AP331A

SINGLE DIFFERENTIAL COMPARATOR

Pin Assignments

(1) SOT25

(Top View)

```
IN-  [1]         [5]  Vcc

GND  [2]

IN+  [3]         [4]  Output
```

(2) SOT25R

(Top View)

```
IN+  [1]         [5]  Vcc

GND  [2]

IN-  [3]         [4]  Output
```
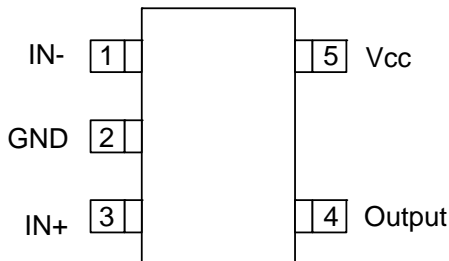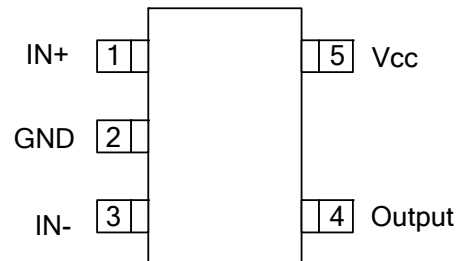
Pin Descriptions

Pin Name	Pin No.		Description
	SOT25	SOT25R	
IN-	1	3	Input-
GND	2	2	Ground
IN+	3	1	Input+
Output	4	4	Output
Vcc	5	5	Power Supply

Block Diagram
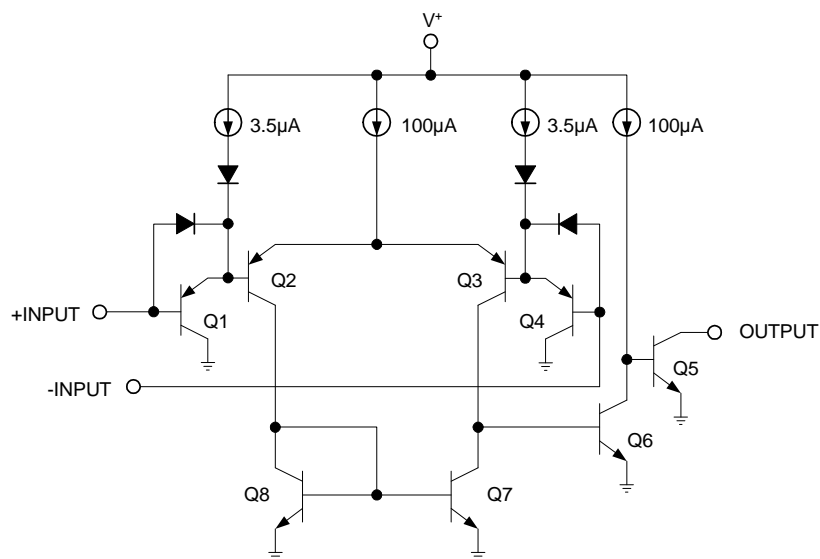
Absolute Maximum Ratings

Symbol	Parameter		Rating	Unit
V_{CC}	Supply Voltage		40	V
V_{IN}	Differential Input Voltage **(Note 10)**		40	V
V_{IN}	Input Voltage		-0.3 to +40	V
I_{CC}	Input Current (V_{IN}-0.3V) **(Note 5)**		50	mA
P_D	Power Dissipation **(Note 3)**	SOT25	740	mW
		SOT25R		
	Output Short-Circuit to Ground **(Note 4)**		Continuous	
T_J	Operating Junction Temperature Range		+150	^oC
T_{ST}	Storage Temperature Range		-65 to +150	^oC

Recommended Operating Conditions

Symbol	Parameter	Min	Max	Unit
V_{CC}	Supply Voltage (Single)	+2	+36	V
V_{CC}	Supply Voltage (Dual Supplies, V^+ VS GND)	±1	±18	V
V_{IN}	Differential Input Voltage (Single) **(Note 10)**	0	±36	V
V_{IN}	Differential Input Voltage (Dual Supplies) **(Note 10)**	-18	+18	V
T_A	Operating Ambient Temperature	0	70	^oC

Electrical Characteristics

(V_{CC} = 5V, T_A = 25°C, unless otherwise stated)

Symbol	Parameter	Conditions		Min	Typ.	Max	Unit
V_{OFFSET}	Input Offset Voltage	**(Note 11)**		-	4.0	8.0	mV
I_{BIAS}	Input Bias Current	I_{IN}(+) or I_{IN}(-) with Output In Linear Range, V_{CM} = 0V **(Note 7)**		-	150	400	nA
I_{OFFSET}	Input Offset Current	I_{IN}(+) - I_{IN}(-) V_{CM} = 0V		-	70	150	nA
	Input Common Mode Voltage Range	V^+ = 30V **(Note 8)**		0	-	V^+-1.5	V
I_{CC}	Supply Current	R_L = ∞	V^+ = 5V	-	0.4	1	mA
			V^+ = 36V	-	1	2.5	
	Voltage Gain	$R_L \geq$ 15kΩ, V^+ = 15V V_O = 1V to 11V		50	200	-	V/mV
	Large Signal Response Time	V_{IN} = TTL Logic Swing, V_{REF} = 1.4V, V_{RL} = 5V, R_L = 5.1kΩ		-	300	-	ns
	Response Time	V_{RL} = 5V, R_L = 5.1kΩ **(Note 9)**		-	1.3	-	µs
$I_{O(Sink)}$	Output Sink Current	V_{IN}(-) = 1V, V_{IN}(+) = 0, $V_O \leq$ 1.5V		6.0	16	-	mA
V_{SAT}	Saturation Voltage	V_{IN}(-) = 1V, V_{IN}(+) = 0, $I_{SINK} \leq$ 4mA		-	250	400	mV
$I_{O(Leak)}$	Output Leakage Current	V_{IN}(-) = 0, V_{IN}(+) = 1V, V_O = 5V		-	0.1	-	nA

Electrical Characteristics

(V_{CC} = 5V) (Note 6)

Symbol	Parameter	Conditions	Min	Typ.	Max	Unit
V_{OFFSET}	Input Offset Voltage	(Note 11)	-	-	9	mV
I_{OFFSET}	Input Offset Current	$I_{IN}(+)$ - $I_{IN}(-)$, V_{CM} = 0V	-	-	200	nA
I_{BIAS}	Input Bias Current	$I_{IN}(+)$ or $I_{IN}(-)$ with Output In Linear Range, V_{CM} = 0V (Note 7)	-	-	500	nA
	Input Common Mode Voltage Range	V^+ =30V (Note 8)	0	-	V^+-2.0	V
V_{SAT}	Saturation Voltage	$V_{IN}(-)$ = 1V, $V_{IN}(+)$ = 0, $I_{SINK} \leq$ 4mA	-	-	700	mV
$I_{O(Leak)}$	Output Leakage Current	$V_{IN}(-)$ = 0, $V_{IN}(+)$ = 1V, V_O = 30V	-	-	1.0	μA
	Differential Input Voltage	Keep All V_{IN}'s $\geq$ 0V (or V^-, if Used), (Note 10)	-	-	36	V
θ_{JA}	Thermal Resistance Junction-to-Ambient	SOT25 (Note 12)		173		°C/W
		SOT25R (Note 12)				
θ_{JC}	Thermal Resistance Junction-to-Case	SOT25 (Note 12)		37		°C/W
		SOT25R (Note 12)				

Notes: 3. The maximum power dissipation is a function of $T_{J(MAX)}$, θ_{JA}. The maximum allowable power dissipation at any ambient temperature is $P_D = (T_{J(MAX)} - T_A)/ \theta_{JA}$. All numbers apply for packages soldered directly onto a PC board.

4. Short circuits from the output to V^+ can cause excessive heating and eventual destruction. When considering short circuits to ground, the maximum output current is approximately 20mA independent of the magnitude of V^+.

5. This input current will only exist when the voltage at any of the input leads is driven negative. It is due to the collector-base junction of the input PNP transistors becoming forward biased and thereby acting as input diode clamps. In addition to this diode action, there is also lateral NPN parasitic transistor action on the IC chip. This transistor action can cause the output voltages of the comparators to go to the V^+ voltage level (or to ground for a large overdrive) for the time duration that an input is driven negative. This is not destructive and normal output states will re-establish when the input voltage, which is negative, again returns to a value greater than -0.3V.

6. The AP331A temperature specifications are limited to 0°C $\leq$ T_{OP} $\leq$ +70°C.

7. The direction of the input current is out of the IC due to the PNP input stage. This current is essentially constant, independent of the state of the output so no loading change exists on the reference or input lines.

8. The input common-mode voltage or either input signal voltage should not be allowed to go negative by more than 0.3V. The upper end of the common-mode voltage range is V^+-1.5V at 25°C, but either or both inputs can go to 36V without damage, independent of the magnitude of V^+.

9. The response time specified is for a 100mV input step with 5mV overdrive. For larger overdrive signals 300ns can be obtained, see typical performance characteristics section.

10. Positive excursions of input voltage may exceed the power supply level. As long as the other voltage remains within the common-mode range, the comparator will provide a proper output state. The low input voltage state must not be less than -0.3V (or 0.3V below the magnitude of the negative power supply, if used).

11. At output switch point, $V_O \simeq$ 1.4V, R_S=0Ω with V^+ from 5V to 30V; and over the full input common-mode range (0V to V^+-1.5V), at 25°C.

12. Test condition for SOT25 and sot25R: Device mounted on FR-4 substrate PC board, 2oz copper, with minimum recommended pad layout.

Application Information

The AP331A is high gain, wide bandwidth devices, like most comparators, can easily oscillate if the output lead is inadvertently allowed to capacitively couple to the inputs via stray capacitance. This shows up only during the output voltage transition intervals as the comparator change states. Power supply bypassing is not required to solve this problem. Standard PC board layout is helpful as it reduces stray input-output coupling. Reducing the input resistors to < 10kΩ reduces the feedback signal levels and finally, adding even a small amount (1.0 to 10 mV) of positive feedback (hysteresis) causes such a rapid transition that oscillations due to stray feedback are not possible. Simply socketing the IC and attaching resistors to the pins will cause input-output oscillations during the small transition intervals unless hysteresis is used. If the input signal is a pulse waveform, with relatively fast rise and fall times, hysteresis is not required. All input pins of any unused comparators should be tied to the negative supply.

The bias network of the AP331A establishes a drain current independent of the magnitude of the power supply voltage over the range of from 2.0 V_{DC} to 30 V_{DC}. It is usually unnecessary to use a bypass capacitor across the power supply line.

The differential input voltage may be larger than V+ without damaging the device (Note 10). Protection should be provided to prevent the input voltages from going negative more than -0.3 V_{DC} (at 25°C). An input clamp diode can be used as shown in the applications section.

The output of the AP331A is the uncommitted collector of a grounded-emitter NPN output transistor. Many collectors can be tied together to provide an output OR'ing function. An output pull-up resistor can be connected to any available power supply voltage within the permitted supply voltage range and there is no restriction on this voltage due to the magnitude of the voltage applied to the V^+ terminal of the AP331A package. The output can also be used as a simple SPST switch to ground (when a pull-up resistor is not used). The amount of current the output device can sink is limited by the drive available (which is independent of V^+) and the β of this device. When the maximum current limit is reached (approximately 16mA), the output transistor will come out of saturation and the output voltage will rise very rapidly. The output saturation voltage is limited by the approximately 60Ω r_{SAT} of the output transistor. The low offset voltage of the output transistor (1.0 mV) allows the output to clamp essentially to ground level for small load currents.
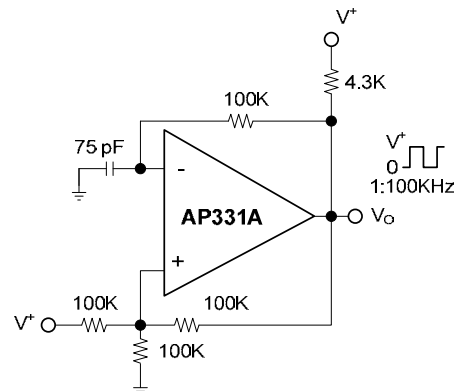
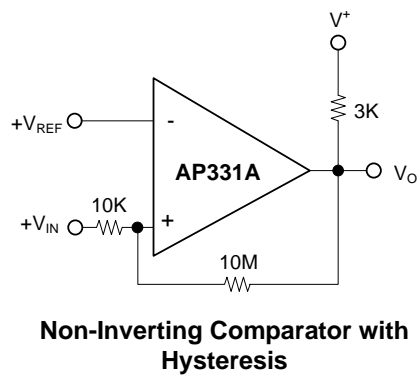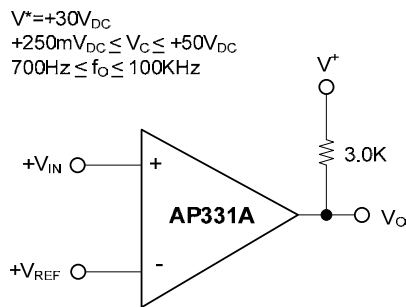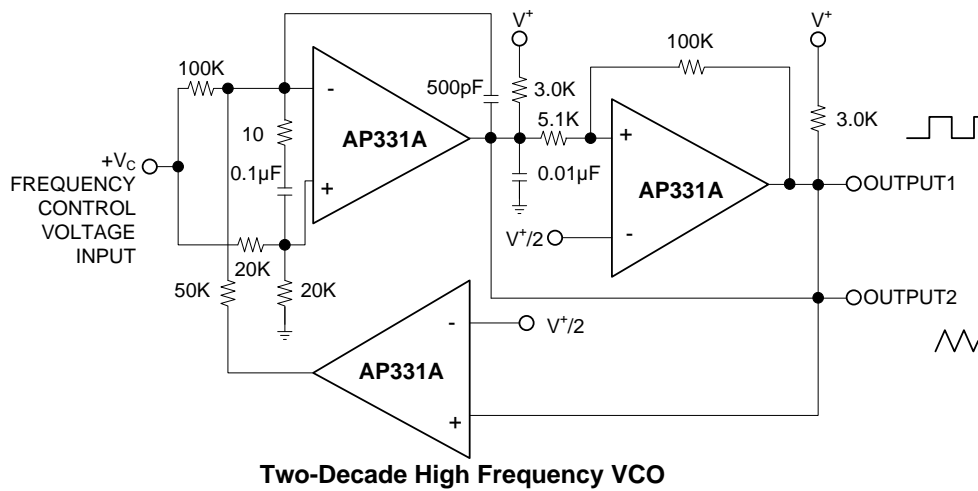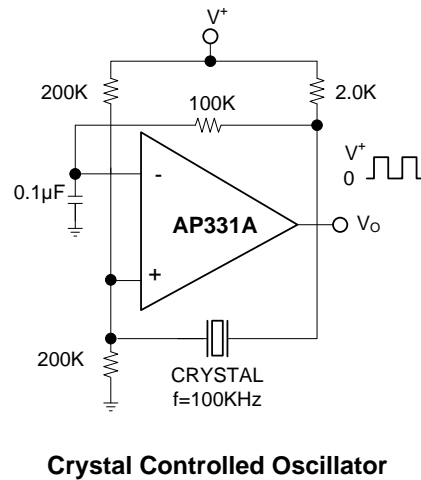# Typical Circuit   (V_{CC} = 5.0V_{DC})
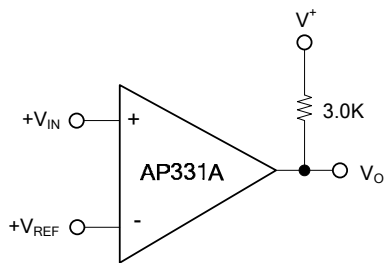
Basic Comparator
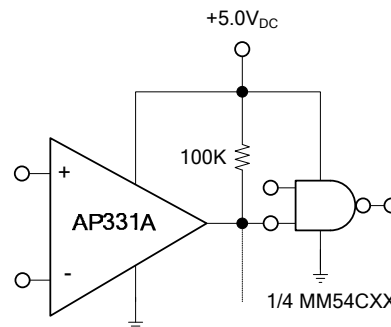
Driving CMOS
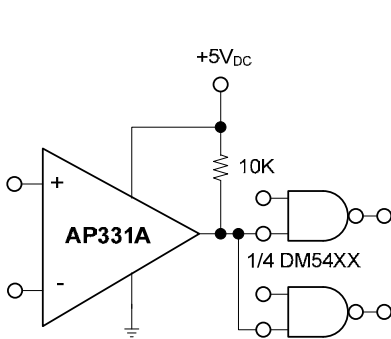
Driving TTL

Squarewave Oscillator

Typical Circuit (Continued) (V_{CC} = 5.0V_{DC})

Pulse Generator

*For large ratios of R1/R2, D1 can be omitted

Crystal Controlled Oscillator

Two-Decade High Frequency VCO

V^+=+30V_{DC}
+250m$V_{DC} \leq V_C \leq$ +50V_{DC}
700Hz $\leq f_O \leq$ 100KHz

Basic Comparator

Non-Inverting Comparator with Hysteresis
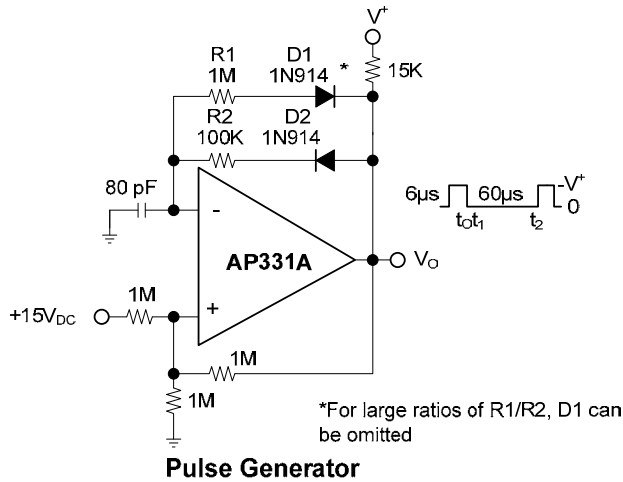
Typical Circuit (Continued) (V_{CC} = 5.0V_{DC})
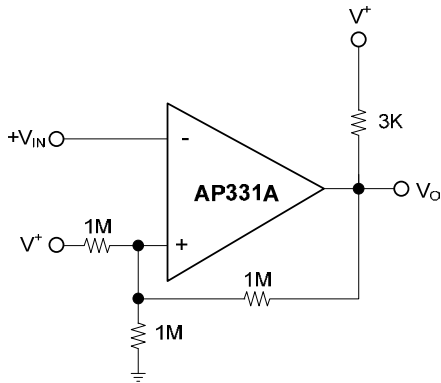
Inverting Comparator with Hysteresis
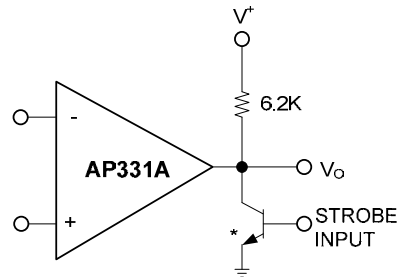
Output Strobing

* OR LOGIC GATE WITHOUT PULL-UP RESISTOR

AND Gate

Or Gate

ALL DIODES 1N914
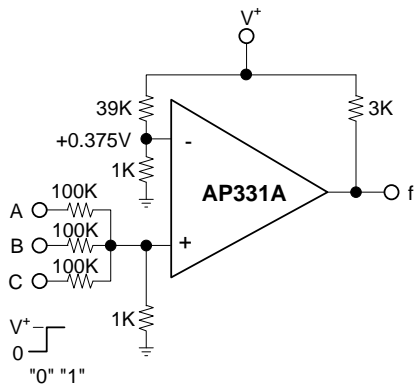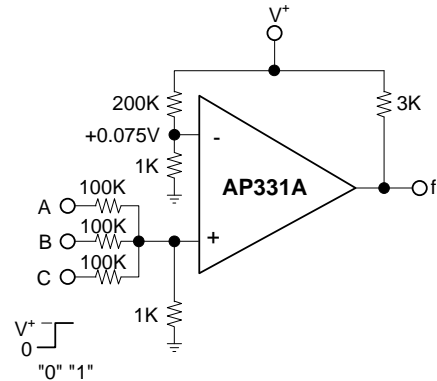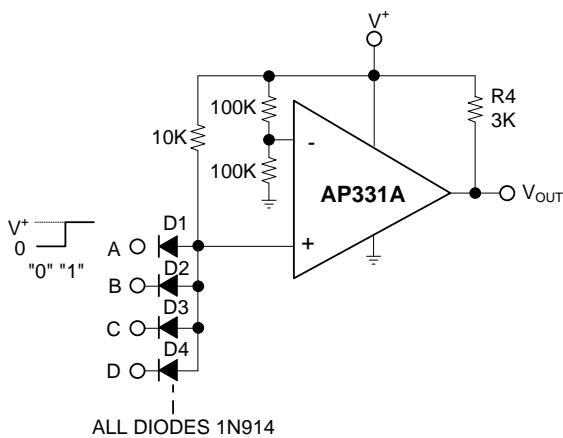
Large Fan-in AND Gate
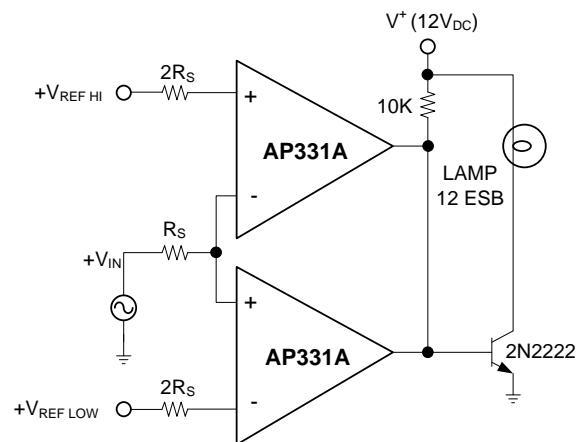
Limit Comparator

Typical Circuit   (Continued)
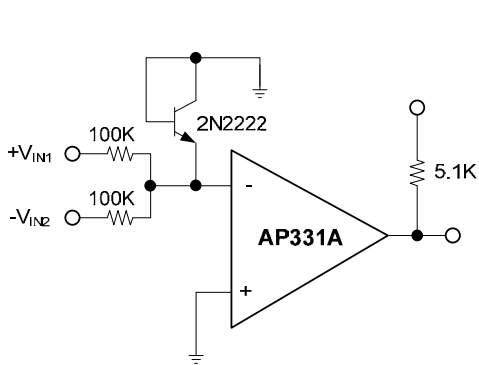
($V_{CC} = 5.0V_{DC}$)

**Comparing Input Voltages of
Opposite Polarity**
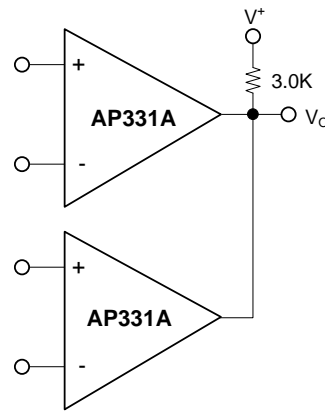
ORing the Outputs
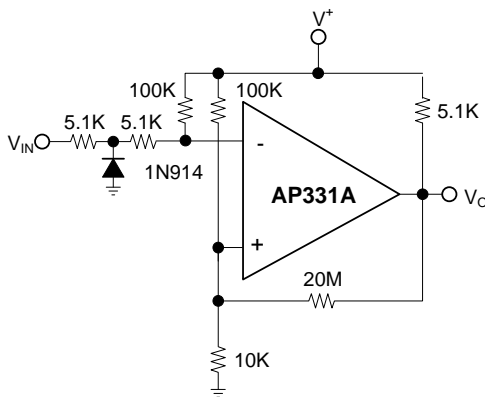
**Zero Crossing Detector
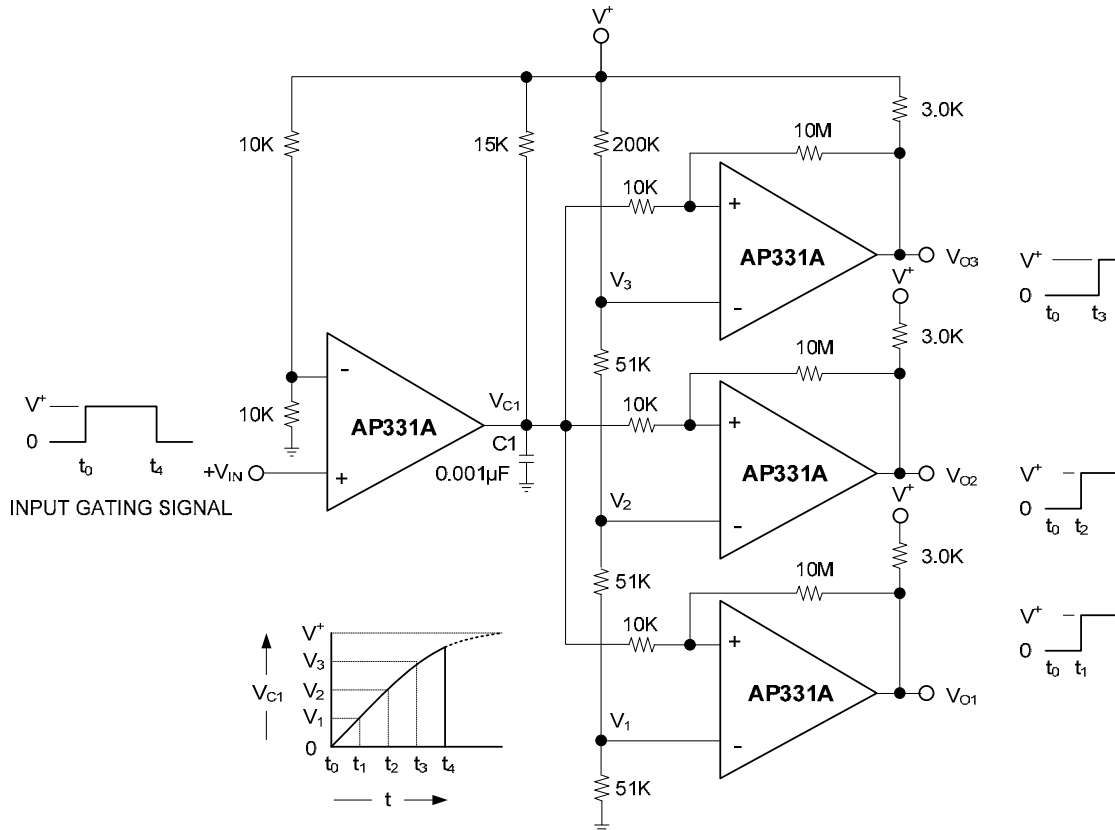(Single Power Supply)**
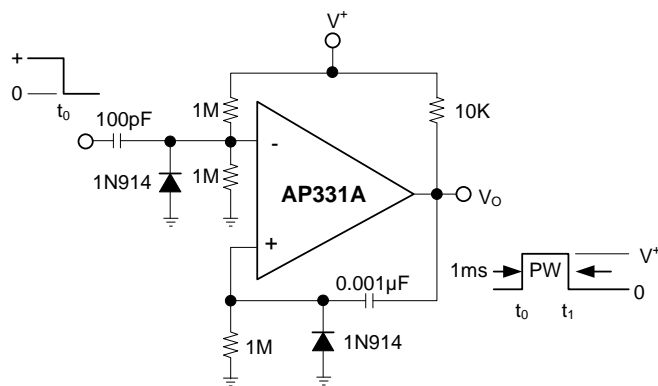
One-Shot Multivibrator
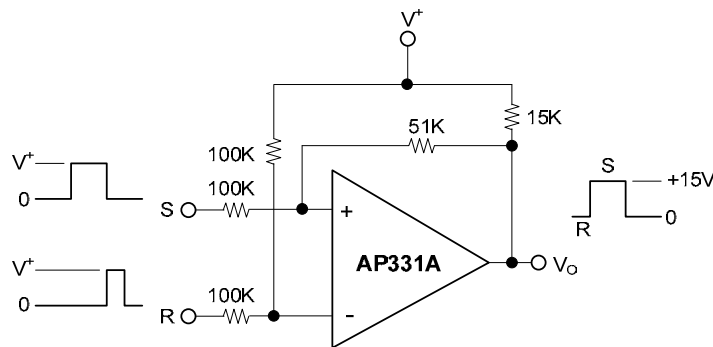
Bi-Stable Multivibrator

Typical Circuit   (Continued)

$(V_{CC} = 5.0V_{DC})$
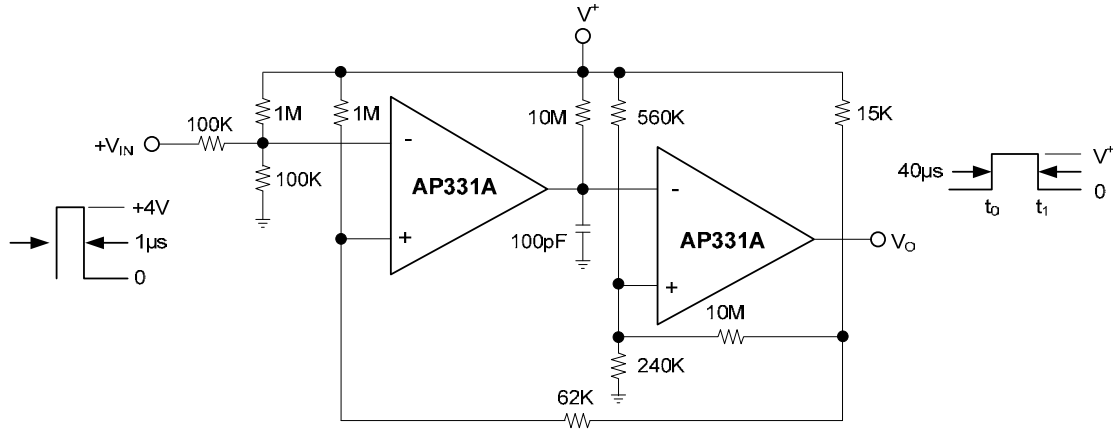
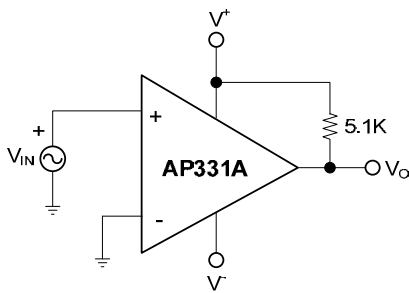**One-Shot Multivibrator with Input Lock Out**

Time Delay Generator
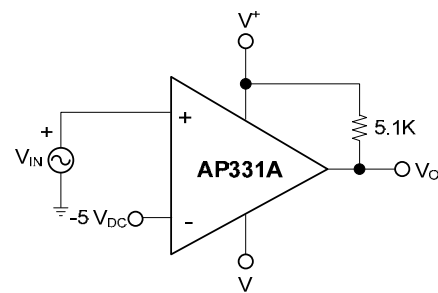
Split-Supply Applications

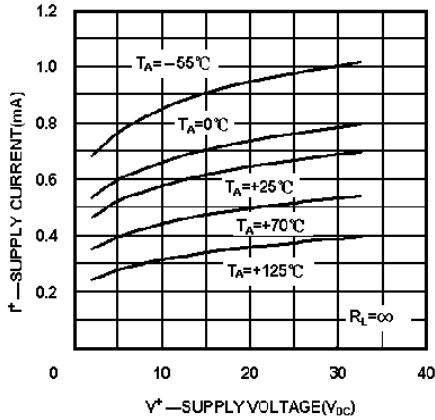$(V^+ = +15V_{DC}$ and $V^- = -15\ V_{DC})$
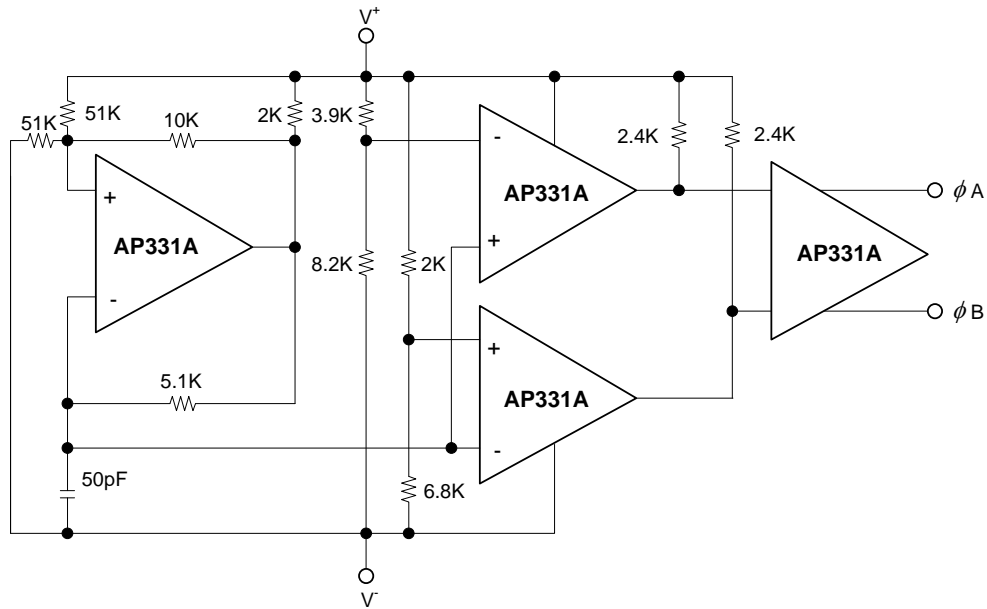
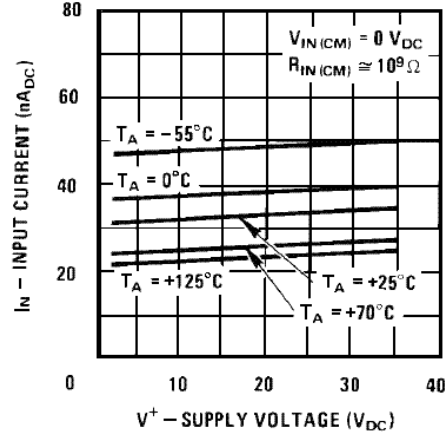
MOS Clock Driver

Zero Crossing Detector

**Comparator With a
Negative Reference**
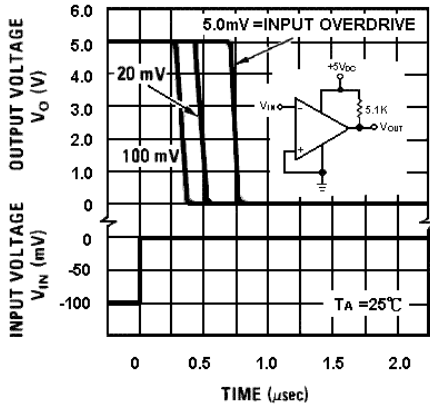
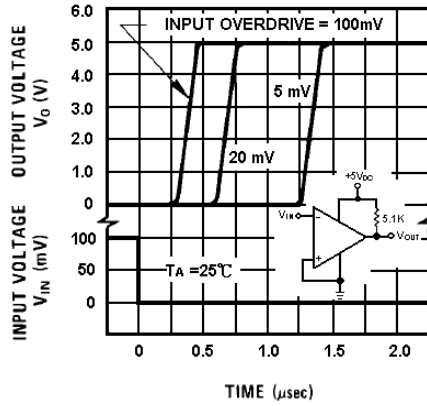## Typical Characteristics

Supply Current

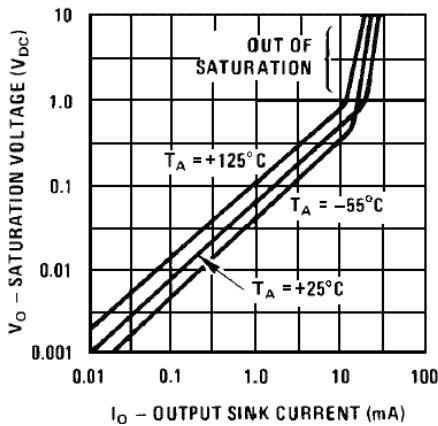### Input Current

Response Time for Various Input Overdrives—Negative Transition

Response Time for Various Input Overdrives—Positive Transition
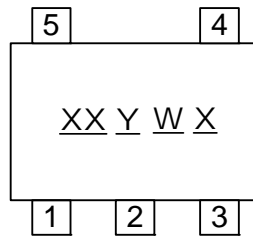

Output Saturation Voltage

Marking Information

(1) SOT25 and SOT25R

(Top View)

```
5        4
XX Y W X
1   2    3
```
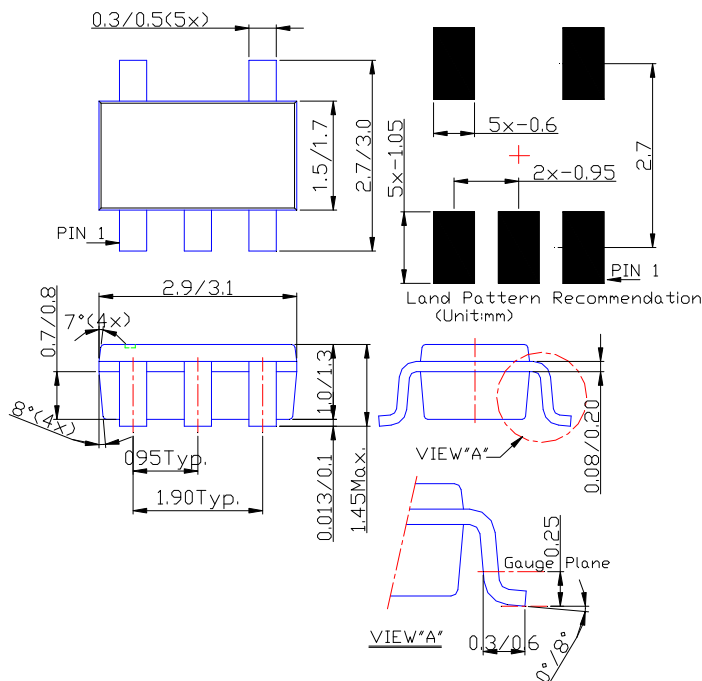
XX : Identification code
Y : Year 0~9
W : Week : A~Z : 1~26 week;
     a~z : 27~52 week; z represents
     52 and 53 week
X : A~Z : Green

Part Number	Package	Identification Code
AP331A	SOT25	R9
AP331A	SOT25R	R3

Package Information   (All Dimensions in mm)

(1) Package Type: SOT25 and SOT25R

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