Temperature Ranges
MAX3158C_I0°C to +70°C
MAX3158E_I40°C to +85°C
Junction Temperature+150°C
Storage Temperature Range65°C to +160°C
Lead Temperature (soldering, 10s)+300°C

Stresses beyond those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

ELECTRICAL CHARACTERISTICS

(V_{CC} = +5V ±5%, YR = ZR = AR = BR = ISOCOM, C1 = C2 = 0.047 μ F, C3 = C4 = 1 μ F, TA = T_{MIN} to T_{MAX}, unless otherwise noted. Typical values are at V_{CC} = +5V and T_A = +25°C.) (Note 1)

PARAMETER	SYMBOL	CONDITIONS	MIN	ТҮР	MAX	UNITS
DRIVER						
Differential Driver Output, No Load	V _{OD1}	Figure 1, R = $10M\Omega$			6	V
Differential Driver Output,	Vene	Figure 1, R = 50Ω (RS-422)	2.0			V
Loaded	V _{OD2}	Figure 1, R = 27Ω (RS-485)	1.5			
Change in Magnitude of Differential Output Voltage (Note 2)	ΔV _{OD}	Figure 1, R = 50Ω or R = 27Ω			0.2	V
Driver Common-Mode Voltage	Voc	Figure 1, R = 50Ω or R = 27Ω , V _{OC} relative to ISOCOM			3.5	V
Change in Magnitude of Common-Mode Voltage (Note 2)	ΔV _{OC}	Figure 1, R = 50Ω or R = 27Ω			0.2	V
Input High Voltage	VIH	DE, DI, $\overline{\text{RE}}$, TXP, RXP, relative to GND, H/ $\overline{\text{F}}$ relative to ISOCOM	2.0			V
Input Low Voltage VIL		DE, DI, $\overline{\text{RE}}$, TXP, RXP, relative to GND, H/ $\overline{\text{F}}$ relative to ISOCOM			0.8	V
DI Input Hysteresis	V _{HYS}			100		mV

ELECTRICAL CHARACTERISTICS (continued)

(V_{CC} = +5V ±5%, YR = ZR = AR = BR = ISOCOM, C1 = C2 = 0.047 μ F, C3 = C4 = 1 μ F, T_A = T_{MIN} to T_{MAX}, unless otherwise noted. Typical values are at V_{CC} = +5V and T_A = +25°C.) (Note 1)

PARAMETER	SYMBOL	CONDITIONS	MIN	ТҮР	MAX	UNITS	
Input Current	I _{IN1}	DE, DI, RE			±2		
Input Current	I _{IN2}	H/F, TXP, RXP internal pulldown	10		40	μA	
Input Current (A and B)	luur	DE = GND, V _{CC} = GND or 5.25V, -7V \leq (V _{IN} - V _{ISOCOM}) \leq +12V, AR = BR = ISOCOM	-5.0		+8.0		
	l _{IN3}	DE = GND, V _{CC} = GND or 5.25V, -7V \leq (V _{IN} - V _{ISOCOM}) \leq +12V, AR open, BR open	-0.58		+1.0	mA	
Output Leakage (Y and Z) Full Duplex		DE = GND, V _{CC} = GND or 5.25V, -7V \leq (V _{IN} - V _{ISOCOM}) \leq +12V, YR = ZR = ISOCOM	-5.0		+8.0		
	IO .	DE = GND, V _{CC} = GND or 5.25V, -7V \leq (V _{IN} - V _{ISOCOM}) \leq +12V, YR open, ZR open	-0.58		+1.0	mA	
Driver Short-Circuit Output	IOSD1	$V_{ISOCOM} - 7V \le V_{OUT} \le V_{ISOVCC}$	-250			m (
Current (Note 3)	IOSD2	$V_{ISOCOM} - 7V \le V_{OUT} \le V_{ISOCOM} + 12V$			250	mA	
RECEIVER	-		-			_	
Receiver Differential Threshold Voltage	V _{TH}	$-7V \le V_{CM} - V_{ISOCOM} \le +12V$	-200	-125	-50	mV	
Receiver Input Hysteresis	ΔV_{TH}			25		mV	
Receiver Output High Voltage	VOH	$I_O = -4mA$, $V_{ID} = -50mV$	V _{CC} - 1.5			V	
Receiver Output Low Voltage	VOL	$I_{O} = 4mA, V_{ID} = -200mV$			0.4	V	
Three-State Output Current at Receiver	IOZR	$0.4V \le V_O \le 2.4V$			±1	μΑ	
Receiver Input Resistance	_	$-7V \le V_{CM} - V_{ISOCOM} \le +12V$, AR = BR = ISOCOM	1.5				
	R _{IN}	$-7V \le V_{CM} - V_{ISOCOM} \le +12V$, AR open, BR open	12			kΩ	
Receiver Output Short-Circuit Current	IOSR	$0V \le V_{RO} \le V_{CC}$	±7		±100	mA	
SUPPLY CURRENT							
Supply Current	Icc	No load, $\overline{RE} = DE = DI = GND$ or V _{CC}		25	85	mA	
Supply Current in Shutdown		$DE = GND, \overline{RE} = V_{CC}, V_{ISOCOM} = GND$		25	60		
Mode	ISHDN	$DE = GND, \overline{RE} = V_{CC}, V_{ISOCOM} = \pm 60V$			±650	μΑ	
Maximum Ground Differential	ΔV_{GND}	DE = GND, RE = V _{CC} , I ISOCOM leakage I ≤ 650µA	±60			V	

SWITCHING CHARACTERISTICS

 $(V_{CC} = +5V \pm 5\%, YR = ZR = AR = BR = ISOCOM, C1 = C2 = 0.047 \mu F, C3 = C4 = 1 \mu F.$ Typical values are at $V_{CC} = +5V$ and $T_A = +25^{\circ}C.$)

PARAMETER	SYMBOL	CONDITIONS	MIN	ТҮР	MAX	UNITS
Driver Input to Output	t _{DPLH}	Figures 3 and 5, $R_{DIFF} = 54\Omega$, $C_{L1} = C_{L2} = 100 pF$	400	1400	2000	
Driver input to Output	t _{DPHL}	Figures 3 and 5, R _{DIFF} = 54 Ω , C _{L1} = C _{L2} = 100pF	400	1400	2000	ns
Driver Output Skew tDPLH - tDPHL	t DSKEW	Figures 3 and 5, R _{DIFF} = 54 Ω , C _{L1} = C _{L2} = 100pF			250	ns
Driver Rise or Fall Time	t _{DR} , t _{DF}	Figures 3 and 5, $R_{DIFF} = 54\Omega$, $C_{L1} = C_{L2} = 100 pF$	200	1200	2200	ns
Maximum Data Rate	fMAX		250			kbps
Driver Enable to Output High	tdzн	Figures 4 and 6, C_L = 100pF, S2 closed			3500	ns
Driver Enable to Output Low	t _{DZL}	Figures 4 and 6, C_L = 100pF, S1 closed			3500	ns
Driver Disable Time from Low	tDLZ	Figures 4 and 6, $C_L = 15pF$, S1 closed			300	ns
Driver Disable Time from High	t _{DHZ}	Figures 4 and 6, $C_L = 15pF$, S2 closed			300	ns
Receiver Input to Output	t _{RPLH} , t _{RPHL}	Figures 7 and 9; $ V_{ID} \ge 2.0V$		440	600	ns
Differential Receiver Skew tRPLH - tRPHL	^t RSKEW	Figures 7 and 9; $ V_{ID} \ge 2.0V$		20		ns
Receiver Enable to Output Low	t _{RZL}	Figures 2 and 8, C_L = 100pF, S1 closed		20	50	ns
Receiver Enable to Output High	t _{RZH}	Figures 2 and 8, C_L = 100pF, S2 closed		20	50	ns
Receiver Disable Time from Low	t _{RLZ}	Figures 2 and 8, C_L = 100pF, S1 closed		200	500	ns
Receiver Disable Time from High	t _{RHZ}	Figures 2 and 8, C_L = 100pF, S2 closed		200	500	ns
Time to Shutdown	t SHDN	(Note 4)	50	200	700	ns
Driver Enable from Shutdown to Output High	t _{DZH} (SHDN)	Figures 4 and 6, C_L = 15pF, S2 closed		0.2	1	ms
Driver Enable from Shutdown to Output Low	t _{DZL} (SHDN)	Figures 4 and 6, C_L = 15pF, S1 closed		0.2	1	ms
Receiver Enable from Shutdown to Output High	t _{RZH} (SHDN)	Figures 2 and 8, C_L = 100pF, S2 closed		0.2	1	ms
Receiver Enable from Shutdown to Output Low	t _{RZL} (SHDN)	Figures 2 and 8, C_L = 100pF, S1 closed		0.2	1	ms
Charge-Pump Oscillating Frequency	fosc			1.3		MHz

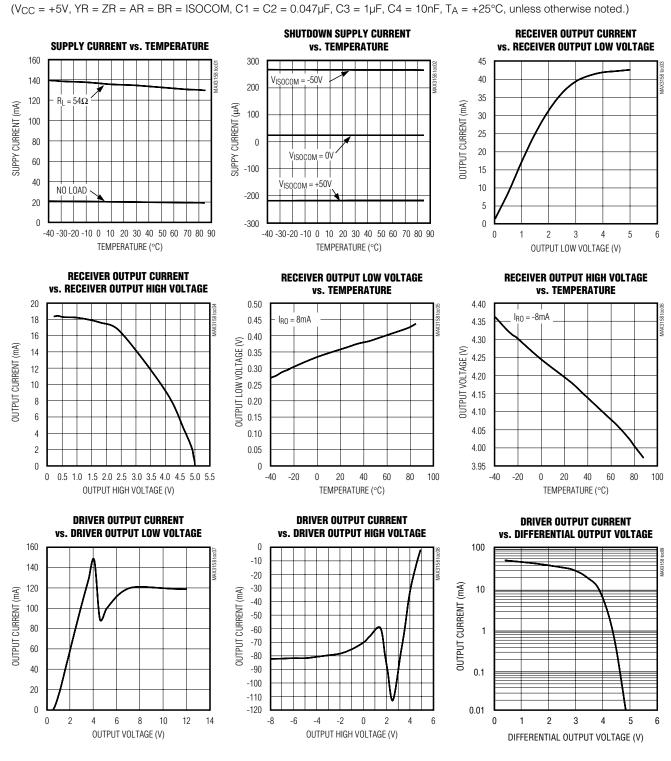
Note 1: All currents into the device are positive; all currents out of the device are negative. All voltages are referred to device ground unless otherwise noted.

Note 2: ΔV_{OD} and ΔV_{OC} are the changes in V_{OD} and V_{OC} , respectively, when the DI input changes state.

Note 3: Current level applies to peak current just prior to foldback-current limiting.

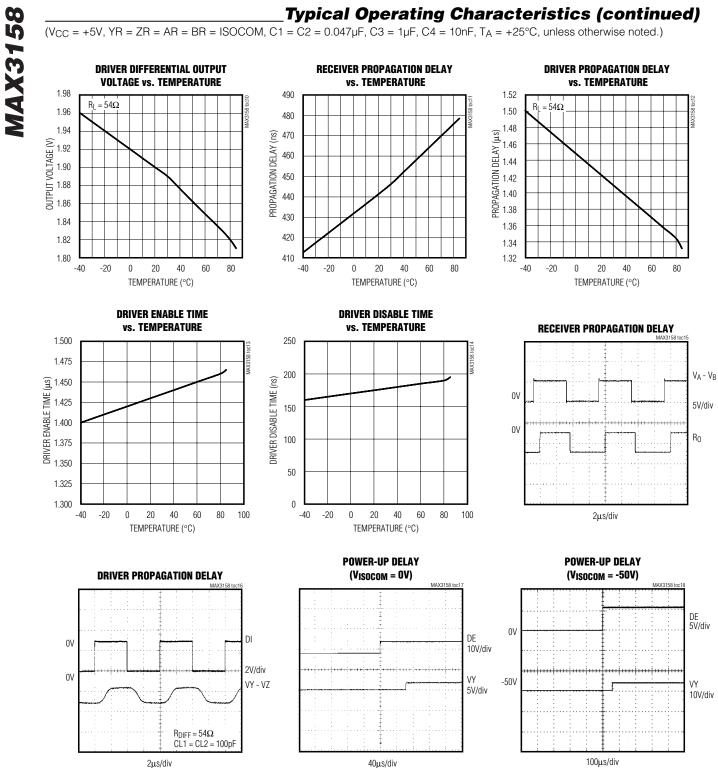
Note 4: The device is put into shutdown by bringing RE high and DE low. If the enable inputs are in this state for less than 50ns, the device is guaranteed not to enter shutdown. If the enable inputs are in this state for at least 700ns, the device is guaranteed to have entered shutdown.

Typical Operating Characteristics



MAX3158

M/IXI/M

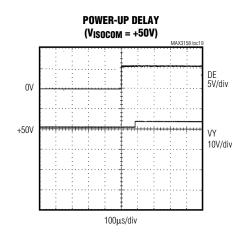


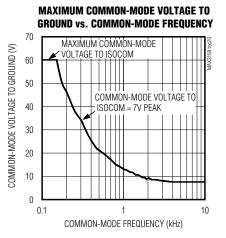
///XI//

6

Typical Operating Characteristics (continued)

 $(V_{CC} = +5V, YR = ZR = AR = BR = ISOCOM, C1 = C2 = 0.047\mu$ F, C3 = 1 μ F, C4 = 10nF, T_A = +25°C, unless otherwise noted.)





$\begin{array}{c} Y \\ \hline \\ V_{OD} \\ Z \\ \hline \\ Z \\ \hline \\ \end{array} \\ \begin{array}{c} R \\ V_{OC} \\ \hline \\ \hline \\ \hline \\ \hline \\ \hline \\ \hline \end{array} \\ \end{array}$

Figure 1. Driver DC Test Load

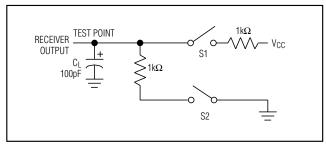


Figure 2. Receiver Enable/Disable Timing Test Load

Test Circuits and Timing Diagrams

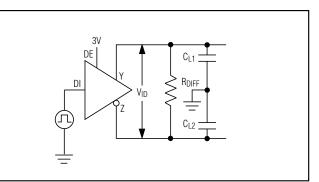


Figure 3. Driver Timing Test Circuit

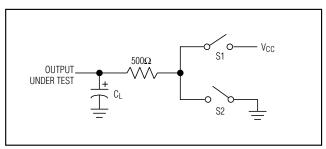


Figure 4. Driver Enable/Disable Timing Test Load

MAX3158

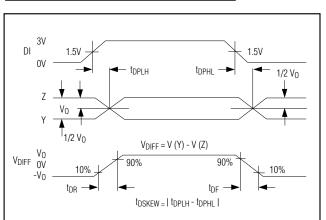


Figure 5. Driver Propagation Delays

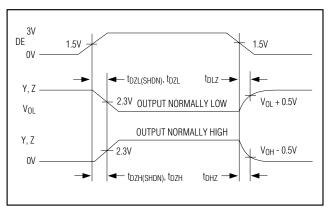


Figure 6. Driver Enable and Disable Times

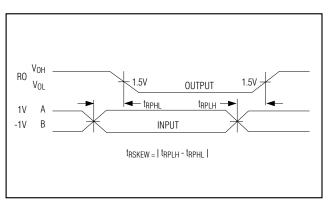


Figure 7. Receiver Propagation Delays

Test Circuits and Timing Diagrams (continued)

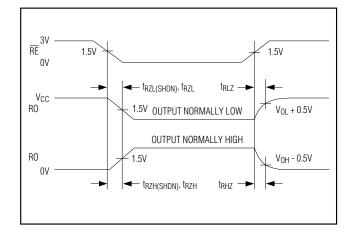


Figure 8. Receiver Enable and Disable Times

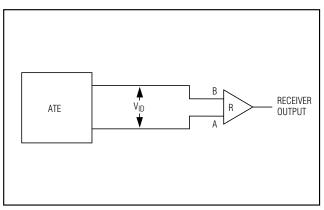


Figure 9. Receiver Propagation Delay Test Circuit

__Pin Description

Р	IN		
SSOP	PDIP	NAME	FUNCTION
1	2	RO	Receiver Output. When \overline{RE} is low and A - B > -50mV, RO will be high; if A - B \leq -200mV, RO will be low.
2	3	RE	Receiver Output Enable. Drive RE low to enable RO.
3	4	DE	Driver Output Enable. Drive DE high to enable driver outputs.
4	5	DI	Driver Input. With DE high, a low on DI forces the noninverting output low and the inverting output high; with DI high, outputs reverse.
5	6	C1LO	Connect to the negative terminal of C1 (0.047µF, 100V).
6, 7, 12, 22, 23	1, 12, 14, 15, 28	N.C.	No Connection. Not internally connected.
8	7	C1HI	Connect to the positive terminal of C1 (0.047µF, 100V).
9	8	ISOVCC	Internally generated power-supply voltage, referenced to the cable ground (ISOCOM). Connect a 1μ F capacitor to ISOCOM.
10	9	H/F	Half/Full-Duplex Selector Pin. Leave open, or connect to ISOCOM to select Full Duplex, or connect to ISOVCC to select Half Duplex.
11	10	ZR	Connect to ISOCOM for a 1.5k Ω input impedance on Z. Leave open for a 12k Ω input impedance on Z.
13	11	YR	Connect to ISOCOM for a 1.5k Ω input impedance on Y. Leave open for a 12k Ω input impedance on Y.
14	13	ISOCOM	Cable Ground
15	16	Z	Inverting Driver Output (and Inverting Receiver Input in Half-Duplex Mode)
16	17	Y	Noninverting Driver Output (and Noninverting Receiver Input in Half-Duplex Mode)
17	18	В	Inverting Receiver Input in Full-Duplex Mode
18	19	А	Noninverting Receiver Input in Full-Duplex Mode
19	20	BR	Connect to ISOCOM for a 1.5k Ω input impedance on B. Leave open for a 12k Ω input impedance on B.
20	21	AR	Connect to ISOCOM for a 1.5k Ω input impedance on A. Leave open for a 12k Ω input impedance on A.
21	22	C2HI	Connect to the positive terminal of C2 (0.047µF, 100V).
24	23	C2LO	Connect to the negative terminal of C2 (0.047µF, 100V).
25	24	GND	Ground
26	25	TXP	Transmitter Phase. Leave open, or connect to GND for normal transmitter polarity, or connect to V_{CC} to invert the transmitter polarity.
27	26	RXP	Receiver Phase. Leave open, or connect to GND for normal receiver polarity, or connect to V_{CC} to invert the receiver polarity.
28	27	Vcc	+4.75V to +5.25V Positive Supply. Connect a 0.1µF capacitor to GND.

Detailed Description

The MAX3158 is a high CMRR RS-485/RS-422 datacommunications interface providing \pm 60V isolation in a hybrid microcircuit. A single \pm 5V supply on the logic side powers both sides of the interface, with external 100V capacitors transferring power from the logic side to the isolated side (see *Block Diagram*). The MAX3158 typically draws 25mA of supply current when unloaded or when fully loaded with the driver disabled. Supply current drops to 25µA when the device is placed in shutdown mode (see *Low-Power Shutdown Mode* section). The MAX3158 transceiver for RS-485/RS-422 communication contains one driver and one receiver.

This device features fail-safe circuitry, which guarantees a logic-high receiver output when the receiver inputs are open or shorted, or when they are connected to a terminated transmission line with all drivers disabled (see *Fail-Safe* section). The MAX3158 is selectable between half- and full-duplex communication by connecting a selector pin to ISOVCC or ISOCOM, respectively. Drivers are output short-circuit current limited. Thermal shutdown circuitry protects drivers against excessive power dissipation. When activated, the thermal shutdown circuitry places the driver outputs into a high-impedance state. The device also features independently programmable receiver and transmitter output phase through separate pins. The MAX3158 is a low-cost replacement for opto-isolated transceivers.

Fail-Safe The MAX3158 **guarantees** a logic-high receiver output when the receiver inputs are shorted or open, or when they are connected to a terminated transmission line with all drivers disabled. The receiver threshold is fixed between -50mV and -200mV. If the differential receiver input voltage (A - B) is greater than or equal to -50mV, RO is logic high. If A - B is less than or equal to -200mV, RO is logic low. In the case of a terminated bus with all transmitters disabled, the receiver's differential input voltage is pulled to 0V by the termination. With the receiver thresholds of the MAX3158, this results in a logic high with a 50mV minimum noise margin. Unlike competitor's fail-safe devices, the -50mV to -200mV threshold complies with the ±200mV EIA/TIA-485 standard.

MAX3158 Programming

The MAX3158 has several programmable operating modes. Occasionally, twisted-pair lines are reversed. The MAX3158 has two pins that invert the phase of the driver and the receiver to correct for this problem. For normal operation, drive TXP and RXP low, connect them to ground, or leave them unconnected (internal pulldown). To invert the driver phase, drive TXP high or

connect it to V_{CC}. To invert the receiver phase, drive RXP high or connect it to V_{CC}. Note that the receiver threshold is positive when RXP is high.

The MAX3158 can operate in full- or half-duplex mode. Connect H/F to ISOCOM for full-duplex mode, and connect it to ISOVCC for half-duplex operation. In halfduplex mode, the receiver inputs are switched to the driver outputs, connecting outputs Y and Z to inputs A and B, respectively. In half-duplex mode, the internal full-duplex receiver input resistors are still connected to pins A and B.

Low-Power Shutdown Mode

The low-power shutdown mode is initiated by bringing both RE high and DE low. In shutdown, this device typically draws only 25µA of supply current, and no power is transferred across the isolation capacitors in this mode. RE and DE may be driven simultaneously; the parts are guaranteed not to enter shutdown if \overline{RE} is high and DE is low for less than 50ns. If the inputs are in this state for at least 700ns, the parts are guaranteed to enter shutdown. Enable times t_ZH and t_ZL in the Switching Characteristics table assume the part was not in a low-power shutdown state. Enable times t_ZH(SHDN) and t_ZL(SHDN) assume the parts were shut down. It takes drivers and receivers longer to become enabled from low-power shutdown mode t_ZH(SHDN), t_ZL(SHDN) than from driver/receiver-disable mode $(t_ZH, t_ZL).$

Driver Output Protection

Two mechanisms prevent excessive output current and power dissipation caused by faults or by bus contention. The first, a foldback current limit on the output stage, provides immediate protection against short circuits over the whole common-mode voltage range (see *Typical Operating Characteristics*). The second, a thermal shutdown circuit, forces the driver outputs into a high-impedance state if the die temperature becomes excessive—typically around +150°C.

Applications Information

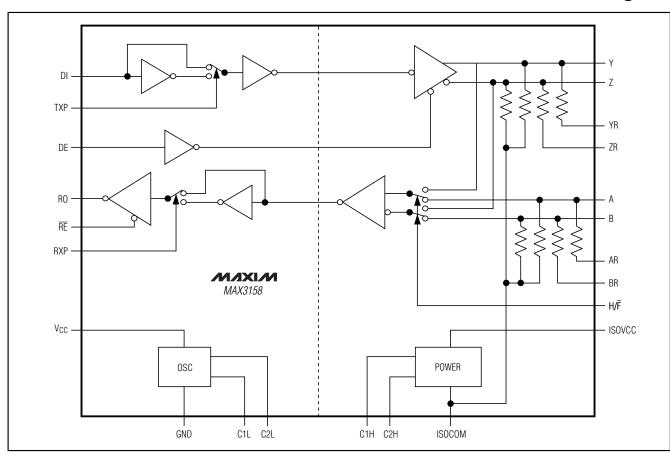
Capacitor and Grounding Resistor Selection

The value for the charge-pump capacitors C1 and C2 should be between 47nF and 100nF. Smaller values will result in insufficient supply voltage on the isolated side. Larger values are allowed but will not result in better charge-pump capacity. The values for C1 and C2, as well as that of C4, determine the maximum frequency and amplitude of the voltage difference (under operating conditions) between the local and isolated ground. Besides the capacitor values, this maximum frequency



_Block Diagram

MAX3158



and amplitude are also determined by the resistance between the remote ground and the ISOCOM pin. The receiver input resistors will cause the isolated common voltage to go to the mean voltage of the receiver inputs, which will be a direct function of the remote ground potential. The receiver input resistance and the capacitors C1, C2, and C4 set up a time constant that limits how fast the ISOCOM pin can follow variations in the remote ground voltage. Connecting YR and ZR in halfduplex operation, or AR and BR in full-duplex operation, to ISOCOM results in a relatively low input impedance of the MAX3158 receiver inputs ($2k\Omega$). This allows for a 60Hz sine wave with a 60V maximum amplitude (see Typical Operating Characteristics). If YR, ZR, AR, and BR are left open, the receiver input impedance is $12k\Omega$ allowing up to 32 transceivers on the bus. To guarantee the same low time constant under those conditions, use a shielded cable with a $1k\Omega$ resistor connected between

the shield and ISOCOM. Using a lower value for this resistor is not recommended because this could trigger a holding current in the internal ESD protection device if the \pm 75V isolation limit is exceeded.

A single point hard-ground connection for the shield is recommended.

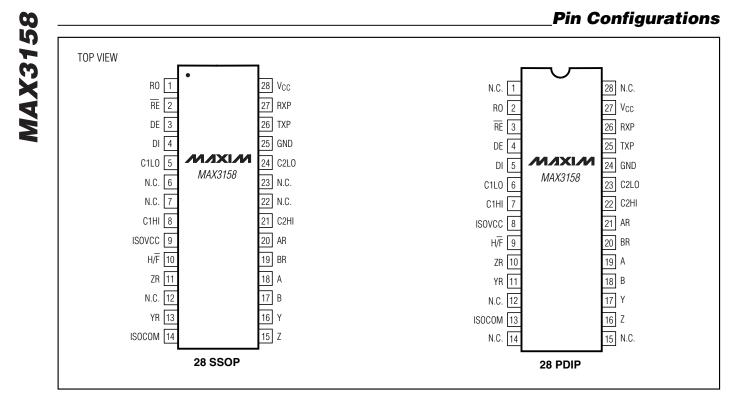
Communication Between Two MAX3158s

If two MAX3158 devices are used to communicate with each other, one of the devices must have ISOCOM connected to local ground. Failure to do so will result in floating ISOCOM pins, with both devices trying to adapt to the isolated ground of the other.

Chip Information

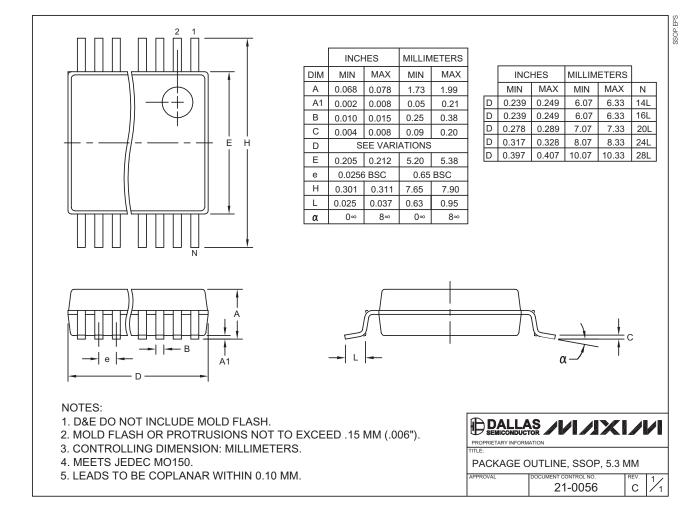
TRANSISTOR COUNT: 1309





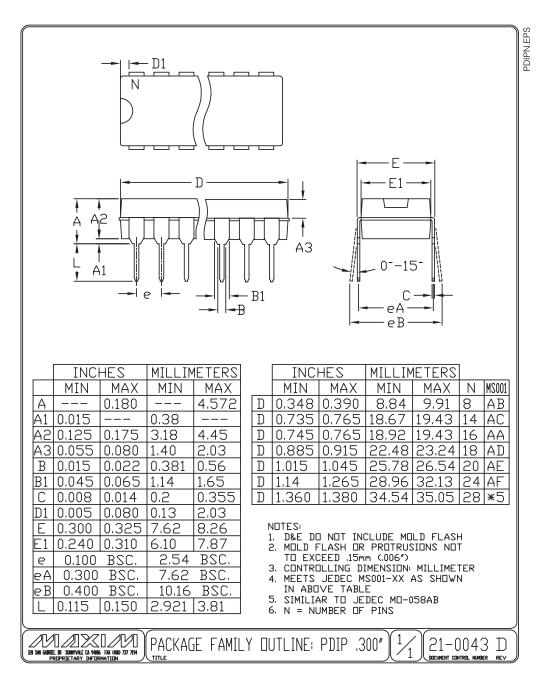
_Package Information

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