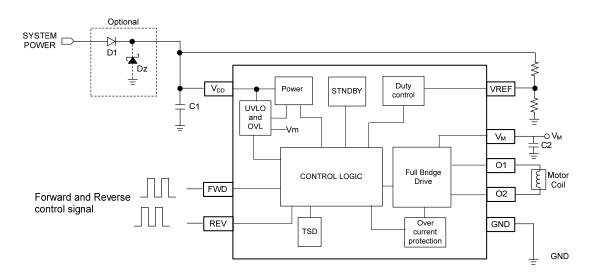


Typical Applications Circuit



Note:

4. C1 is for power stabilization and to strengthen the noise immunity, the recommended capacitance is 100nF to $1\mu F$ or more.

C2 is a re-circulating capacitor for back rush voltage and recommended capacitance is 100nF for low current applications to $10\mu F$ or more for large current applications. See application note section

Pin Descriptions

Package: SO-8

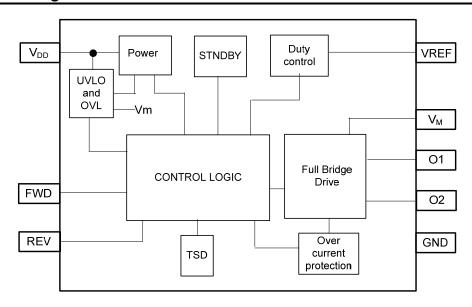
Pin#	Pin Name	Function					
1	OUT1	Driver output					
2	V _M	Power Supply Voltage					
3	V_{DD}	Power Supply Voltage					
4	FWD	Forward Control Input (logic level, 5.5V max)					
5	REV	Reverse Control Input (logic level, 5.5V max)					
6	V_{REF}	Input reference voltage to set the internal PWM oscillator duty ratio					
7	OUT2	Driver Output					
8	GND	Ground					

Package: SO8-EP

Pin #	Pin Name	Function				
1	OUT1	Driver output				
2	V_{M}	Power Supply Voltage				
3	V_{DD}	Power Supply Voltage				
4	FWD	Forward Control Input (logic level, 5.5V max)				
5	REV	Reverse Control Input (logic level, 5.5V max)				
6	V_{REF}	Input reference voltage to set the internal PWM oscillator duty ratio				
7	OUT2	Driver Output				
8	GND	Ground				
Pad	Pad	The exposed pad is for thermal dissipation and it is internally connected to the ground. On the PCB layout, it can be connected to GND or left open circuit.				



Functional Block Diagram



Absolute Maximum Ratings (Note 5) @T_A = +25°C, unless otherwise specified.)

Symbol	Charact	eristic	Values	Unit
V_{M} and $V_{DD,}$	Supply voltage (Note 6)		24	V
$V_{OUT1,}V_{OUT2}$ V_{REF}	All other pins except FWD and REV pir	ns	24	V
V_{FWD} and V_{REV}	FWD and REV pin voltage		7	V
V _{REVERSE}	Reverse supply Voltage on all pins	-0.3	V	
loutput	Output current (source and sink)- Peal	k	1500	mA
Б	Deckers revending in the	SO8 (Note 7)	1043	mW
P_{D}	Package power dissipation	SO8-EP (Note 8)	2980	mW
Ts	Storage temperature range		-65 to +150	°C
TJ	Maximum junction temperature	150	°C	
ESD HBM	Human Body Model ESD withstand capability	6	kV	

Notes:

- 5. Stresses greater than the 'Absolute Maximum Ratings' specified above, may cause permanent damage to the device. These are stress ratings only; functional operation of the device at these or any other conditions exceeding those indicated in this specification is not implied. Device reliability may be affected by exposure to absolute maximum rating conditions for extended periods of time
- 6. The absolute maximum supply voltage of 24V is a transient stress rating and is not meant as a functional operating condition. It is not recommended to operate the device at the absolute maximum rated conditions for any period of time.
- 7. SO-8 soldered to minimum recommended landing pads (see Package Outline Dimension section) on a 1"x1" two-layer 2oz.copper FR4 PCB (1.6mm thickness) without any via or copper flood on the bottom layer. See thermal de-rating curves in the thermal performance section.
- 8. SO-8EP exposed pad soldered to minimum recommended landing pads (see Package Outline Dimension section) on a 2"x2" two-layer 2oz.copper FR4 PCB (1.6mm thickness) with four thermal vias in the exposed PAD to the copper flood on the bottom layer. See thermal de-rating curves in the thermal performance section.

Recommended Operating Conditions

Symbol	Characteristic	Characteristic Conditions				
V_{DD}	Supply Voltage	Operating	3	18	V	
т.	Operating Temperature Bange	Operating, SO8 package	-40	+85	°C	
IA	Operating Temperature Range	Operating, SO8-EP package	-40	+105	°C	



Electrical Characteristics (Note 9)(@T_A = +25°C, V_{DD} = 12V, unless otherwise specified.)

Symbol	Characteristics	Conditions	Min	Тур	Max	Unit
I_{DD}	Supply Current	No Load	_	0.85	2.5	mA
I _{DD_STNDBY}	Standby Supply Current	FWD=REV=LOW (GND)	_	32	45	uA
V _{UV_TH}	Under Voltage lock Out Threshold	Voltage Decreasing	_	2.6		V
V_{UVLO_R}	Under Voltage Lock Out Release Threshold	Voltage Increasing	_	2.8	3.0	V
V_{UV_HYS}	Under Voltage Hysteresis	Voltage Increasing	_	200	l	mV
V _{OV_TH}	Overvoltage Threshold	Voltage Increasing	_	20.7	24	V
V_{OV_RLTH}	Overvoltage Release Threshold	Voltage Decreasing	17.0	19		V
.,	Output Voltage High	I _{OUT} = 300mA, T _A = -40°C to +105°C	V _{DD} - 0.25	V _{DD} - 0.15		V
V _{OH}		I _{OUT} =500mA T _A = -40°C to +105°C	V _{DD} - 0.43	V _{DD} - 0.25	_	V
	Output Valtage Lavy	I _{OUT} = 300mA, T _A = -40°C to +105°C	_	0.15	0.25	V
V _{OL}	Output Voltage Low	I _{OUT} = 500mA, T _A = -40°C to +105°C	_	0.25	0.43	V
		I _{OUT} = 300mA, T _A = -40°C to +105°C	_	0.3	0.5	V
V _{OH} + V _{OL}	Output voltage of N- and PMOS and bond wire voltage drop combined	$I_{OUT} = 500$ mA, $T_A = -40$ °C to +105°C	_	0.5	0.86	V
		$I_{OUT} = 300 \text{mA}, V_{DD} = 3V$ $T_A = -40^{\circ}\text{C to } +105^{\circ}\text{C}$	_	0.39	0.66	V
Pou T.	Combined N- and PMOS R _{DSON} including	$I_{OUT} = 500$ mA, $T_A = -40$ °C to +105°C	_	1	1.72	Ω
R _{ON_Total}	bond wire resistance	I_{OUT} = 300mA, V_{DD} = 3V T_A = -40°C to +105°C	_	1.3	2.2	Ω
I _{LIM_TH}	Over current protection threshold		1.2	1.5		Α
V_{REF}	V _{REF} voltage range (DC voltage speed control mode)		3	_	V _{DD} (18V max)	V
I _{VREF}	V _{REF} bias current (DC voltage speed control mode)	V _{REF} = V _{DD}	-15	0	15	μΑ
Fout	Output PWM switching frequency	VREF control mode	20	26.5	35	kHz
1 001	(Internal PWM oscillator)	PWM speed control mode	20	_	100	kHz
F _{FWD_REV}	Input PWM frequency of speed control signal	PWM control mode	20	_	100	kHz
T _{DEAD}	Dead time between current reversal	$V_{DD} = 3V \text{ to } 18V$ $T_A = -40^{\circ}\text{C to } +105^{\circ}\text{C}$	2.1	3	3.9	μs
t _{SDN_DELAY}	Shutdown delay – Internal circuits active after FWD = REV = L (except from brake mode)	FWD = GND REV = GND	125	180	_	μs
V_{FWDH}	FWD Input H Level		2	_	5.5	V
V_{FWDL}	FWD Input L Level		0	_	0.8	V
I _{FWDH}	FWD pin current – H Level	FWD pin: V _{FWD} = 5V	_	50	_	μΑ
I _{FWDL}	FWD pin current – L Level	FWD pin: V _{FWD} = 0V		50	_	μΑ
V_{REVH}	REV Input H Level		2	_	5.5	V
V_{REVL}	REV Input L Level		0	_	0.8	V
I _{REVH}	REV pin current – H Level	REV pin: V _{RVS} = 5V	_	50	_	μΑ
I _{REVL}	REV pin current – L Level	REV pin: V _{RVS} = 0V	_	50	_	μΑ
D _{PWM_MIN}	Output minimum duty ratio		0%		100	%
Tj_ _{SDN_TH}	IC junction temperature thermal shutdown threshold		_	165	_	°C
Tj_sdn_hyst	IC junction temperature thermal shutdown hysteresis		_	25	_	°C

9. Typical data is at T_A = +25°C, V_{DD} = 12V. The maximum and minimum parameters values over the operating temperature range are not tested in production, they are guaranteed by design, characterization and process control. Note:

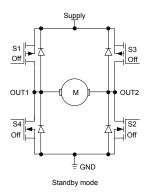
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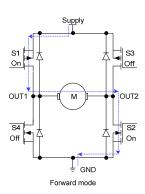


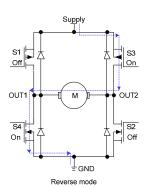
Application Note

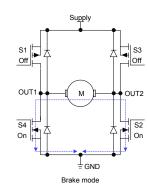
Rotation Control and Standby Modes

The device has FWD and REV pins for controlling the motor rotation directions. The device has four motor operation modes: 1) Standby mode, 2) Forward mode, 3) Reverse mode and 4) Brake mode. The four modes are controlled by the FWD and REV logic pins.









FWD	REV	V _{REF}	OUT1	OUT2	Operating mode		
L	L	x	Open	Open	Standby mode – All switches are off		
Н	L	3V to V _{DD}	Н	L	L Forward mode – Current flows from OUT1 to OUT2; V _{REF} duty control		
L	Н	3V to V _{DD}	L	Н	H Reverse mode – Current flows from OUT2 to OUT1; V _{REF} duty control		
Н	Н	х	L	L	Brake mode – Short circuit brake with low side switches on		
PWM	L	V_{DD}	Н	PWM	Forward mode – Current flows from OUT1 to OUT2; PWM control mode		
L	PWM	V_{DD}	PWM	Н	Reverse mode – Current flows from OUT2 to OUT1; PWM control mode		
Н	Н	Х	L	L	Brake mode – Short circuit brake with low side switches on		

In the brake mode, switches S2 and S4 are ON allowing the motor to stop quickly. All the internal control circuits are fully operational.

In the standby mode all the output drive switches are off and additionally, the internal circuits are also turned off to minimize power consumption. The power consumption in the standby mode is less than in the brake mode. If running motor enters the Standby mode, due to the body diodes the motor free wheels to idle state. Whenever the motor enters the standby mode from any mode (except the brake mode) the control logic will remain active in previous mode for at least 125µs before shutting down the internal circuits. To prevent device entering the standby mode during operating mode changes, the mode change signals should be completed within 125µs.

In the forward mode, with switches S1-S2 ON and S3-S4 OFF, OUT1 is high and OUT2 is low. The motor current flows from OUT1 to OUT2. In the reverse mode, switches S1-S4 are ON while S1-S2 are OFF to allow motor current flow from OUT2 to OUT1.

In the forward or reverse mode, for V_{REF} speed control, the output drive duty ratio is generated internally based on the voltage on the V_{REF} pin. For PWM speed control, external PWM signals applied to the FWD or REV pins control the PWM switching of the low side S2 (forward mode) or S4 (reverse mode). See application section for further details.

The ZXBM5210 has three modes of speed control: VREF speed control mode, PWM speed control mode and by adjusting the supply voltage



Application Note (cont.)

Motor Speed Control with DC Voltage on V_{REF} Pin

Motor speed can be controlled by adjusting the DC voltage into the V_{REF} pin. The output drive PWM duty ratio is defined by the ratio of the V_{REF} voltage to the supply V_{DD} voltage.

In V_{REF} speed control mode, FWD and REV pins are only used for direction control and therefore high frequency PWM control signal should not be applied to the FWD and REV pins. If repetitive direction changes required, it is recommended to keep direction change frequency of below 400Hz.

The speed and direction control is given by:

FWD	REV	V _{REF}	OUT1	OUT2	Operating mode		
L	L	х	Open	Open	Standby mode – All switches are off		
Н	L	V_{DD}	Н	L	L Forward mode – Current flows from OUT1 to OUT2; 100% duty		
Н	L	3V to V _{DD}	Н	L	L Forward mode – Current flows from OUT1 to OUT2; V _{REF} duty control		
L	Н	V_{DD}	L	Н	Reverse mode – Current flows from OUT2 to OUT1; 100% duty		
L	Н	3V to V _{DD}	L	Н	Reverse mode – Current flows from OUT2 to OUT1; V _{REF} duty control		
Н	Н	х	L	L	Brake mode – Short circuit brake with low side switches on		

Motor Speed Control with a PWM Input Signal

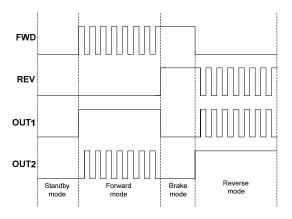
Motor speed can be controlled by adjusting the duty cycle of the PWM speed control signal into the FWD or REV while keeping the V_{DD} pin at the nominal motor voltage. In this mode the input voltage on the V_{REF} pin must be greater than or equal to V_{DD} .

In PWM speed control mode the high side switches S1 and S3 are kept fixed while the low side switches S2 or S4 are switched. In the forward mode, S1 is kept switched on, S2 is switched in accordance with the PWM signal and S3 and S4 are switched off. In reverse mode, S3 is switched on, S4 is switched in accordance with the PWM signal and S1 and S2 are switched off.

The speed and direction control is given by:

FWD	REV	V _{REF}	OUT1	OUT2	Operating mode
L	L	Х	Open	Open	Standby mode – All switches are off
Н	L	V_{DD}	Н	L	Forward mode – Current flows from OUT1 to OUT2; 100% duty
L	Н	V_{DD}	L	Н	Reverse mode – Current flows from OUT2 to OUT1; 100% duty
Н	Н	Х	L	L	Brake mode – Short circuit brake with low side switches on
PWM	L	V_{DD}	Н	PWM	Forward mode – Current flows from OUT1 to OUT2; PWM control mode
L	PWM	V_{DD}	PWM	Н	Reverse mode – Current flows from OUT2 to OUT1 PWM control mode
Н	Н	Х	L	L	Brake mode – Short circuit brake with low side switches on

The motor speed is proportional to the input PWM signal duty. For example, for a 12V motor the V_{DD} pin is maintained at 12V while varying the PWM control signal duty to adjust the motor speed linearly. The timing diagram below shows the output O1 and O2 in relation to PWM speed control signal at PWM pin.



Frequency of PWM speed control signal can be between 8kHz to 100kHz. Recommended typical PWM signal frequency is 25kHz to keep switching frequency away from the audible band.



Application Note (cont.)

Motor Speed Control with a PWM Input Signal (cont.)

Depending on the motor design and its inertia the minimum start-up PWM duty required can be typically between 30% - 50%. While the motor is rotating minimum PWM duty can generally be reduced down to 20%. How small the PWM duty can be without stalling the motor depends on the motor mechanical and coil design parameters and not limited by the output capability of the device. If voltage at V_{DD} is lower than the nominal motor voltage, both start-up PWM duty and minimum running PWM duty required will be higher.

Motor Speed Control by DC Supply Voltage

Motor speed can be controlled by varying the V_{DD} supply voltage while the FWD and REV pins are set to either a logic high or low depending on forward or reverse direction needed. The V_{REF} must be equal to V_{DD} in this mode. For example, if the V_{DD} for a 12V motor is changed from 12V to 3V the speed will be reduced from 100% to 25%.

Re-Circulating Capacitor

During motor operation when the low side switch is turned off the bridge or the motor voltage may overshoot to high levels if there is no current path for the energy in the motor to flow. Such high voltages can damage the IC. A current path can be provided by adding a bypass capacitor from the V_{DD} or V_{M} to the GND. The value of the bypass capacitor depends on the motor coil design, motor current, motor voltage and the IC voltage limits. This could be in the range of $0.47\mu F$ for low current applications to 10uF or more for large current applications.

Dead-Time

During motor current reversal (for motor rotation direction reversal), switch position changes between S1-S2 and S3-S4. Such change may result in cross conduction between high side and low side MOSFETS, e.g. S1 and S4 or S2 and S3. To prevent cross conduction the IC provides a dead time 3µs typical during current reversal or fast turn on of the low side MOSFETs.

Back-Rush Voltage

Depending on motor characteristics, the environment and the ambient conditions back-rush voltage (at the bridge) may fluctuate during brush commutation and PWM switching. Due to the energy in the coil this back-rush voltage can reach high levels if no adequate alternative current path is provided when inductor current path is interrupted. The back-rush voltage overshoot should not be allowed to go beyond the operating voltage range of the IC. This backrush voltage overshoot can be minimized by using a re-circulating bypass capacitor at the V_{DD} and V_M pins.

The value of the re-circulating bypass capacitor depends on the motor coil design, motor current, motor voltage and the IC voltage limits. This could be 0.47µF for low current applications to 10uF or more for large current applications.

Under Voltage Lockout

To make sure the minimum voltage needed to operate the driver is supplied, the driver has an under voltage lock out. At start up the device will only start if the supply voltage is typically 2.8 or greater. During normal operation, the device will switch off all the output switches and power down if the supply voltage drops below 2.6V typical.

Over Voltage Protection

When the supply voltage exceeds 20.7V (typical) the driver will turn-off all the output switches. The driver will return to normal condition if the supply voltage drops below 19V (typical) provided no other fault condition or signals are preventing it to enter normal operation.

In-Rush Current

It is recommended to use the PWM duty cycle to control the average voltage supplied to the motor during power up, standby mode, brake mode or during motor direction reversal. If a PWM signal is not available it is recommended to use a current limiting resistor or other protection devices if needed.

Over-Current Setting and Protection

The internal over current protection (OCP) threshold is 1.5A typical at 12V supply +25°C.

When the motor current exceeds the OCP threshold for longer than 10µs typical on any of the H-Bridge switches, the device will switch of all the output switches and remain off for 5ms typical. The IC returns to normal operation after the 5ms if over current condition has gone away. If the motor current is still higher than the OCP threshold, the device will enter another 5ms standby mode.

Thermal Shutdown

The device has an internal thermal shutdown to prevent a thermal run-away scenario. The thermal shutdown is triggered when the junction temperature of the device reaches +165°C. It will remain in standby mode until the junction temperature falls by +25°C.

Reverse Voltage Protection

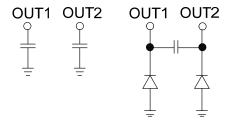
If reverse protection is needed this can be achieved by adding an external diode to the V_{DD} and V_M pins.



Application Note (cont.)

Motor Electromagnetic and Audible Noise

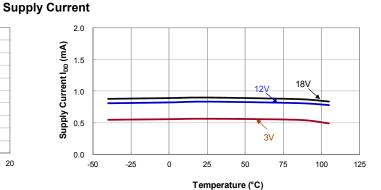
To help reduce electromagnetic and audible nose, capacitor can be connected from OUT1 to GND and OUT2 to GND pins. Alternatively, a capacitor can be connected between OUT1 and OUT2 with diodes between the output pins and GND as shown below. To prevent large currents it is recommended to keep any capacitor used at the output pins as small as possible and less than $1\mu F$.





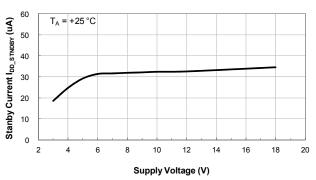
2.0 T_A = +25 °C 1.8 Supply Current I_{DD} (mA) 1.8 1.0 1.0 0.8 0.6 0.4 0.4 0.3 0.2 0.0 20 Supply Voltage (V)

Supply Current vs. Supply Voltage

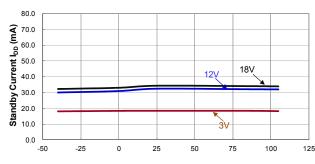


Supply Current vs. Temperature

Standby Current

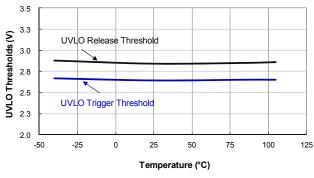


Standby Supply Current vs. Supply Voltage



Standby Supply Current vs. Temperature

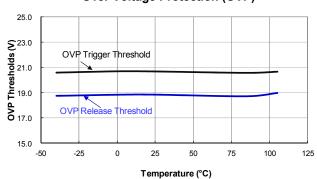
Under Voltage Lockout (UVLO)



UVLO Trigger and Release Thresholds vs Temperature

Over Voltage Protection (OVP)

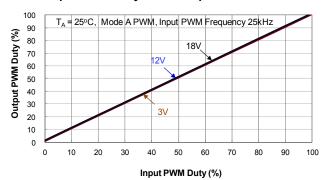
Temperature (°C)



OVP Trigger and Release Thresholds vs Temperature

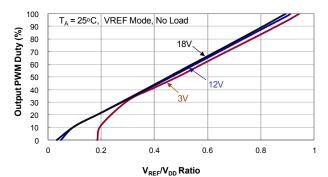


Output PWM Duty in PWM Speed Control Mode



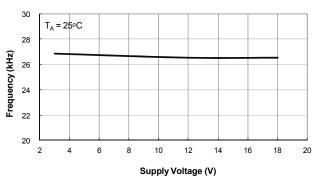
Output PWM Duty vs Input PWM Duty

Output PWM Duty in VREF Speed Control Mode

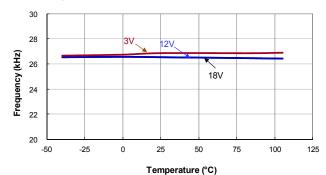


OutputPWM Duty vs V_{REF}/V_{DD} Ratio

PWM Oscillator Frequency



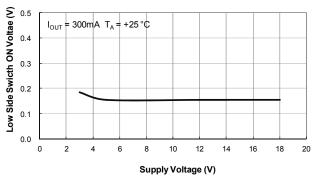
PWM Oscillator Frequency vs. Supply Voltage



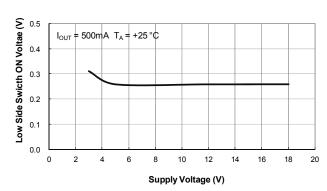
PWM Oscillator Frequency vs. Temperature



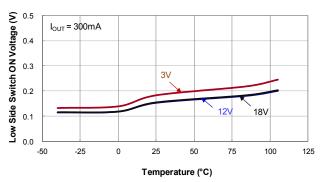
Low Side Switch On Voltage (Vol)



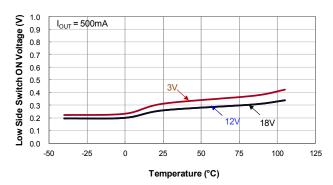
Low Side Switch ON Voltage V_{OL} vs. Supply Voltage



Low Side Switch ON Voltage V_{OL} vs. Supply Voltage

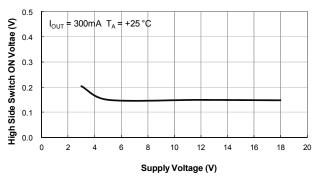


Low Side Switch On Voltage V_{OL} vs. Temperature

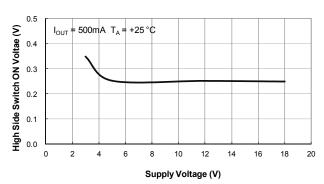


Low Side Switch On Voltage V_{OL} vs. Temperature

High Side Switch On Voltage (VDD - VOH)



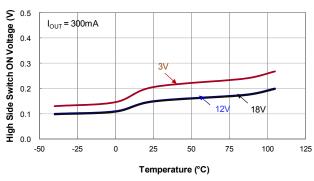
High Side Switch ON Voltage (V_{DD} - V_{OH}) vs Supply Voltage

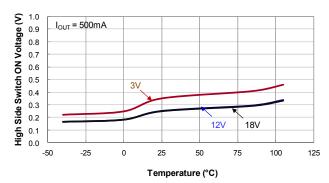


High Side Switch ON Voltage ($V_{\rm DD}\text{-}V_{\rm OH}$) vs Supply Voltage



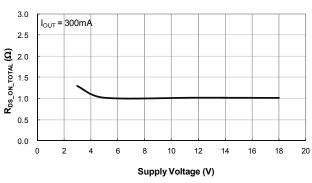
High Side Switch On Voltage (V_{DD} - V_{OH}) (cont.)



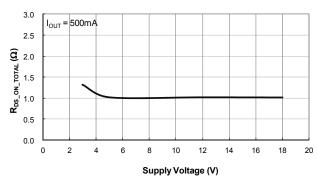


 $\label{eq:high-Side-Switch-On-Voltage} \ \ (V_{DD}\text{-}V_{OL}) \ vs. \ Temperature$

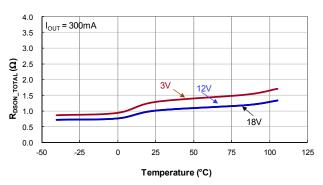
H-Bridge Resistance - Total Resistance On (RDSON_TOTAL) of the High and the Low Side Switches



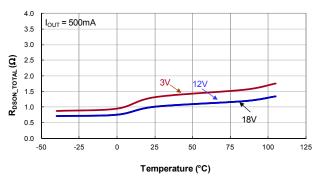
High+Low Side Resistance R_{DSON_TOTAL} vs. Voltage



 $\label{eq:high+LowSide} \textbf{High+Low Side Resistance} \; \textbf{R}_{\textbf{DSON_TOTAL}} \; \textbf{vs. Voltage}$



High+Low Side ON Resistance R_{DSON_TOTAL} vs. Temperature



High+Low Side ON Resistance $R_{\mbox{\scriptsize DSON_TOTAL}}$ vs. Temperature

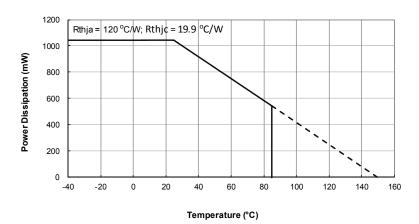


Thermal Performance

(1) Package Type: SO-8

MSOP8-EP Power Dissipation De-rating Curve (Note 11)

T _A (°C)	-40	0	25	50	60	70	80	85	90	95	100	105	110	120	125	130	140	150
P _D (mW)	1043	1043	1043	835	751	668	584	543	501	459	417	376	334	250	209	167	83	0



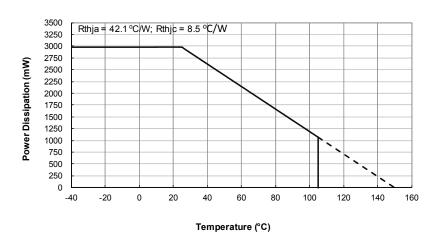
SO-8 Thermal Derating Curve

Note: 11. SO-8 soldered to minimum recommended landing pads (see Package Outline Dimension section) on a 1"x1" two-layer 2oz.copper FR4 PCB (1.6mm thickness) without any via or copper flood on the bottom layer.

(2) Package Type: SO-8EP

SO-8EP Power Dissipation De-rating Curve (Note 12)

00-0E1 1 0V	VCI DIS	Sipatio	וו ספ-וו	ating C	ul ve (iv	IUIC IZ												
T _A (°C)	-40	0	25	50	60	70	80	85	90	95	100	105	110	120	125	130	140	150
Pn (mW)	2980	29080	2980	2384	2146	1907	1669	1550	1430	1430	1192	1073	954	715	596	477	238	0

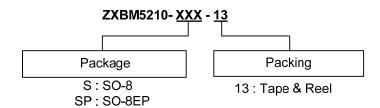


SO-8EP Thermal Derating Curve

Note: 12. SO-8EP exposed pad soldered to minimum recommended landing pads (see Package Outline Dimension section) on a 2"x2" two-layer 2oz.copper FR4 PCB (1.6mm thickness) with four thermal vias in the exposed PAD to the copper flood on the bottom layer



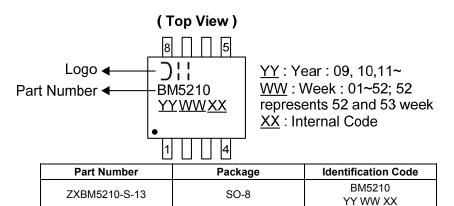
Ordering Information



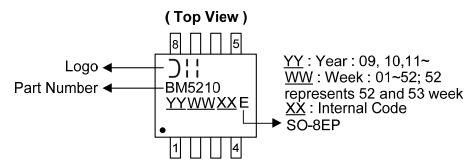
Part Number	Backage Code	Packaging	13" Tape and Reel				
Part Number	Package Code	Packaging	Quantity	Part Number Suffix			
ZXBM5210-S-13	S	SO-8	2500/Tape & Reel	-13			
ZXBM5210-SP-13	SP	SO-8EP	2500/Tape & Reel	-13			

Marking Information

(1) Package type: SO-8



(2) Package type: SO-8EP



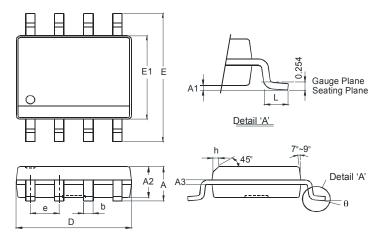
Part Number	Package	Identification Code
ZXBM5210-SP-13	SO-8EP	BM5210
ZABINI3210-3F-13	30-6EF	YY WW XX E



Package Outline Dimensions (All dimensions in mm.)

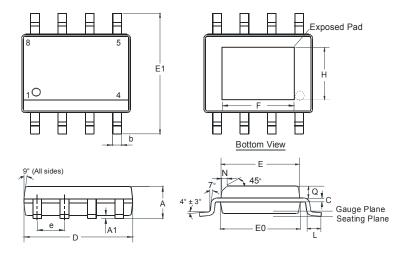
Please see AP02002 at http://www.diodes.com/datasheets/ap02002.pdf for latest version.

(1) Package Type: SO-8



SO-8				
Dim	Min	Max		
Α	-	1.75		
A1	0.10	0.20		
A2	1.30	1.50		
A3	0.15	0.25		
b	0.3	0.5		
D	4.85	4.95		
Е	5.90	6.10		
E1	3.85	3.95		
е	1.27 Typ			
h	-	0.35		
L	0.62	0.82		
θ	0°	8°		
All Dimensions in mm				

(2) Package Type: SO-8EP



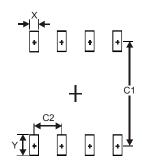
SO-8EP (SOP-8L-EP)				
Dim	Min	Max	Тур	
Α	1.40	1.50	1.45	
A1	0.00	0.13	-	
b	0.30	0.50	0.40	
ဂ	0.15	0.25	0.20	
D	4.85	4.95	4.90	
Е	3.80	3.90	3.85	
E0	3.85	3.95	3.90	
E1	5.90	6.10	6.00	
е	-	-	1.27	
F	2.75	3.35	3.05	
I	2.11	2.71	2.41	
П	0.62	0.82	0.72	
N	-	-	0.35	
Ø	0.60	0.70	0.65	
All Dimensions in mm				



Suggested Pad Layout

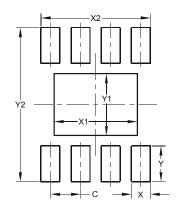
Please see AP02001 at http://www.diodes.com/datasheets/ap02001.pdf for the latest version.

(1) Package Type: SO-8



Dimensions	Value (in mm)
X	0.60
Y	1.55
C1	5.4
C2	1.27

(2) Package Type: SO-8EP



Dimensions	Value (in mm)	
С	1.270	
Х	0.802	
X1	3.502	
X2	4.612	
Y	1.505	
Y1	2.613	
Y2	6.500	



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