

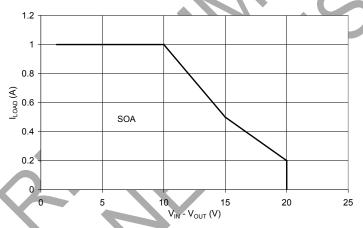
# **Pin Descriptions**

Pin Name	I/O	Pin Number	Function
Adj (GND)	I	1	A resistor divider from this pin to the $V_{\text{OUT}}$ pin and ground sets the output voltage (Ground only for Fixed-Mode).
V <sub>out</sub>	0	2	The output of the regulator. A minimum of $4.7\mu\text{F}$ capacitor $(0.05\Omega \leq \text{ESR} \leq 0.5\Omega)$ must be connected from this pin to ground to insure stability. For improved ac load response a larger output capacitor is recommended.
V <sub>IN</sub>	I	3	The input pin of regulator. Typically a large storage capacitor $(0.05\Omega \le \text{ESR} \le 0.5\Omega)$ is connected from this pin to ground to ensure that the input voltage does not sag below the minimum dropout voltage during the load transient response. This pin must always be 1.3V higher than $V_{\text{OUT}}$ in order for the device to regulate properly.

### Absolute Maximum Ratings (@TA = +25°C, unless otherwise specified.)

Symbol	Parameter	Rating	Unit
V <sub>IN</sub>	Input Supply Voltage (Relative to Ground)	-0.03 to +18	V
$T_J$	Junction Temperature	+150	°C
	Power Dissipation	See SOA Curve	
T <sub>ST</sub>	Storage Temperature	-65 to +150	°C

Unless otherwise stated voltages specified are relative to the ANODE pin.



Safe Operation Area (SOA) Curve

ESD Susceptibility			
Symbol	Parameter	Rating	Unit
НВМ	Human Body Model	4	kV
MM	Machine Model	400	V

Stresses greater than the 'Absolute Maximum Ratings' specified above, may cause permanent damage to the device. These are stress ratings only; functional operation of the device at these or any other conditions exceeding those indicated in this specification is not implied. Device reliability may be affected by exposure to absolute maximum rating conditions for extended periods of time.

Semiconductor devices are ESD sensitive and may be damaged by exposure to ESD events. Suitable ESD precautions should be taken when handling and transporting these devices



# Recommended Operating Conditions (@TA = +25°C, unless otherwise specified.)

Symbol	Parameter	Min	Max	Unit
$V_{IN}$	Input voltage	2.7 (Note 8)	18	V
Io	Output current	0.01	1	Α
TJ	Operating Junction Temperature Range (Note 5)	-40	+125	°C

### **Package Thermal Data**

Thermal Resistance	Package	Rating	Unit
Junction-to-Ambient, $\theta_{JA}$	SOT223 (Note 6) TO252 (Note 7)	107 73	°C/W
Junction-to-Case, $\theta_{JC}$	SOT223 (Note 6) TO252 (Note 7)	16 12	°C/W

Notes:

- 5. ZLDO1117 contains an internal thermal limiting circuit that is designed to protect the regulator in the event that the maximum junction temperature exceeded. When activated, typically at 150°C, the regulator Output switches off and then back on as the die cools.
  6. Test condition for SOT223: T<sub>A</sub> = +27°C, no air flow, device mounted on 2"X2" polyimide PCB, 2 oz copper, 5.6mmX5.6mm pad.
  7. Test condition for TO252: T<sub>A</sub> = +27°C, no air flow, device mounted on 2"X2" polyimide PCB, 1 oz copper, 2cmX2cm pad.
  8. Ensures correct operation without entering dropout. Device will continue to operate below this minimum input voltage under dropout conditions.

### **Electrical Characteristics** (@T<sub>A</sub> = +25°C, unless otherwise specified.)

Parameter		Conditions		Min	Тур	Max	Unit	
		$(V_{IN}-V_{OUT}) = 2V, I_O = 10mA$	25	1.238	1.250	1.263		
Reference Voltage	ZLDO1117-ADJ	$V_{OUT}$ +1.4V < $V_{IN}$ < 10V, 10mA < $I_{O}$ < 1A	FT	1.225		1.275	V	
	ZLDO1117-1.2	I <sub>O</sub> = 10mA, V <sub>IN</sub> = 3.2V	25	1.188	1.200	1.212	V	
	ZLDO1117-1.2	10mA< I <sub>0</sub> < 1A, 2.7V <v<sub>IN&lt; 12V</v<sub>	FT	1.176		1.224		
	ZLDO1117-1.5	$I_0 = 10 \text{mA}, V_{IN} = 3.5 \text{V}$	25	1.485	1.500	1.515	V	
	ZLDO1117-1.5	0 < I <sub>0</sub> < 1A, 2.9V <v<sub>IN&lt; 12V</v<sub>	FT	1.470		1.530	V	
	ZLDO1117-1.8	$I_{O} = 10 \text{mA}, V_{IN} = 3.8 \text{V}$	25	1.782	1.800	1.818	V	
Output Valtage	ZLDO1117-1.0	$0 < I_{O} < 1A$ , $3.2V < V_{IN} < 12V$	FT	1.764		1.836		
Output Voltage	ZLDO1117-2.5	I <sub>O</sub> = 10mA, V <sub>IN</sub> = 4.5V	25	2.475	2.500	2.525	V	
	ZLDO1117-2.5	0 < I <sub>0</sub> < 1A, 3.9V <v<sub>IN&lt; 12V</v<sub>	FT	2.450		2.550	V	
	ZLDO1117-3.3	$I_{\rm O}$ = 10mA, $V_{\rm IN}$ = 5.3V	25	3.267	3.300	3.333	V	
	ZLDO1117-3.3	0 < I <sub>0</sub> < 1A, 4.7V <v<sub>IN &lt; 12V</v<sub>	FT	3.235		3.365		
	ZLDO1117-5.0	$I_{O}$ = 10mA, $V_{IN}$ = 7V	25	4.95	5.000	5.05	V	
	ZLDO1117-5.0	$0 < I_0 < 1A, 6.4V < V_{IN} < 12V$	FT	4.900		5.100		
Line Degulation	ZLDO1117-ADJ	$I_O = 10mA$ ,	25			0.1	%	
	ZLDO1117-1.2	V <sub>OUT</sub> +1.5V <v<sub>IN&lt;12V</v<sub>	FT			0.2	70	
Line Regulation	ZLDO1117-xx	$I_O = 0mA$ ,	25			0.1	%	
	ZLDOTTI7-XX	V <sub>OUT</sub> +1.5V <v<sub>IN&lt;12V</v<sub>	FT			0.2	/0	

- 9. See thermal regulation specifications for changes in output voltage due to heating effects. Line and load regulation are measured at a constant junction temperature by low duty cycle pulse testing. Load regulation is measured at the output lead = 1/18" from the package.
- 10. Line and load regulation are guaranteed up to the maximum power dissipation of 15W. Power dissipation is determined by the difference between input and output differential and the output current. Guaranteed maximum power dissipation will not be available over the full input/output range.



# Electrical Characteristics (cont.) (@TA = +25°C, unless otherwise specified.)

Parameter		Conditions	T <sub>A</sub>	Min	Тур	Max	Unit
	71 DO4447 AD4	V <sub>IN</sub> =3.3V,V <sub>ADJ</sub> =0, 10mA <i<sub>O&lt;1A,</i<sub>	25			0.2	%
	ZLDO1117-ADJ	(Notes 9, 10)	FT			0.4	70
	ZLDO1117-1.2	$V_{IN}$ =2.7V, 10mA < $I_{O}$ < 1A,	25			0.2	%
	ZLDO1117-1.2	(Notes 9, 10)	FT			0.4	
	ZLDO1117-1.5	$V_{IN} = 3V, 0 < I_O < 1A,$	25			3	mV
	ZEDO1117-1.5	(Notes 9, 10)	FT			6	1117
Load Regulation	ZLDO1117-1.8	$V_{IN} = 3.3V, 0 < I_O < 1A,$	25			4	mV
Loud Regulation	2601117-1.0	(Notes 9, 10)	FT			8	1110
	ZLDO1117-2.5	$V_{IN} = 4V, 0 < I_{O} < 1A,$	25			5	mV
		(Notes 9, 10)	FT			10	
	ZLDO1117-3.3	$V_{IN} = 4.8V, 0 < I_{O} < 1A,$	25			6.6	mV
		(Notes 9, 10)	FT			13	
	ZLDO1117-5.0	$V_{IN} = 6.5V, 0 < I_0 < 1A,$	25			10	mV
	22001117 0.0	(Notes 9, 10)	FT			20	
Dropout Voltage	ZLDO1117- ADJ/1.2/1.5/1.8/2.5/		25		1.11	1.2	
(V <sub>IN</sub> -V <sub>OUT</sub> )		$I_O = 1A$ , $\Delta V_{OUT} = 1\% V_{OUT}$	0 ~ 125			1.3	V
(*IN *001)	3.3/5.0		FT			1.35	
	ZLDO1117-	0.4 3.4 3.4	25				
Current Limit	ADJ/1.2/1.5/1.8/2.5/ 3.3/5.0	$(V_{IN}-V_{OUT}) = 5V$	FT	1. 1			Α
Minimum Load Current (Note 8)	ZLDO1117-ADJ ZLDO1117-1.2	V <sub>IN</sub> = <18V	FT		2	5	mA
Quiescent current	ZLDO1117-xx	$V_{IN}$ < 18V, $I_{O}$ = 0mA	FT		4	10	mA
GND current	ZLDO1117-ADJ ZLDO1117-1.2	V <sub>IN</sub> = 7V	FT		35	120	μA
Thermal Regulation	30ms pulse		25			0.1	%/W
Ripple Rejection	f = 120Hz, C <sub>OUT</sub> = 25μF Tantalum, I <sub>OUT</sub> = 100mA, ZLDO1117-XXX V <sub>IN</sub> = V <sub>OUT</sub> +3V		25	60	80		dB
Temperature Stability	I <sub>O</sub> = 10mA			•	0.5	_	%

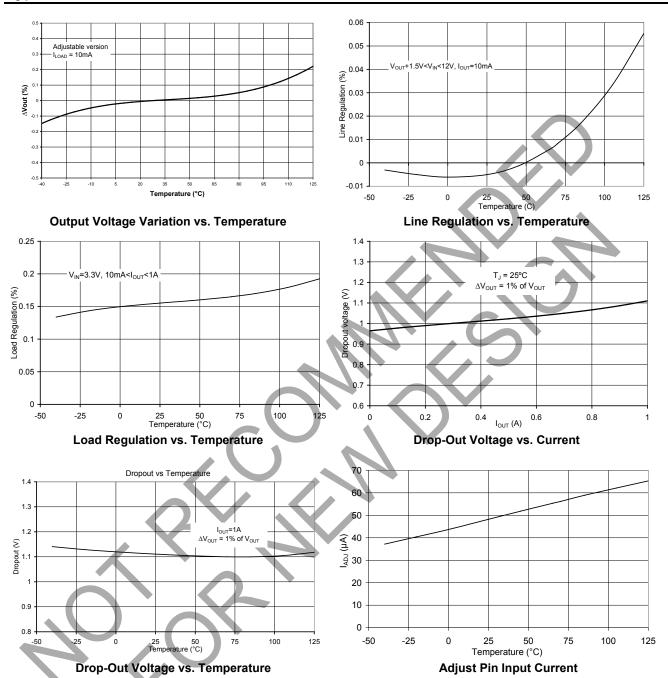
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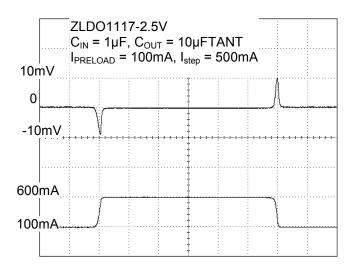


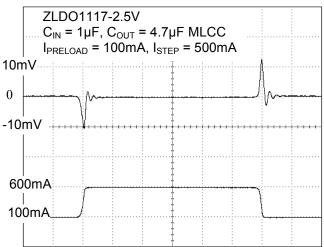
# **Typical Characteristics**





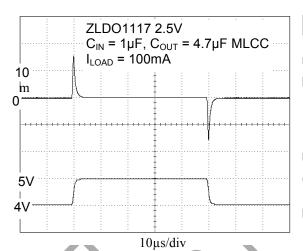
# **Typical Characteristics**



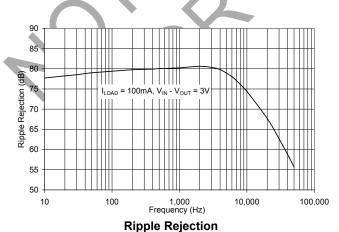


Transient Load Regulation with 10µF Tantalum Capacitor

Transient Load Regulation with 4.7µF MLCC Capacitor



Transient Line Regulation with 4.7µF MLCC Capacitor





### **Application Information**

The ZLDO1117 family of quasi-LDO regulators is easy to use. They are protected against short circuit and thermal overloads. (see block diagram).

Thermal protection circuitry will shut down the regulator should the junction temperature exceed +150°C at the sense point. The ZLDO1117 is pin compatible with similar '1117 regulators and offers extended temperature range and improved regulation specifications.

#### Operation

The ZLDO1117 develops a 1.25V reference voltage between the output and the adjust terminal (see block diagram). By placing a resistor between these two terminals, a constant current is caused to flow through R1 and down through R2. For fixed output variants Resistors R1 and R2 are internal.

#### Stability

The ZLDO1117 requires an output capacitor as part of the device frequency compensation. As part of its improved performance over industry standard 1117 the ZLDO1117 is suitable for use with MLCC (Multi Layer Ceramic Chip) capacitors. A minimum of  $4.7\mu\text{F}$  ceramic X7R,  $4.7\mu\text{F}$  tantalum, or  $47~\mu\text{F}$  of aluminum electrolytic is required. The ESR of the output capacitor should be less than  $0.5\Omega$ . Surface mount tantalum capacitors, which have very low ESR, are available from several manufacturers. When using MLCC capacitors avoid the use of Y5V dielectrics.

#### **Load Regulation**

For improved load regulation the ZLDO1117-ADJ should have the upper feedback resistor, R<sub>1</sub>, connected as close as possible to V<sub>OUT</sub> and the lower resistor, R2, connected as close as possible to the load GND return. This helps reduce any parasitic resistance in series with the load.

#### **Thermal Considerations**

ZLDO1117 series regulators have internal thermal limiting circuitry designed to protect the device during overload conditions. For continuous normal load conditions however, the maximum junction temperature rating of +125°C must not be exceeded.

It is important to give careful consideration to all sources of thermal resistance from junction to ambient. For the SOT223-3L and TO252-3L packages, which are designed to be surface mounted, additional heat sources mounted near the device must also be considered. Heat sinking is accomplished using the heat spreading capability of the PCB and its copper traces. The  $\theta_{JC}$  (junction to tab)of the TO252-3L and SOT223-3L are +12°C/W and +16°C/W respectively.

Thermal resistances from tab to ambient can be as low as  $+30^{\circ}$ C/W. The total thermal resistance from junction to ambient can be as low as +42 to  $+46^{\circ}$ C/W. This requires a reasonable sized PCB with at least one layer of copper to spread the heat across the board and couple it into the surrounding air. Datasheet specifications using 2 oz copper and a 5mmx5mm pad with  $T_A = +27^{\circ}$ C, no air flow yielded  $\theta_{JA}$  (junction to tab) of  $+73^{\circ}$ C/W and  $+107^{\circ}$ C/W for TO252-3L and SOT223-3L respectively.

The thermal resistance for each application will be affected by thermal interactions with other components on the board. Some experimentation will be necessary to determine the actual value.

#### Ripple Rejection

When using the ZLDO1117 adjustable device the adjust terminal can be bypassed to improve ripple rejection. When the adjust terminal is bypassed the required value of the output capacitor increases.

The device will require an output capacitor of  $22\mu\text{F}$  tantalum or  $150\mu\text{F}$  aluminum electrolytic when the adjust pin is bypassed. Normally, capacitor values on the order of  $100\mu\text{F}$  are used in the output of many regulators to ensure good load transient response with large load current changes. Output capacitance can be increased without limit and larger values of output capacitance further improve stability and transient response.

The curves for Ripple Rejection were generated using an adjustable device with the adjust pin bypassed. These curves will hold true for all values of output voltage. For proper bypassing, and ripple rejection approaching the values shown, the impedance of the adjust pin capacitor, at the ripple frequency, should be < R1. R1 is normally in the range of  $100\Omega$  to  $200\Omega$ . The size of the required adjust pin capacitor is a function of the input ripple frequency. At 120Hz, with R1 =  $100\Omega$ , the adjust pin capacitor should be  $>13\mu$ F. At 10kHz only  $0.16\mu$ F is needed.

For fixed voltage devices, and adjustable devices without an adjust pin capacitor, the output ripple will increase as the ratio of the output voltage to the reference voltage ( $V_{OUT}/V_{REF}$ ). For example, with the output voltage equal to 5V, the output ripple will be increased by the ratio of 5V/1.25V. It will increase by a factor of four. Ripple rejection will be degraded by 12dB from the value shown on the curve.



### **Typical Application Circuits**

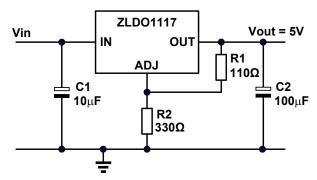


Figure 1 Basic Adjustable Regulator with 5V Output

Using

$$V_{OUT} = 1.25 \bullet \left\{ 1 + \frac{R2}{R1} \right\}$$

then the output voltage becomes:

$$V_{OUT} = 1.25 \bullet \left\{ 1 + \frac{330}{110} \right\} = 5.0V$$

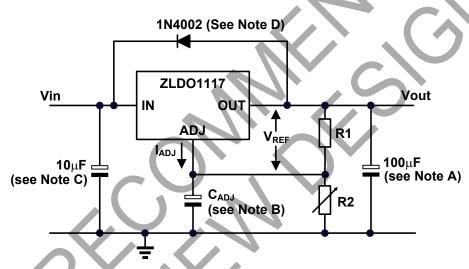


Figure 2 Adjustable Regulator with IADJ Errors

$$V_{OUT} = 1.25 \bullet \left\{ 1 + \frac{R2}{R1} \right\} + I_{ADJ} \bullet R2$$

Because I<sub>ADJ</sub> typically is 55µA, its effect is negligible in most applications.

$$V_{OUT} = 1.25 \bullet \left\{ 1 + \frac{330}{110} + 55 \bullet 10^{-6} \bullet 330 \right\} = 5.02 V \sim 0.4\%$$

- A. Output capacitor selection is critical for regulator stability. Larger C<sub>out</sub> values benefit the regulator by improving transient response and loop stability.
- B. C<sub>ADJ</sub> can be used to improve ripple rejection. If C<sub>ADJ</sub> is used, a C<sub>out</sub> that is larger in value than C<sub>ADJ</sub> must be used.
- C. C<sub>in</sub> is recommended if ZLDO1117 is not located near the power supply filter.
- D. An external diode is recommended to protect the regulator if the input instantaneously is shorted to GND.
- E. This device is designed to be stable with tantalum and MLCC capacitors with an ESR less than 0.47Ω.



# **Other Application Circuits**

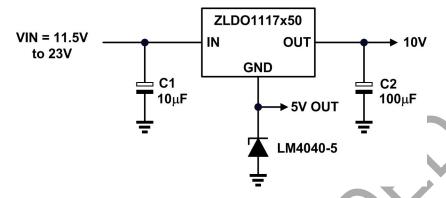


Figure 3 ZLDO1117 with Extended Output Voltage

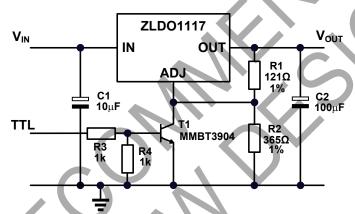
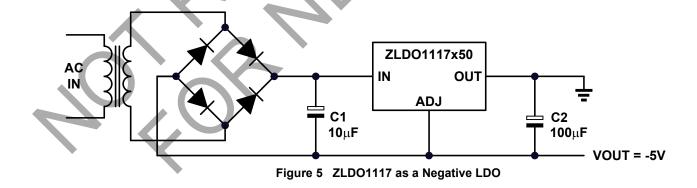


Figure 4 ZLDO1117 with Disable Function





# **Ordering Information**

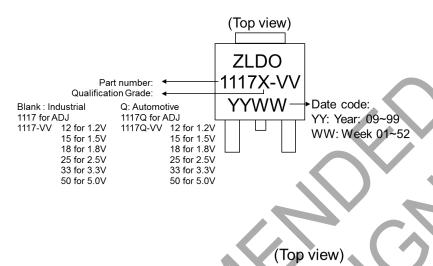
Part Number	Output Voltage	Packaging (Note 11)	Status	Reel Quantity	Tape width	Reel size
ZLDO1117KTC	Adimatable	TO252	Active	2500	16 mm	13"
ZLDO1117GTA	Adjustable	SOT223	Active	1000	12 mm	7"
ZLDO1117K12TC	1.2V	TO252	Active	2500	16 mm	13"
ZLDO1117G12TA	1.20	SOT223	Active	1000	12 mm	7"
ZLDO1117K15TC	4.5)/	TO252	Active	2500	16 mm	13"
ZLDO1117G15TA	1.5V	SOT223	Active	1000	12 mm	7"
ZLDO1117K18TC	1.8V	TO252	Active	2500	16 mm	13"
ZLDO1117G18TA	1.87	SOT223	Active	1000	12 mm	7"
ZLDO1117K25TC	2.5V	TO252	Active	2500	16 mm	13"
ZLDO1117G25TA	2.50	SOT223	Active	1000	12 mm	7"
ZLDO1117K33TC	3.3V	TO252	Active	2500	16 mm	13"
ZLDO1117G33TA	3.30	SOT223	Active	1000	12 mm	7"
ZLDO1117K50TC	5.0V	TO252	Active	2500	16 mm	13"
ZLDO1117G50TA	5.00	SOT223	Active	1000	12 mm	7"
ZLDO1117QKTC	Adimatable	TO252	Active	2500	16 mm	13"
ZLDO1117QGTA	Adjustable	SOT223	Active	1000	12 mm	7"
ZLDO1117QK12TC	1.2V	TO252	Active	2500	16 mm	13"
ZLDO1117QG12TA	1.20	SOT223	Active	1000	12 mm	7"
ZLDO1117QK15TC	1.5V	TO252	Active	2500	16 mm	13"
ZLDO1117QG15TA	1.5V	SOT223	Active	1000	12 mm	7"
ZLDO1117QK18TC	1.8V	TO252	Active	2500	16 mm	13"
ZLDO1117QG18TA	1.87	SOT223	Active	1000	12 mm	7"
ZLDO1117QK25TC	2.51/	TO252	Active	2500	16 mm	13"
ZLDO1117QG25TA	2.5V	SOT223	Active	1000	12 mm	7"
ZLDO1117QK33TC	2.01/	TO252	Active	2500	16 mm	13"
ZLDO1117QG33TA	3.3V	SOT223	Active	1000	12 mm	7"
ZLDO1117QK50TC	F 0)/	TO252	Active	2500	16 mm	13"
ZLDO1117QG50TA	5.0V	SOT223	Active	1000	12 mm	7"

Note: 11. Pad layout as shown on Diodes Inc. suggested pad layout document AP02001, which can be found on our website at http://www.diodes.com/datasheets/ap02001.pdf.

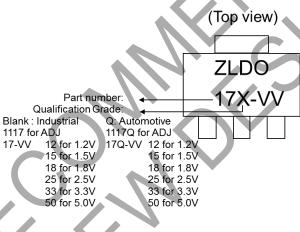


# **Marking Information**

TO252



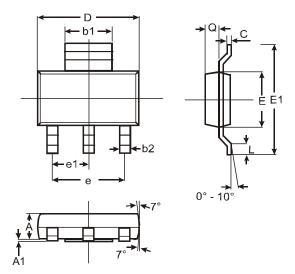
**SOT223** 





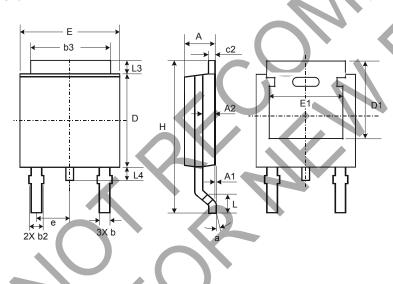
# Package Outline Dimensions (All dimensions in mm.)

### **SOT223**



	SOT223				
Dim	Min	Max	Тур		
Α	1.55	1.65	1.60		
A1	0.010	0.15	0.05		
b1	2.90	3.10	3.00		
b2	0.60	0.80	0.70		
С	0.20	0.30	0.25		
D	6.45	6.55	6.50		
Е	3.45	3.55	3.50		
E1	6.90	7.10	7.00		
е	1		4.60		
e1	-		2.30		
J	0.85	1.05	0.95		
Q	0.84	0.94	0.89		
All Dimensions in mm					

### TO252

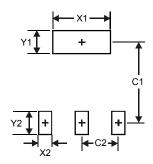


_					
	TC	252			
Dim	Min	Max	Тур		
Α	2.19	2.39	2.29		
<b>A1</b>	0.00	0.13	0.08		
<b>A2</b>	0.97	1.17	1.07		
b	0.64	0.88	0.783		
b2	0.76	1.14	0.95		
b3	5.21	5.46	5.33		
c2	0.45	0.58	0.531		
D	6.00	6.20	6.10		
D1	5.21	_	_		
е	_	_	2.286		
Е	6.45	6.70	6.58		
E1	4.32	_	_		
H	9.40	10.41	9.91		
Г	1.40	1.78	1.59		
L3	0.88	1.27	1.08		
L4	0.64	1.02	0.83		
а	0°	10°	_		
All	All Dimensions in mm				



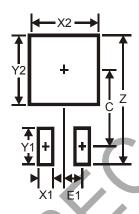
# **Suggested Pad Layout**

### SOT223



Dimensions	Value (in mm)
X1	3.3
X2	1.2
Y1	1.6
Y2	1.6
C1	6.4
C2	2.3

TO252



Dimensions	Value (in mm)
Z	11.6
X1	1.5
X2	7.0
Y1	2.5
Y2	7.0
С	6.9
F1	2.3



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