MAXIMUM RATINGS ($T_A = +25^{\circ}C$, unless otherwise noted.)

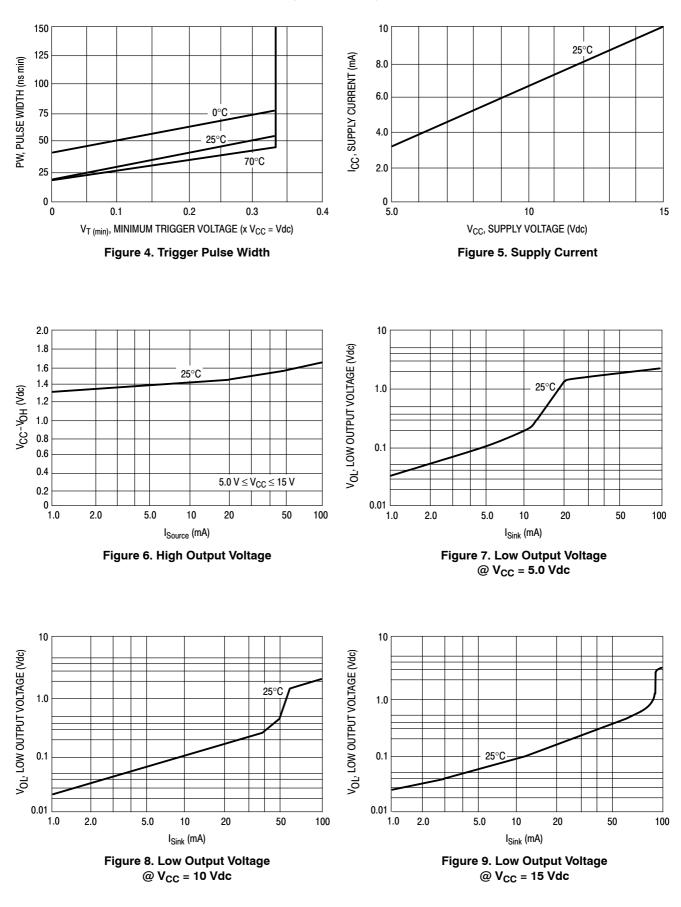
Rating	Symbol	Value	Unit
Power Supply Voltage	V _{CC}	+18	Vdc
Discharge Current (Pin 7)	ا ₇	200	mA
Power Dissipation (Package Limitation) P1 Suffix, Plastic Package Derate above $T_A = +25^{\circ}C$ D Suffix, Plastic Package Derate above $T_A = +25^{\circ}C$	P _D P _D	625 5.0 625 160	mW mW/°C mW °C/W
Operating Temperature Range (Ambient) MC1455B MC1455 NCV1455B	T _A	-40 to +85 0 to +70 -40 to +125	°C
Maximum Operating Die Junction Temperature	TJ	+150	°C
Storage Temperature Range	T _{stg}	-65 to +150	°C

Stresses exceeding Maximum Ratings may damage the device. Maximum Ratings are stress ratings only. Functional operation above the Recommended Operating Conditions is not implied. Extended exposure to stresses above the Recommended Operating Conditions may affect device reliability.

ELECTRICAL CHARACTERISTICS (T _F	$_{A}$ = +25°C, V _{CC} = +5.0 V to +15 V, unless otherwise noted.)
--	---

Characteristics	Symbol	Min	Тур	Max	Unit
Operating Supply Voltage Range	V _{CC}	4.5	_	16	V
Supply Current $V_{CC} = 5.0 \text{ V}, \text{ R}_{L} = \infty$ $V_{CC} = 15 \text{ V}, \text{ R}_{L} = \infty$, Low State (Note 1)	Icc		3.0 10	6.0 15	mA
Timing Error (R = 1.0 k Ω to 100 k Ω) (Note 2) Initial Accuracy C = 0.1 μ F Drift with Temperature Drift with Supply Voltage			1.0 50 0.1	_ _ _	% PPM/°C %/V
Threshold Voltage/Supply Voltage	V _{th} /V _{CC}	-	2/3	-	
Trigger Voltage $V_{CC} = 15 V$ $V_{CC} = 5.0 V$	V _T		5.0 1.67		V
Trigger Current	Ι _Τ	-	0.5	-	μA
Reset Voltage	V _R	0.4	0.7	1.0	V
Reset Current	I _R	_	0.1	-	mA
Threshold Current (Note 3)	l _{th}	_	0.1	0.25	μΑ
Discharge Leakage Current (Pin 7)	I _{dischg}	-	-	100	nA
Control Voltage Level $V_{CC} = 15 V$ $V_{CC} = 5.0 V$	V _{CL}	9.0 2.6	10 3.33	11 4.0	V
$\begin{array}{l} \text{Output Voltage Low} \\ I_{Sink} = 10 \text{ mA } (V_{CC} = 15 \text{ V}) \\ I_{Sink} = 50 \text{ mA } (V_{CC} = 15 \text{ V}) \\ I_{Sink} = 100 \text{ mA } (V_{CC} = 15 \text{ V}) \\ I_{Sink} = 200 \text{ mA } (V_{CC} = 15 \text{ V}) \\ I_{Sink} = 8.0 \text{ mA } (V_{CC} = 5.0 \text{ V}) \\ I_{Sink} = 5.0 \text{ mA } (V_{CC} = 5.0 \text{ V}) \end{array}$	V _{OL}	- - - - -	0.1 0.4 2.0 2.5 - 0.25	0.25 0.75 2.5 - - 0.35	V
Output Voltage High $V_{CC} = 15 \text{ V} (I_{Source} = 200 \text{ mA})$ $V_{CC} = 15 \text{ V} (I_{Source} = 100 \text{ mA})$ $V_{CC} = 5.0 \text{ V} (I_{Source} = 100 \text{ mA})$	V _{OH}	_ 12.75 2.75	12.5 13.3 3.3		V
Rise Time Differential Output	t _r	-	100	-	ns
Fall Time Differential Output	t _f	-	100	-	ns

'Supply current when output is high is typically 1.0 mA less.
Tested at V_{CC} = 5.0 V and V_{CC} = 15 V Monostable mode.
This will determine the maximum value of R_A + R_B for 15 V operation. The maximum total R = 20 MΩ.
T_{low} = 0°C for MC1455, T_{low} = -40°C for MC1455B, NCV1455B T_{high} = +70°C for MC1455, T_{high} = +85°C for MC1455B, T_{high} = +125°C for NCV1455B
NCV prefix is for Automotive and other applications requiring site and change control.



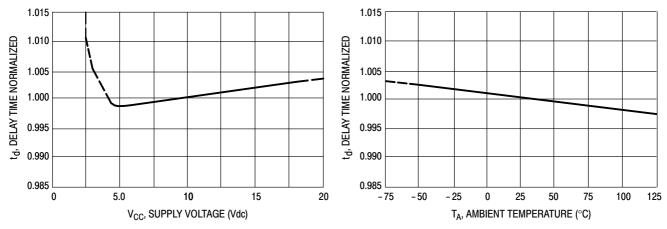
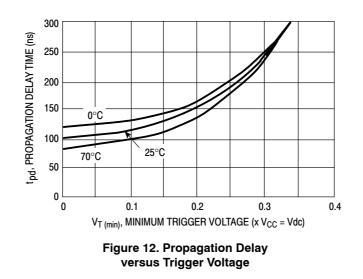


Figure 10. Delay Time versus Supply Voltage

Figure 11. Delay Time versus Temperature



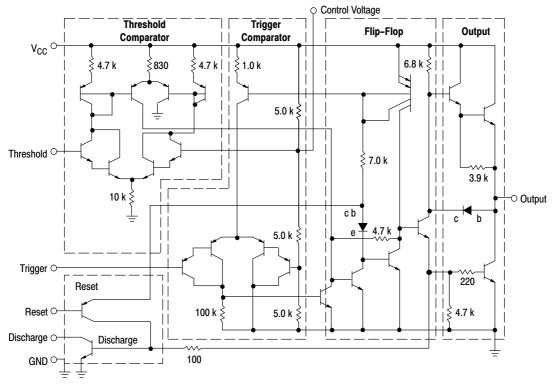


Figure 13. Representative Circuit Schematic

GENERAL OPERATION

The MC1455 is a monolithic timing circuit which uses an external resistor – capacitor network as its timing element. It can be used in both the monostable (one–shot) and astable modes with frequency and duty cycle controlled by the capacitor and resistor values. While the timing is dependent upon the external passive components, the monolithic circuit provides the starting circuit, voltage comparison and other functions needed for a complete timing circuit. Internal to the integrated circuit are two comparators, one for the input signal and the other for capacitor voltage; also a flip–flop and digital output are included. The comparator reference voltages are always a fixed ratio of the supply voltage thus providing output timing independent of supply voltage.

Monostable Mode

In the monostable mode, a capacitor and a single resistor are used for the timing network. Both the threshold terminal and the discharge transistor terminal are connected together in this mode (refer to circuit in Figure 14). When the input voltage to the trigger comparator falls below $1/3 V_{CC}$, the comparator output triggers the flip–flop so that its output sets low. This turns the capacitor discharge transistor "off" and drives the digital output to the high state. This condition allows the capacitor to charge at an exponential rate which is set by the RC time constant. When the capacitor voltage reaches $2/3 V_{CC}$, the threshold comparator resets the flip–flop. This action discharges the timing capacitor and returns the digital output to the low state. Once the flip–flop has been triggered by an input signal, it cannot be retriggered until the present timing period has been completed. The time that the output is high is given by the equation $t = 1.1 R_A C$. Various combinations of R and C and their associated times are shown in Figure 16. The trigger pulse width must be less than the timing period.

A reset pin is provided to discharge the capacitor, thus interrupting the timing cycle. As long as the reset pin is low, the capacitor discharge transistor is turned "on" and prevents the capacitor from charging. While the reset voltage is applied the digital output will remain the same. The reset pin should be tied to the supply voltage when not in use.

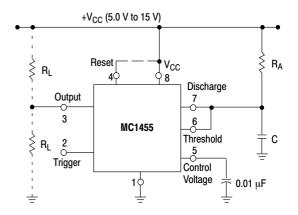
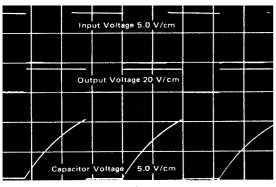


Figure 14. Monostable Circuit



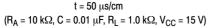


Figure 15. Monostable Waveforms

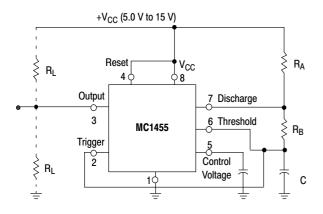


Figure 17. Astable Circuit

Astable Mode

In the astable mode the timer is connected so that it will retrigger itself and cause the capacitor voltage to oscillate between $1/3 V_{CC}$ and $2/3 V_{CC}$. See Figure 17.

The external capacitor changes to 2/3 V_{CC} through R_A and R_B and discharges to 1/3 V_{CC} through $R_B.$ By varying the ratio of these resistors the duty cycle can be varied. The charge and discharge times are independent of the supply voltage.

The charge time (output high) is given by:

 $t_1 \,=\, 0.695 \, (R_A \,+\, R_B) \, C$

The discharge time (output low) is given by:

$$t_2 = 0.695 (R_B) C$$

Thus the total period is given by:

 $T = t_1 + t_2 = 0.695 (R_A + 2R_B)C$

The frequency of oscillation is then:
$$f = \frac{1}{1} = \frac{1.44}{(\mathsf{R}_\mathsf{A} + 2\mathsf{R}_\mathsf{B})\mathsf{C}}$$

and may be easily found as shown in Figure 19.

The duty cycle is given by:

$$DC = \frac{R_B}{R_A + 2R_B}$$

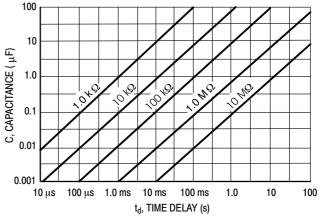


Figure 16. Time Delay

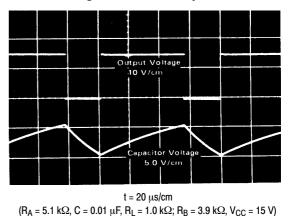


Figure 18. Astable Waveforms

To obtain the maximum duty cycle R_A must be as small as possible; but it must also be large enough to limit the discharge current (Pin 7 current) within the maximum rating of the discharge transistor (200 mA).

The minimum value of R_A is given by:

$$R_A \ge \frac{V_{CC}(Vdc)}{17 (A)} \ge \frac{V_{CC}(Vdc)}{0.2}$$

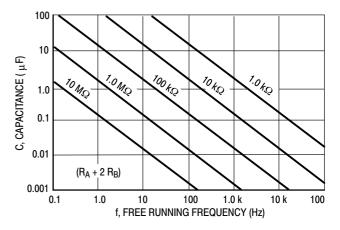


Figure 19. Free Running Frequency

APPLICATIONS INFORMATION

Linear Voltage Ramp

In the monostable mode, the resistor can be replaced by a constant current source to provide a linear ramp voltage. The capacitor still charges from 0 V_{CC} to 2/3 V_{CC}. The linear ramp time is given by:

$$t = \frac{2}{3} \frac{V_{CC}}{1}, \text{ where } I = \frac{V_{CC} - V_B - V_{BE}}{R_F}$$

If V_{B} is much larger than $V_{BE}\!,$ then t can be made independent of $V_{CC}\!.$

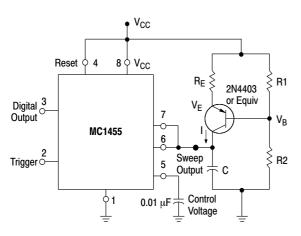
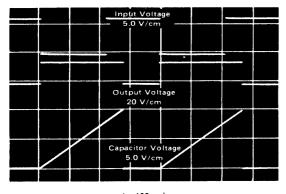


Figure 20. Linear Voltage Sweep Circuit



t = 100 µs/cm (R_E = 10 kΩ, R2 = 100 kΩ, R1 = 39 kΩ, C = 0.01 µF, V_{CC} = 15 V)



Missing Pulse Detector

The timer can be used to produce an output when an input pulse fails to occur within the delay of the timer. To accomplish this, set the time delay to be slightly longer than the time between successive input pulses. The timing cycle is then continuously reset by the input pulse train until a change in frequency or a missing pulse allows completion of the timing cycle, causing a change in the output level.

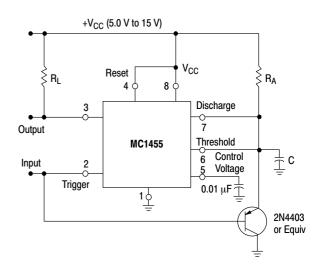
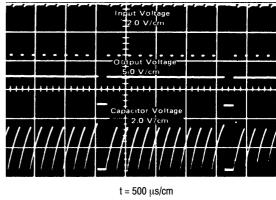


Figure 21. Missing Pulse Detector







Pulse Width Modulation

If the timer is triggered with a continuous pulse train in the monostable mode of operation, the charge time of the capacitor can be varied by changing the control voltage at Pin 5. In this manner, the output pulse width can be modulated by applying a modulating signal that controls the threshold voltage.

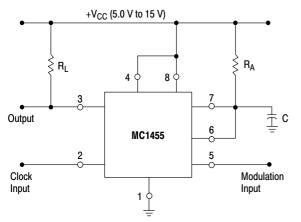
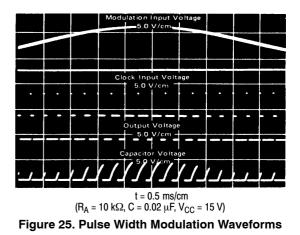


Figure 24. Pulse Width Modulator



Test Sequences

Several timers can be connected to drive each other for sequential timing. An example is shown in Figure 26 where the sequence is started by triggering the first timer which runs for 10 ms. The output then switches low momentarily and starts the second timer which runs for 50 ms and so forth.

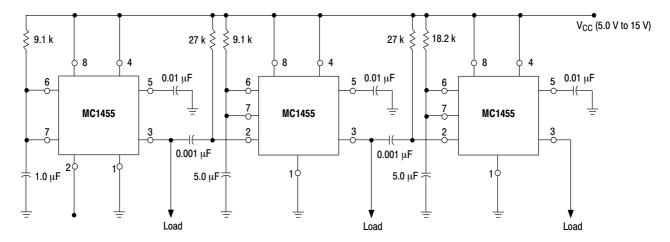


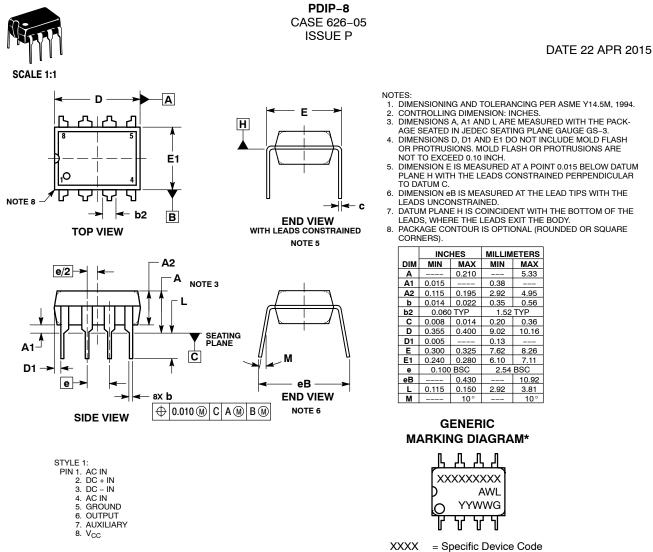
Figure 26. Sequential Timer

ORDERING INFORMATION

Device	Operating Temperature Range	Package	Shipping [†]
MC1455P1		PDIP-8	50 Units / Rail
MC1455P1G		PDIP-8 (Pb-Free)	50 Units / Rail
MC1455D		SOIC-8	98 Units / Rail
MC1455DG	$T_A = 0^{\circ}C$ to +70°C	SOIC-8 (Pb-Free)	98 Units / Rail
MC1455DR2		SOIC-8	2500 Units / Tape & Reel
MC1455DR2G		SOIC-8 (Pb-Free)	2500 Units / Tape & Reel
MC1455BD		SOIC-8	98 Units / Rail
MC1455BDG		SOIC-8 (Pb-Free)	98 Units / Rail
MC1455BDR2		SOIC-8	2500 Units / Tape & Reel
MC1455BDR2G	$T_A = -40^{\circ}C \text{ to } +85^{\circ}C$	SOIC-8 (Pb-Free)	2500 Units / Tape & Reel
MC1455BP1		PDIP-8	50 Units / Rail
MC1455BP1G		PDIP-8 (Pb-Free)	50 Units / Rail
NCV1455BDR2*		SOIC-8	2500 Units / Tape & Reel
NCV1455BDR2G*	$T_A = -40^{\circ}C \text{ to } +125^{\circ}C$	SOIC-8 (Pb-Free)	2500 Units / Tape & Reel

†For information on tape and reel specifications, including part orientation and tape sizes, please refer to our Tape and Reel Packaging Specifications Brochure, BRD8011/D.
*NCV prefix is for automotive and other applications requiring site and control changes.





A = Assembly Location

- WL = Wafer Lot
- YY = Year
- WW = Work Week
- G = Pb-Free Package

*This information is generic. Please refer to device data sheet for actual part marking. Pb–Free indicator, "G" or microdot " ■", may or may not be present.







*For additional information on our Pb–Free strategy and soldering details, please download the ON Semiconductor Soldering and Mounting Techniques Reference Manual, SOLDERRM/D.

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STYLE 1: PIN 1. EMITTER COLLECTOR 2. COLLECTOR 3. 4. EMITTER 5. EMITTER BASE 6. 7 BASE EMITTER 8. STYLE 5: PIN 1. DRAIN 2. DRAIN 3. DRAIN DRAIN 4. GATE 5. 6. GATE SOURCE 7. 8. SOURCE STYLE 9: PIN 1. EMITTER, COMMON COLLECTOR, DIE #1 COLLECTOR, DIE #2 2. З. EMITTER, COMMON 4. 5. EMITTER, COMMON 6 BASE. DIE #2 BASE, DIE #1 7. 8. EMITTER, COMMON STYLE 13: PIN 1. N.C. 2. SOURCE 3 GATE 4. 5. DRAIN 6. DRAIN DRAIN 7. DRAIN 8. STYLE 17: PIN 1. VCC 2. V2OUT V10UT З. TXE 4. 5. RXE 6. VFF 7. GND 8. ACC STYLE 21: PIN 1. CATHODE 1 2. CATHODE 2 3 CATHODE 3 CATHODE 4 4. 5. CATHODE 5 6. COMMON ANODE COMMON ANODE 7. 8. CATHODE 6 STYLE 25: PIN 1. VIN 2 N/C REXT З. 4. GND 5. IOUT 6. IOUT IOUT 7. 8. IOUT STYLE 29: BASE, DIE #1 PIN 1. 2 EMITTER, #1 BASE, #2 З. EMITTER, #2 4. 5 COLLECTOR, #2

STYLE 2: PIN 1. COLLECTOR, DIE, #1 2. COLLECTOR, #1 COLLECTOR, #2 3. 4 COLLECTOR, #2 BASE, #2 5. EMITTER, #2 6. 7 BASE #1 EMITTER, #1 8. STYLE 6: PIN 1. SOURCE 2. DRAIN 3. DRAIN SOURCE 4. SOURCE 5. 6. GATE GATE 7. 8. SOURCE STYLE 10: GROUND PIN 1. BIAS 1 OUTPUT 2. З. GROUND 4. 5. GROUND 6 BIAS 2 INPUT 7. 8. GROUND STYLE 14: PIN 1. N-SOURCE 2. N-GATE P-SOURCE 3 P-GATE 4. P-DRAIN 5 6. P-DRAIN N-DRAIN 7. N-DRAIN 8. STYLE 18: PIN 1. ANODE 2. ANODE SOURCE 3. GATE 4. 5. DRAIN 6 DRAIN CATHODE 7. CATHODE 8. STYLE 22 PIN 1. I/O LINE 1 2. COMMON CATHODE/VCC 3 COMMON CATHODE/VCC 4. I/O LINE 3 5. COMMON ANODE/GND 6. I/O LINE 4 7. I/O LINE 5 8. COMMON ANODE/GND STYLE 26: PIN 1. GND 2 dv/dt З. ENABLE 4. ILIMIT 5. SOURCE SOURCE 6. SOURCE 7. 8. VCC STYLE 30: DRAIN 1 PIN 1. DRAIN 1 2 GATE 2 З. SOURCE 2 4. SOURCE 1/DRAIN 2 SOURCE 1/DRAIN 2 5. 6.

STYLE 3: PIN 1. DRAIN, DIE #1 DRAIN, #1 2. DRAIN, #2 З. 4. DRAIN, #2 GATE, #2 5. SOURCE, #2 6. 7 GATE #1 8. SOURCE, #1 STYLE 7: PIN 1. INPUT 2. EXTERNAL BYPASS THIRD STAGE SOURCE GROUND З. 4. 5. DRAIN 6. GATE 3 SECOND STAGE Vd 7. FIRST STAGE Vd 8. STYLE 11: PIN 1. SOURCE 1 GATE 1 SOURCE 2 2. 3. GATE 2 4. 5. DRAIN 2 6. DRAIN 2 DRAIN 1 7. 8. DRAIN 1 STYLE 15: PIN 1. ANODE 1 2. ANODE 1 ANODE 1 3 ANODE 1 4. 5. CATHODE, COMMON CATHODE, COMMON CATHODE, COMMON 6. 7. CATHODE, COMMON 8. STYLE 19: PIN 1. SOURCE 1 GATE 1 SOURCE 2 2. 3. GATE 2 4. 5. DRAIN 2 6. MIRROR 2 7. DRAIN 1 8. **MIRROR 1** STYLE 23: PIN 1. LINE 1 IN COMMON ANODE/GND COMMON ANODE/GND 2. 3 LINE 2 IN 4. LINE 2 OUT 5. COMMON ANODE/GND COMMON ANODE/GND 6. 7. LINE 1 OUT 8. STYLE 27: PIN 1. ILIMIT 2 OVI 0 UVLO З. 4. INPUT+ 5. SOURCE SOURCE 6. SOURCE 7. 8 DRAIN

DATE 16 FEB 2011

STYLE 4: PIN 1. 2. ANODE ANODE ANODE З. 4. ANODE ANODE 5. 6. ANODE 7 ANODE COMMON CATHODE 8. STYLE 8: PIN 1. COLLECTOR, DIE #1 2. BASE, #1 BASE, #2 З. COLLECTOR, #2 4. COLLECTOR, #2 5. 6. EMITTER, #2 EMITTER, #1 7. 8. COLLECTOR, #1 STYLE 12: PIN 1. SOURCE SOURCE 2. 3. GATE 4. 5. DRAIN 6. DRAIN DRAIN 7. 8. DRAIN STYLE 16: PIN 1. EMITTER, DIE #1 2. BASE, DIE #1 EMITTER, DIE #2 3 BASE, DIE #2 4. 5. COLLECTOR, DIE #2 6. COLLECTOR, DIE #2 COLLECTOR, DIE #1 7. COLLECTOR, DIE #1 8. STYLE 20: PIN 1. SOURCE (N) GATE (N) SOURCE (P) 2. 3. 4. GATE (P) 5. DRAIN 6. DRAIN DRAIN 7. 8. DRAIN STYLE 24: PIN 1. BASE 2. EMITTER 3 COLLECTOR/ANODE COLLECTOR/ANODE 4. 5. CATHODE 6. CATHODE COLLECTOR/ANODE 7. 8. COLLECTOR/ANODE STYLE 28: PIN 1. SW_TO_GND 2. DASIC OFF DASIC_SW_DET З. 4. GND 5. 6. V MON VBULK 7. VBULK 8 VIN

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