

Figure 3. Representative Schematic Diagram Component values are nominal

MAXIMUM RATINGS (Full operating ambient temperature range applies, unless otherwise noted.)

Rating	Symbol	Value	Unit
Cathode to Anode Voltage	V _{KA}	37	V
Cathode Current Range, Continuous	۱ _K	-100 to +150	mA
Reference Input Current Range, Continuous	I _{ref}	-0.05 to +10	mA
Operating Junction Temperature	Т _Ј	150	°C
Operating Ambient Temperature Range	T _A		°C
TL431I, TL431AI, TL431BI		-40 to +85	
TL431C, TL431AC, TL431BC		0 to +70	
NCV431AI, NCV431B, TL431BV, SCV431AI		-40 to +125	
Storage Temperature Range	T _{stg}	-65 to +150	°C
Total Power Dissipation @ $T_A = 25^{\circ}C$	P _D		W
Derate above 25°C Ambient Temperature			
D, LP Suffix Plastic Package		0.70	
P Suffix Plastic Package		1.10	
DM Suffix Plastic Package		0.52	
Total Power Dissipation @ $T_C = 25^{\circ}C$	PD		W
Derate above 25°C Case Temperature			
D, LP Suffix Plastic Package		1.5	
P Suffix Plastic Package		3.0	
ESD Rating (Note 1)			V
Human Body Model per JEDEC JESD22-A114F	HBM	>2000	
Machine Model per JEDEC JESD22-A115C Charged Device Model per JEDEC JESD22-C101E		>200 >500	
Charged Device Model per JEDEC JESD22-C101E	CDM	>500	

Stresses exceeding those listed in the Maximum Ratings table may damage the device. If any of these limits are exceeded, device functionality should not be assumed, damage may occur and reliability may be affected.

1. This device contains latch-up protection and exceeds ±100 mA per JEDEC standard JESD78.

RECOMMENDED OPERATING CONDITIONS

Condition	Symbol	Min	Max	Unit
Cathode to Anode Voltage	V _{KA}	V _{ref}	36	V
Cathode Current	١ _K	1.0	100	mA

Functional operation above the stresses listed in the Recommended Operating Ranges is not implied. Extended exposure to stresses beyond the Recommended Operating Ranges limits may affect device reliability.

THERMAL CHARACTERISTICS

Characteristic	Symbol	D, LP Suffix Package	P Suffix Package	DM Suffix Package	Unit
Thermal Resistance, Junction-to-Ambient	R_{\thetaJA}	178	114	240	°C/W
Thermal Resistance, Junction-to-Case	$R_{ ext{ heta}JC}$	83	41	_	°C/W

ELECTRICAL CHARACTERISTICS (T_A = 25°C, unless otherwise noted.)

			TL431I			TL431C		
Characteristic	Symbol	Min	Тур	Max	Min	Тур	Max	Unit
$ \begin{array}{l} \mbox{Reference Input Voltage (Figure 1)} \\ V_{KA} = V_{ref}, \ I_K = 10 \ mA \\ T_A = 25^\circ C \\ T_A = T_{low} \ to \ T_{high} \ (Note 2) \end{array} $	V _{ref}	2.44 2.41	2.495 -	2.55 2.58	2.44 2.423	2.495 -	2.55 2.567	V
Reference Input Voltage Deviation Over Temperature Range (Figure 1, Notes 3, 4) V_{KA} = V_{ref} , I_K = 10 mA	ΔV_{ref}	-	7.0	30	-	3.0	17	mV
$\begin{array}{l} \mbox{Ratio of Change in Reference Input Voltage to Change} \\ \mbox{in Cathode to Anode Voltage} \\ \mbox{I}_{K} = 10 \mbox{ mA (Figure 2),} \\ \mbox{$\Delta V_{KA} = 10 \mbox{ V to V_{ref}}$} \\ \mbox{$\Delta V_{KA} = 36 \mbox{ V to 10 \mbox{ V}}$} \end{array}$	$rac{\Delta V_{ref}}{\Delta V_{KA}}$		-1.4 -1.0	-2.7 -2.0		-1.4 -1.0	-2.7 -2.0	mV/V
$ \begin{array}{l} \mbox{Reference Input Current (Figure 2)} \\ I_K = 10 \mbox{ mA, R1} = 10 \mbox{ k, R2} = \infty \\ T_A = 25^\circ \mbox{C} \\ T_A = T_{low} \mbox{ to } T_{high} \mbox{ (Note 2)} \end{array} $	I _{ref}		1.8 _	4.0 6.5		1.8 _	4.0 5.2	μΑ
Reference Input Current Deviation Over Temperature Range (Figure 2, Note 3) $I_K = 10 \text{ mA}, \text{ R1} = 10 \text{ k}, \text{ R2} = \infty$	ΔI_{ref}	_	0.8	2.5	-	0.4	1.2	μΑ
Minimum Cathode Current For Regulation $V_{KA} = V_{ref}$ (Figure 1)	I _{min}	-	0.5	1.0	-	0.5	1.0	mA
Off-State Cathode Current (Figure 3) $V_{KA} = 36 V, V_{ref} = 0 V$	I _{off}	-	20	1000	-	20	1000	nA
Dynamic Impedance (Figure 1, Note 5) $V_{KA} = V_{ref}, \Delta I_K = 1.0 \text{ mA to } 100 \text{ mA}, f \le 1.0 \text{ kHz}$	Z _{KA}	-	0.22	0.5	-	0.22	0.5	Ω

Product parametric performance is indicated in the Electrical Characteristics for the listed test conditions, unless otherwise noted. Product performance may not be indicated by the Electrical Characteristics if operated under different conditions.

2. T_{low} = -40°C for TL431AIP TL431AILP, TL431IP, TL431ILP, TL431BID, TL431BIP, TL431BIP, TL431AIDM, TL431IDM, TL431BIDM; 0°C for TL431ACP, TL431ACLP, TL431CP, TL431CD, TL431ACD, TL431ACD, TL431BCD, TL431BCP, TL431BCLP, TL431CDM, = TL431ACDM, TL431BCDM

+85°C for TL431AIP, TL431AILP, TL431IP, TL431ILP, TL431BID, TL431BIP, TL431BILP, TL431BIDM, TL431AIDM, TL431BIDM T_{high} = = +70°C for TL431ACP, TL431ACLP, TL431CP, TL431ACD, TL431BCD, TL431BCP, TL431BCLP, TL431CDM, TL431ACDM, TL431BCDM

3. Guaranteed by design.

4. The deviation parameter ΔV_{ref} is defined as the difference between the maximum and minimum values obtained over the full operating ambient temperature range that applies.

V_{ref} max DV_{ref} = V_{ref} max -V_{ref} min $\Delta T_{A} = T_{2} - T_{1}$ V_{ref} mir T2 T1 $\frac{ppm}{{}^{\circ}C} = \frac{\left(\frac{\Delta V_{ref}}{V_{ref} @ 25{}^{\circ}C}\right) X \ 10^{6}}{\Delta T_{A}} = \frac{\Delta V_{ref} x \ 10^{6}}{\Delta T_{A} (V_{ref} @ 25{}^{\circ}C)}$ Ambient Temperature The average temperature coefficient of the reference input voltage, αV_{ref} is defined as:

aV_{ref} can be positive or negative depending on whether V_{ref} Min or V_{ref} Max occurs at the lower ambient temperature. (Refer to Figure 6.)

Example : $\Delta V_{ref} = 8.0 \text{ mV}$ and slope is positive,

$$V_{\text{ref}} @ 25^{\circ}\text{C} = 2.495 \text{ V}, \Delta T_{\text{A}} = 70^{\circ}\text{C}$$
 $\alpha \text{ V}_{\text{ref}} = \frac{0.008 \times 10^6}{70 (2.495)} = 45.8 \text{ ppm/}^{\circ}\text{C}$

5. The dynamic impedance Z_{KA} is defined as: $|Z_{KA}| = \frac{\Delta V_{KA}}{\Delta I_{K}}$. When the device is programmed with two external resistors, R1 and R2,

(refer to Figure 2) the total dynamic impedance of the circuit is defined as: $|Z_{KA}'| \approx |Z_{KA}| \left(1 + \frac{R1}{R2}\right)$

ELECTRICAL CHARACTERISTICS (T_A = 25°C, unless otherwise noted.)

		TL431AI / NCV431AI/ SCV431AI		TL431AC		TL431BC / TL431BI / TL431BV / NCV431BV					
Characteristic	Symbol	Min	Тур	Max	Min	Тур	Max	Min	Тур	Max	Unit
$ \begin{array}{l} \mbox{Reference Input Voltage (Figure 1)} \\ V_{KA} = V_{ref}, \ I_K = 10 \ mA \\ T_A = 25^\circ C \\ T_A = T_{low} \ to \ T_{high} \ (Note \ 6) \end{array} $	V _{ref}	2.47 2.44	2.495 -	2.52 2.55	2.47 2.453	2.495 -	2.52 2.537	2.485 2.475	2.495 2.495	2.505 2.515	V
Reference Input Voltage Deviation Over Temperature Range (Figure 1, Notes 7, 8) V_{KA} = V_{ref} , I_K = 10 mA	ΔV_{ref}	-	7.0	30	_	3.0	17	-	3.0	17	mV
$\begin{array}{l} \mbox{Ratio of Change in Reference Input Voltage to} \\ \mbox{Change in Cathode to Anode Voltage} \\ \mbox{I}_{K} = 10 \mbox{ mA (Figure 2),} \\ \mbox{$\Delta V_{KA} = 10 \mbox{ V to } V_{ref}$} \\ \mbox{$\Delta V_{KA} = 36 \mbox{ V to } 10 \mbox{ V} $} \end{array}$	$\frac{\Delta V_{ref}}{\Delta V_{KA}}$		-1.4 -1.0	-2.7 -2.0		-1.4 -1.0	-2.7 -2.0		-1.4 -1.0	-2.7 -2.0	mV/V
$ \begin{array}{l} \mbox{Reference Input Current (Figure 2)} \\ I_K = 10 \mbox{ mA}, \mbox{ R1} = 10 \mbox{ k}, \mbox{ R2} = \infty \\ T_A = 25^\circ \mbox{C} \\ T_A = T_{low} \mbox{ to } T_{high} \mbox{ (Note 6)} \end{array} $	I _{ref}		1.8 _	4.0 6.5		1.8 _	4.0 5.2		1.1 -	2.0 4.0	μΑ
Reference Input Current Deviation Over Temperature Range (Figure 2, Note 7) $I_{K} = 10 \text{ mA}, \text{ R1} = 10 \text{ k}, \text{ R2} = \infty$	ΔI_{ref}	-	0.8	2.5	_	0.4	1.2	-	0.8	2.5	μA
Minimum Cathode Current For Regulation $V_{KA} = V_{ref}$ (Figure 1)	I _{min}	-	0.5	1.0	-	0.5	1.0	-	0.5	1.0	mA
Off–State Cathode Current (Figure 3) $V_{KA} = 36 V, V_{ref} = 0 V$	l _{off}	-	20	1000	_	20	1000	-	0.23	500	nA
Dynamic Impedance (Figure 1, Note 9) $ V_{KA} = V_{ref}, \Delta I_K = 1.0 \text{ mA to } 100 \text{ mA} \\ f \leq 1.0 \text{ kHz} $	Z _{KA}	-	0.22	0.5	_	0.22	0.5	-	0.14	0.3	Ω

Product parametric performance is indicated in the Electrical Characteristics for the listed test conditions, unless otherwise noted. Product performance may not be indicated by the Electrical Characteristics if operated under different conditions.

6. T_{low} = -40°C for TL431AIP TL431AILP, TL431IP, TL431ILP, TL431BID, TL431BIP, TL431BILP, TL431BILP, TL431BV, TL431AIDM, TL431IDM,

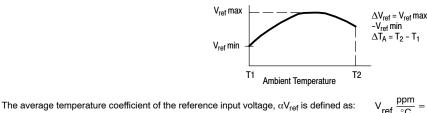
TL431BIDM, NCV431AIDMR2G, NCV431AIDR2G, NCV431BVDR2G, SCV431AIDMR2G = 0°C for TL431ACP, TL431ACLP, TL431CP, TL431CLP, TL431CD, TL431ACD, TL431BCD, TL431BCP, TL431BCLP, TL431CDM, TL431ACDM, TL431BCDM, SCV431AIDMR2G

T_{high} = +85°C for TL431AIP, TL431AIP, TL431IP, TL431IP, TL431BID, TL431BIP, TL431BIP, TL431BIDM, TL431AIDM, TL431AIDM, TL431BIDM = +70°C for TL431ACP, TL431ACLP, TL431CP, TL431ACD, TL431BCD, TL431BCP, TL431BCLP, TL431CDM, TL431ACDM, TL431ACDM, TL431BCDM

= +125°C TL431BV, NCV431AIDMR2G, NCV431AIDR2G, NCV431BVDMR2G, NCV431BVDR2G, SCV431AIDMR2G

7. Guaranteed by design.

 The deviation parameter ΔV_{ref} is defined as the difference between the maximum and minimum values obtained over the full operating ambient temperature range that applies.



$$\frac{\left(\frac{\Delta V_{\text{ref}}}{V_{\text{ref}} @ 25^{\circ}\text{C}}\right) \times 10^{6}}{\Delta T_{\text{A}}} = \frac{\Delta V_{\text{ref}} \times 10^{6}}{\Delta T_{\text{A}} (V_{\text{ref}} @ 25^{\circ}\text{C})}$$

 αV_{ref} can be positive or negative depending on whether V_{ref} Min or V_{ref} Max occurs at the lower ambient temperature. (Refer to Figure 6.)

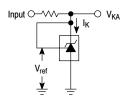
Example : $\Delta V_{ref} = 8.0 \text{ mV}$ and slope is positive, $V_{ref} @ 25^{\circ}C = 2.495 \text{ V}, \Delta T_{A} = 70^{\circ}C$

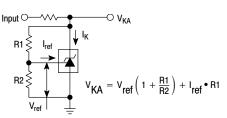
$$\alpha V_{ref} = \frac{0.008 \times 10^6}{70 (2.495)} = 45.8 \text{ ppm/}^{\circ}\text{C}$$

9. The dynamic impedance Z_{KA} is defined as $|Z_{KA}| = \frac{\Delta V_{KA}}{\Delta I_{K}}$ When the device is programmed with two external resistors, R1 and R2, (refer

to Figure 2) the total dynamic impedance of the circuit is defined as: $|Z_{KA}'| \approx |Z_{KA}| \left(1 + \frac{R_1}{R_2}\right)$

10. NCV431AIDMR2G, NCV431AIDR2G, NCV431BVDMR2G, NCV431BVDR2G, SCV431AIDMR2G T_{low} = -40°C, T_{high} = +125°C. NCV prefix is for automotive and other applications requiring unique site and control change requirements.





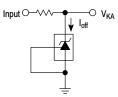
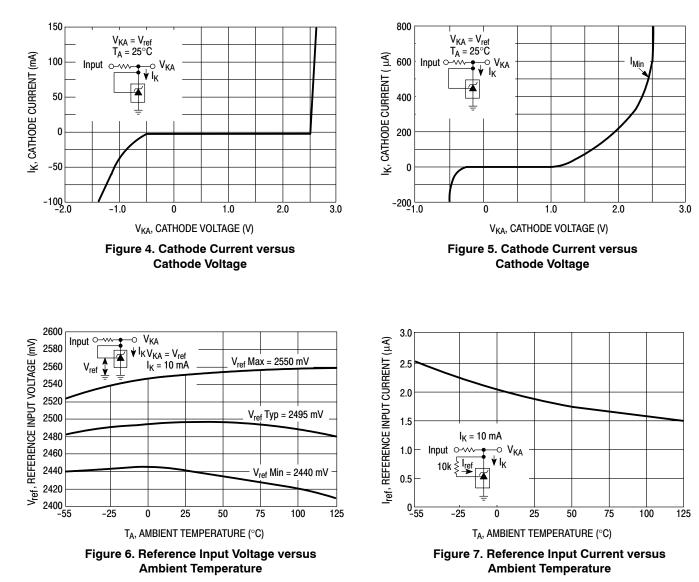
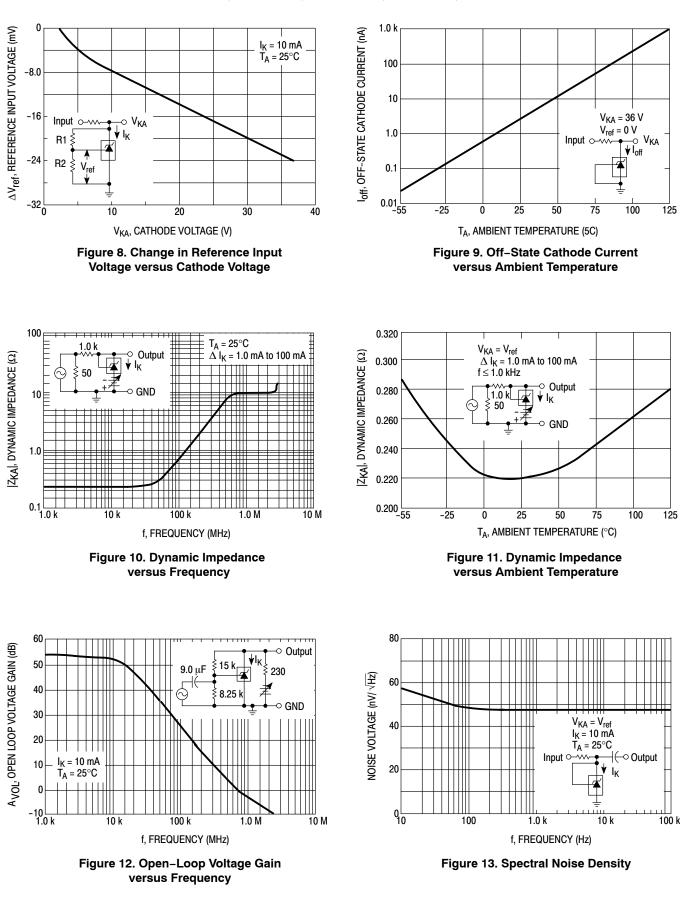


Figure 1. Test Circuit for V_{KA} = V_{ref}

Figure 2. Test Circuit for V_{KA} > V_{ref}

Figure 3. Test Circuit for Ioff





TL431A, B Series, NCV431A, B Series, SCV431A

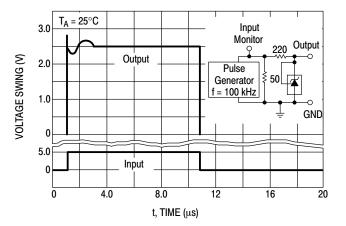


Figure 14. Pulse Response

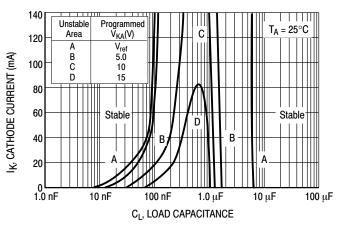


Figure 15. Stability Boundary Conditions

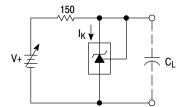


Figure 16. Test Circuit For Curve A of Stability Boundary Conditions

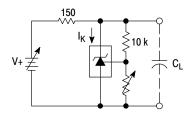


Figure 17. Test Circuit For Curves B, C, And D of Stability Boundary Conditions

TYPICAL APPLICATIONS

TL431A, B Series, NCV431A, B Series, SCV431A

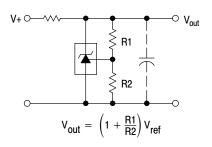


Figure 18. Shunt Regulator

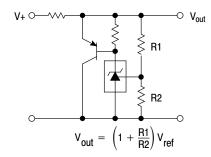


Figure 19. High Current Shunt Regulator

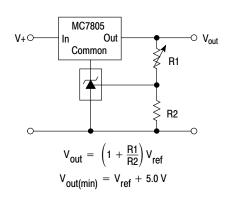


Figure 20. Output Control for a Three–Terminal Fixed Regulator

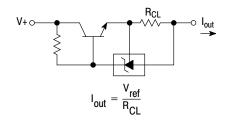


Figure 22. Constant Current Source

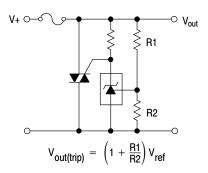


Figure 24. TRIAC Crowbar

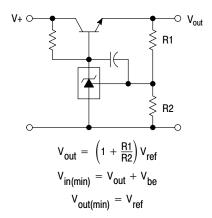


Figure 21. Series Pass Regulator

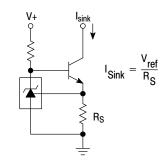


Figure 23. Constant Current Sink

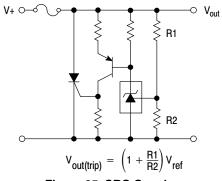
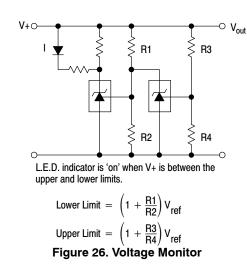


Figure 25. SRC Crowbar



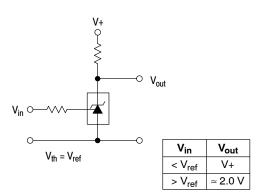


Figure 27. Single–Supply Comparator with Temperature–Compensated Threshold

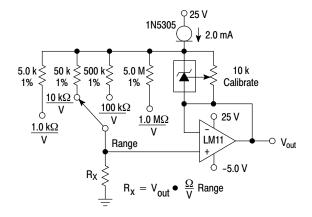


Figure 28. Linear Ohmmeter

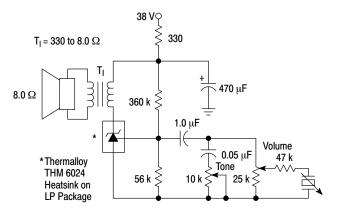


Figure 29. Simple 400 mW Phono Amplifier

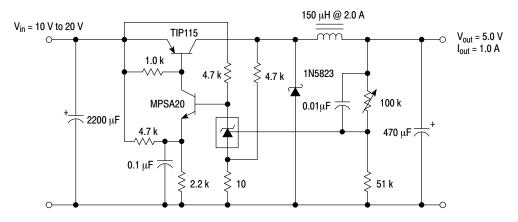


Figure 30. High Efficiency Step-Down Switching Converter

Test	Conditions	Results
Line Regulation	V_{in} = 10 V to 20 V, I_o = 1.0 A	53 mV (1.1%)
Load Regulation	V_{in} = 15 V, I_{o} = 0 A to 1.0 A	25 mV (0.5%)
Output Ripple	V _{in} = 10 V, I _o = 1.0 A	50 mVpp P.A.R.D.
Output Ripple	$V_{in} = 20 \text{ V}, I_o = 1.0 \text{ A}$	100 mVpp P.A.R.D.
Efficiency	V _{in} = 15 V, I _o = 1.0 A	82%

APPLICATIONS INFORMATION

The TL431 is a programmable precision reference which is used in a variety of ways. It serves as a reference voltage in circuits where a non-standard reference voltage is needed. Other uses include feedback control for driving an optocoupler in power supplies, voltage monitor, constant current source, constant current sink and series pass regulator. In each of these applications, it is critical to maintain stability of the device at various operating currents and load capacitances. In some cases the circuit designer can estimate the stabilization capacitance from the stability boundary conditions curve provided in Figure 15. However, these typical curves only provide stability information at specific cathode voltages and at a specific load condition. Additional information is needed to determine the capacitance needed to optimize phase margin or allow for process variation.

A simplified model of the TL431 is shown in Figure 31. When tested for stability boundaries, the load resistance is 150 Ω . The model reference input consists of an input transistor and a dc emitter resistance connected to the device anode. A dependent current source, Gm, develops a current whose amplitude is determined by the difference between the 1.78 V internal reference voltage source and the input transistor emitter voltage. A portion of Gm flows through compensation capacitance, C_{P2}. The voltage across C_{P2} drives the output dependent current source, Go, which is connected across the device cathode and anode.

Model component values are:

 $V_{ref} = 1.78 V$

 $Gm = 0.3 + 2.7 \exp(-I_C/26 mA)$

where I_C is the device cathode current and Gm is in mhos

Go = 1.25 (V_{cp}2) µmhos.

Resistor and capacitor typical values are shown on the model. Process tolerances are $\pm 20\%$ for resistors, $\pm 10\%$ for capacitors, and $\pm 40\%$ for transconductances.

An examination of the device model reveals the location of circuit poles and zeroes:

P1 =
$$\frac{1}{2\pi R_{GM} C_{P1}} = \frac{1}{2\pi * 1.0 M * 20 pF} = 7.96 \text{ kHz}$$

P2 =
$$\frac{1}{2\pi R_{P2}C_{P2}} = \frac{1}{2\pi * 10 M * 0.265 pF} = 60 \text{ kHz}$$

Z1 = $\frac{1}{2\pi R_{Z1}C_{P1}} = \frac{1}{2\pi * 15.9 \text{ k} * 20 pF} = 500 \text{ kHz}$

In addition, there is an external circuit pole defined by the load:

$$\mathsf{P}_{\mathsf{L}} = \frac{1}{2\pi \,\mathsf{R}_{\mathsf{L}}\mathsf{C}_{\mathsf{L}}}$$

Also, the transfer dc voltage gain of the TL431 is:

$$G = G_M R_{GM} G_0 R_L$$

Example 1:

 $\rm I_{C}$ = 10 mA, $\rm R_{L}$ = 230 $\Omega, \rm C_{L}$ = 0. Define the transfer gain .

The DC gain is:

(2

$$G = G_M R_{GM} G_0 R_L =$$

.138)(1.0 M)(1.25 μ)(230) = 615 = 56 dB

Loop gain = G
$$\frac{8.25 \text{ k}}{8.25 \text{ k} + 15 \text{ k}}$$
 = 218 = 47 dB

The resulting transfer function Bode plot is shown in Figure 32. The asymptotic plot may be expressed as the following equation:

$$Av = 615 \frac{\left(1 + \frac{jf}{500 \text{ kHz}}\right)}{\left(1 + \frac{jf}{8.0 \text{ kHz}}\right)\left(1 + \frac{jf}{60 \text{ kHz}}\right)}$$

The Bode plot shows a unity gain crossover frequency of approximately 600 kHz. The phase margin, calculated from the equation, would be 55.9 degrees. This model matches the Open–Loop Bode Plot of Figure 12. The total loop would have a unity gain frequency of about 300 kHz with a phase margin of about 44 degrees.

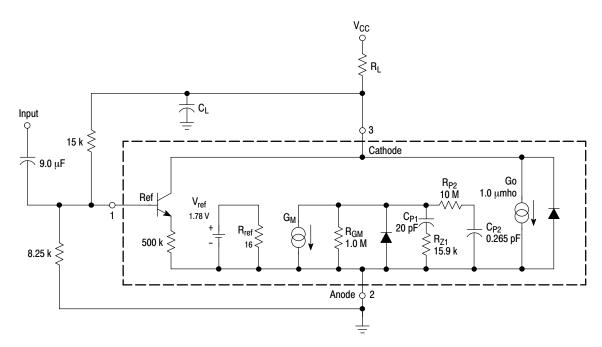


Figure 31. Simplified TL431 Device Model

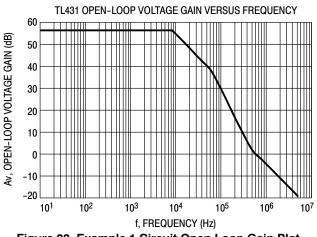


Figure 32. Example 1 Circuit Open Loop Gain Plot Example 2.

 $I_C = 7.5 \text{ mA}$, $R_L = 2.2 \text{ k}\Omega$, $C_L = 0.01 \mu\text{F}$. Cathode tied to reference input pin. An examination of the data sheet stability boundary curve (Figure 15) shows that this value of load capacitance and cathode current is on the boundary. Define the transfer gain.

The DC gain is:

$$G = G_M R_{GM} G_0 R_L =$$

 $(2.323)(1.0 \text{ M})(1.25 \mu)(2200) = 6389 = 76 \text{ dB}$

The resulting open loop Bode plot is shown in Figure 33. The asymptotic plot may be expressed as the following equation:

$$Av = 615 \frac{\left(1 + \frac{jf}{500 \text{ kHz}}\right)}{\left(1 + \frac{jf}{8.0 \text{ kHz}}\right)\left(1 + \frac{jf}{60 \text{ kHz}}\right)\left(1 + \frac{jf}{7.2 \text{ kHz}}\right)}$$

Note that the transfer function now has an extra pole formed by the load capacitance and load resistance.

Note that the crossover frequency in this case is about 250 kHz, having a phase margin of about -46 degrees. Therefore, instability of this circuit is likely.

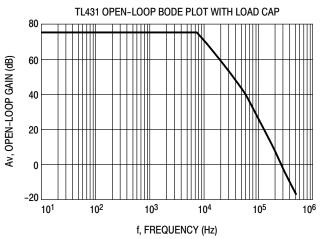


Figure 33. Example 2 Circuit Open Loop Gain Plot

With three poles, this system is unstable. The only hope for stabilizing this circuit is to add a zero. However, that can only be done by adding a series resistance to the output capacitance, which will reduce its effectiveness as a noise filter. Therefore, practically, in reference voltage applications, the best solution appears to be to use a smaller value of capacitance in low noise applications or a very large value to provide noise filtering and a dominant pole rolloff of the system.

ORDERING INFORMATION

Device	Marking Code	Operating Temperature Range	Package Code	Shipping Information [†]	Tolerance
TL431ACDG	AC				1.0%
TL431BCDG	BC			98 Units / Rail	0.4%
TL431CDG	С		SOIC-8		2.2%
TL431ACDR2G	AC		(Pb-Free)		1.0%
TL431BCDR2G	BC			2500 / Tape & Reel	0.4%
TL431CDR2G	С				2.2%
TL431ACDMR2G	TAC				1.0%
TL431BCDMR2G	TBC		Micro8 (Pb-Free)	4000 / Tape & Reel	0.4%
TL431CDMR2G	T–C		(FD-FIEE)		2.2%
TL431ACPG	ACP				1.0%
TL431BCPG	BCP		PDIP-8 (Pb-Free)	50 Units / Rail	0.4%
TL431CPG	СР		(PD-Free)		2.2%
TL431ACLPG	ACLP	0°C to 70°C			1.0%
TL431BCLPG	BCLP			2000 Units / Bag	0.4%
TL431CLPG	CLP				2.2%
TL431ACLPRAG	ACLP				1.0%
TL431BCLPRAG	BCLP				0.4%
TL431CLPRAG	CLP			2000 / Tape & Reel	2.2%
TL431ACLPREG	ACLP		TO-92		1.0%
TL431BCLPREG	BCLP		(Pb-Free)		0.4%
TL431CLPREG	CLP				2.2%
TL431ACLPRPG	ACLP			2000 / Tape & Ammo Box	1.0%
TL431BCLPRMG	BCLP			· · · ·	0.4%
TL431CLPRMG	CLP			2000 / Fan-Fold	
TL431CLPRPG	CLP				2.2%
TL431AIDG	AI		SOIC-8 (Pb-Free)	98 Units / Rail 2500s / Tape & Reel	1.0%
TL431BIDG	BI				0.4%
TL431IDG	I				2.2%
TL431AIDR2G	AI				1.0%
TL431BIDR2G	BI				0.4%
TL431IDR2G	I				2.2%
TL431AIDMR2G	TAI				1.0%
TL431BIDMR2G	TBI		Micro8 (Pb-Free)	4000 / Tape & Reel	0.4%
TL431IDMR2G	T–I		(FD-FIEE)		2.2%
TL431AIPG	AIP				1.0%
TL431BIPG	BIP		PDIP-8	50 Units / Rail	0.4%
TL431IPG	IP	–40°C to 85°C	(Pb–Free)		2.2%
TL431AILPG	AILP				1.0%
TL431BILPG	BILP			2000 Units / Bag	0.4%
TL431ILPG	ILP				2.2%
TL431AILPRAG	AILP				1.0%
TL431BILPRAG	BILP		TO-92		0.4%
SC431ILPRAG	ILP		(Pb-Free)	2000 / Tape & Reel	0.001
TL431ILPRAG	ILP				2.2%
TL431AILPRMG	A			2000 / Tape & Ammo Box	
TL431AILPRPG	AILP				1.0%
TL431ILPRPG	ILP				2.2%

+For information on tape and reel specifications, including part orientation and tape sizes, please refer to our Tape and Reel Packaging Specifications Brochure, BRD8011/D.

*NCV/SCV Prefixes for Automotive and Other Applications Requiring Unique Site and Control Change Requirements; AEC–Q100 Qualified and PPAP Capable.

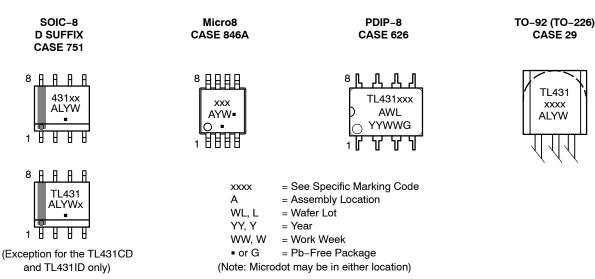
ORDERING INFORMATION

Device	Marking Code	Operating Temperature Range	Package Code	Shipping Information [†]	Tolerance
TL431BVDG	BV		SOIC-8	98 Units / Rail	
TL431BVDR2G	DV		(Pb-Free)	2500 / Tape & Reel	
TL431BVDMR2G	TBV		Micro8 (Pb–Free)	4000 / Tape & Reel	0.4%
TL431BVLPG	BVLP		TO-92	2000 Units / Bag	
TL431BVLPRAG	DVLP		(Pb-Free)	2000 / Tape & Reel	
TL431BVPG	BVP	−40°C to 125°C	PDIP-8 (Pb-Free)	50 Units / Rail	0.4%
NCV431AIDMR2G*	RAN	-40 0 10 125 0	Micro8	4000 / Tana & Daal	
SCV431AIDMR2G*	RAP		(Pb-Free)	4000 / Tape & Reel	1%
NCV431AIDR2G*	AV		SOIC-8 (Pb-Free)	2500 / Tape & Reel	1/0
NCV431BVDMR2G*	NVB		Micro8 (Pb–Free)	4000 / Tape & Reel	0.4%
NCV431BVDR2G*	BV		SOIC-8 (Pb-Free)	2500 / Tape & Reel	0.4%

+For information on tape and reel specifications, including part orientation and tape sizes, please refer to our Tape and Reel Packaging Specifications Brochure, BRD8011/D.

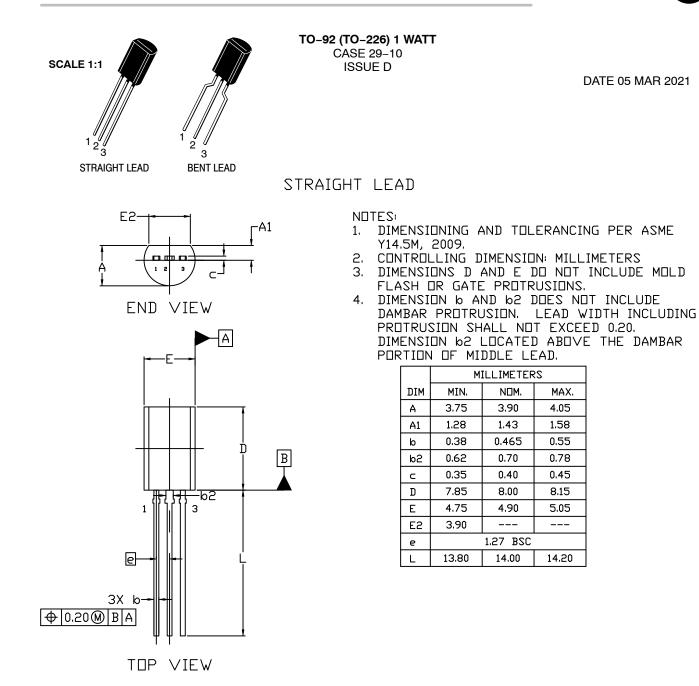
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MARKING DIAGRAMS



MECHANICAL CASE OUTLINE PACKAGE DIMENSIONS





STYLES AND MARKING ON PAGE 3

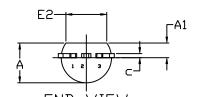
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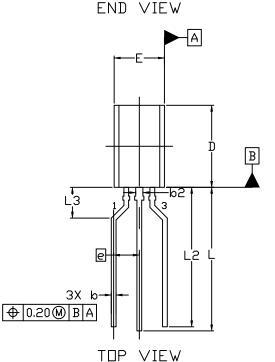


TO-92 (TO-226) 1 WATT CASE 29–10 ISSUE D

DATE 05 MAR 2021

FORMED LEAD





NDTES:

- 1. DIMENSIONING AND TOLERANCING PER ASME Y14.5M, 2009.
- 2. CONTROLLING DIMENSION: MILLIMETERS
- 3. DIMENSIONS D AND E DO NOT INCLUDE MOLD FLASH OR GATE PROTRUSIONS,
- 4. DIMENSION ७ AND ७2 DOES NOT INCLUDE DAMBAR PROTRUSION. LEAD WIDTH INCLUDING PROTRUSION SHALL NOT EXCEED 0.20. DIMENSION ७2 LOCATED ABOVE THE DAMBAR PORTION OF MIDDLE LEAD.

	M	ILLIMETER	22
DIM	MIN.	NDM.	MAX.
Α	3.75	3.90	4.05
A1	1.28	1.43	1.58
σ	0.38	0.465	0.55
b2	0.62	0.70	0.78
с	0.35	0.40	0.45
D	7.85	8.00	8.15
Е	4.75	4.90	5.05
E2	3.90		
e	2.50 BSC		
L	13.80	14.00	14.20
L2	13.20	13.60	14.00
L3		3.00 REF	

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DATE 05 MAR 2021

2.	EMITTER BASE COLLECTOR
STYLE 6: PIN 1. 2. 3.	SOURCE & SUBSTRATE
2.	ANODE CATHODE & ANODE CATHODE
2.	ANODE GATE CATHODE
2.	COLLECTOR EMITTER BASE
	V _{CC} GROUND 2 OUTPUT
	GATE DRAIN SOURCE

STYLE 2: PIN 1. BASE 2. EMITTER 3. COLLECTOR STYLE 7: PIN 1. SOURCE 2. DRAIN 3. GATE STYLE 12: PIN 1. MAIN TERMINAL 1 2. GATE 3. MAIN TERMINAL 2 STYLE 17: PIN 1. COLLECTOR 2. BASE 3. EMITTER STYLE 22: PIN 1. SOURCE 2. GATE 3. DRAIN STYLE 27: PIN 1. MT 2. SUBSTRATE 3. MT STYLE 32 PIN 1. BASE 2. COLLECTOR 3. EMITTER

2	: ANODE ANODE CATHODE
2	: DRAIN GATE SOURCE & SUBSTRATE
2	3: ANODE 1 GATE CATHODE 2
2	8: ANODE CATHODE NOT CONNECTED
2	3: GATE SOURCE DRAIN
2	8: . CATHODE . ANODE . GATE
2	3: . Return . INPUT . Output

STYLE 4: PIN 1. CATHODE STYLE 5: 2. CATHODE 3. ANODE STYLE 9: PIN 1. BASE 1 EMITTER 2. 3. BASE 2 STYLE 14: PIN 1. EMITTER 2. COLLECTOR 3. BASE STYLE 19: PIN 1. GATE 2. ANODE 3. CATHODE STYLE 24: PIN 1. EMITTER 2. COLLECTOR/ANODE 3. CATHODE STYLE 29: PIN 1. NOT CONNECTED 2. ANODE 3. CATHODE STYLE 34: PIN 1. INPUT 2. GROUND

3. LOGIC

PIN 1. DRAIN 2. SOURCE 3. GATE STYLE 10: PIN 1. CATHODE 2. GATE 3. ANODE STYLE 15: PIN 1. ANODE 1 2. CATHODE 3. ANODE 2 STYLE 20: PIN 1. NOT CONNECTED 2. CATHODE 3. ANODE STYLE 25: PIN 1. MT 1 2. GATE 3. MT 2 STYLE 30: PIN 1. DRAIN 2. GATE 3. SOURCE STYLE 35: PIN 1. GATE 2. COLLECTOR 3. EMITTER

GENERIC MARKING DIAGRAM*

XXXXX XXXXX ALYW

XXXX = Specific Device Code

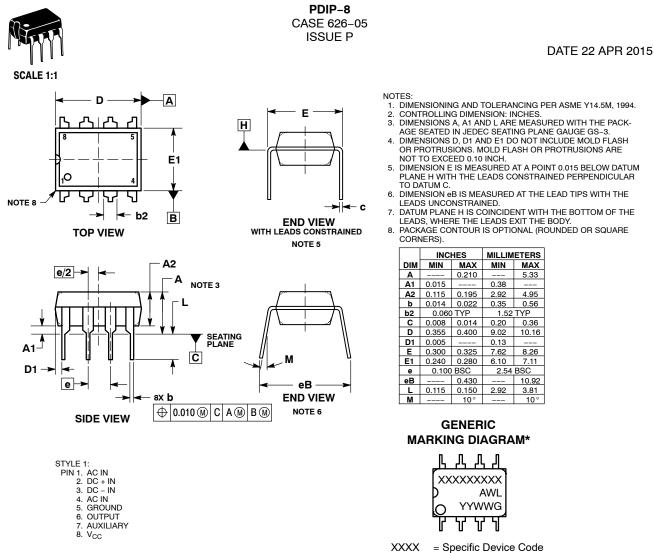
- A = Assembly Location
- L = Wafer Lot
- Y = Year
- W = Work Week
 - = Pb-Free Package

(Note: Microdot may be in either location)

*This information is generic. Please refer to device data sheet for actual part marking. Pb-Free indicator, "G" or microdot "•", may or may not be present. Some products may not follow the Generic Marking.

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A = Assembly Location

- WL = Wafer Lot
- YY = Year
- WW = Work Week
- G = Pb-Free Package

*This information is generic. Please refer to device data sheet for actual part marking. Pb–Free indicator, "G" or microdot " ■", may or may not be present.







*For additional information on our Pb–Free strategy and soldering details, please download the ON Semiconductor Soldering and Mounting Techniques Reference Manual, SOLDERRM/D.

STYLES ON PAGE 2

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STYLE 1: PIN 1. EMITTER COLLECTOR 2. COLLECTOR 3. 4. EMITTER EMITTER 5. BASE 6. 7 BASE EMITTER 8. STYLE 5: PIN 1. DRAIN 2. DRAIN 3. DRAIN DRAIN 4. GATE 5. 6. GATE SOURCE 7. 8. SOURCE STYLE 9: PIN 1. EMITTER, COMMON COLLECTOR, DIE #1 COLLECTOR, DIE #2 2. З. EMITTER, COMMON 4. 5. EMITTER, COMMON 6 BASE. DIE #2 BASE, DIE #1 7. 8. EMITTER, COMMON STYLE 13: PIN 1. N.C. 2. SOURCE 3 GATE 4. 5. DRAIN 6. DRAIN DRAIN 7. DRAIN 8. STYLE 17: PIN 1. VCC 2. V2OUT V10UT З. TXE 4. 5. RXE 6. VFF 7. GND 8. ACC STYLE 21: CATHODE 1 PIN 1. 2. CATHODE 2 3 CATHODE 3 CATHODE 4 4. 5. CATHODE 5 6. COMMON ANODE COMMON ANODE 7. 8. CATHODE 6 STYLE 25: PIN 1. VIN 2 N/C REXT З. 4. GND 5. IOUT 6. IOUT IOUT 7. 8. IOUT STYLE 29: BASE, DIE #1 PIN 1. 2 EMITTER, #1 BASE, #2 З. EMITTER, #2 4. 5 COLLECTOR, #2 COLLECTOR, #2 6.

STYLE 2: PIN 1. COLLECTOR, DIE, #1 2. COLLECTOR, #1 COLLECTOR, #2 3. 4 COLLECTOR, #2 BASE, #2 5. EMITTER, #2 6. 7 BASE #1 EMITTER, #1 8. STYLE 6: PIN 1. SOURCE 2. DRAIN 3. DRAIN SOURCE 4. SOURCE 5. 6. GATE GATE 7. 8. SOURCE STYLE 10: PIN 1. GROUND BIAS 1 OUTPUT 2. З. GROUND 4. 5. GROUND 6 BIAS 2 INPUT 7. 8. GROUND STYLE 14: PIN 1. N-SOURCE 2. N-GATE P-SOURCE 3 P-GATE 4. P-DRAIN 5 6. P-DRAIN N-DRAIN 7. N-DRAIN 8. STYLE 18: PIN 1. ANODE 2. ANODE SOURCE 3. GATE 4. 5. DRAIN 6 DRAIN CATHODE 7. CATHODE 8. STYLE 22 PIN 1. I/O LINE 1 2. COMMON CATHODE/VCC COMMON CATHODE/VCC 3 4. I/O LINE 3 5. COMMON ANODE/GND 6. I/O LINE 4 7. I/O LINE 5 COMMON ANODE/GND 8. STYLE 26: PIN 1. GND 2 dv/dt З. ENABLE 4. ILIMIT 5. SOURCE SOURCE 6. SOURCE 7. 8. VCC STYLE 30: DRAIN 1 PIN 1. DRAIN 1 2 GATE 2 З. SOURCE 2 4. SOURCE 1/DRAIN 2 SOURCE 1/DRAIN 2 5. 6.

STYLE 3: PIN 1. DRAIN, DIE #1 DRAIN, #1 2. DRAIN, #2 З. 4. DRAIN, #2 GATE, #2 5. SOURCE, #2 6. 7 GATE #1 8. SOURCE, #1 STYLE 7: PIN 1. INPUT 2. EXTERNAL BYPASS THIRD STAGE SOURCE GROUND З. 4. 5. DRAIN 6. GATE 3 SECOND STAGE Vd 7. FIRST STAGE Vd 8. STYLE 11: PIN 1. SOURCE 1 GATE 1 SOURCE 2 2. 3. GATE 2 4. 5. DRAIN 2 6. DRAIN 2 DRAIN 1 7. 8. DRAIN 1 STYLE 15: PIN 1. ANODE 1 2. ANODE 1 ANODE 1 3 ANODE 1 4. 5. CATHODE, COMMON CATHODE, COMMON CATHODE, COMMON 6. 7. CATHODE, COMMON 8. STYLE 19: PIN 1. SOURCE 1 GATE 1 SOURCE 2 2. 3. GATE 2 4. 5. DRAIN 2 6. MIRROR 2 DRAIN 1 7. 8. **MIRROR 1** STYLE 23: PIN 1. LINE 1 IN COMMON ANODE/GND COMMON ANODE/GND 2. 3 LINE 2 IN 4. LINE 2 OUT 5. COMMON ANODE/GND COMMON ANODE/GND 6. 7. LINE 1 OUT 8. STYLE 27: PIN 1. ILIMIT 2 OVI 0 UVLO З. 4. INPUT+ 5. SOURCE SOURCE 6. SOURCE 7. 8 DRAIN

DATE 16 FEB 2011

STYLE 4: ANODE ANODE PIN 1. 2. ANODE З. 4. ANODE ANODE 5. 6. ANODE 7 ANODE COMMON CATHODE 8. STYLE 8: PIN 1. COLLECTOR, DIE #1 2. BASE, #1 BASE #2 3. COLLECTOR, #2 4. COLLECTOR, #2 5. 6. EMITTER, #2 EMITTER, #1 7. 8. COLLECTOR, #1 STYLE 12: PIN 1. SOURCE SOURCE 2. 3. 4. GATE 5. DRAIN 6. DRAIN DRAIN 7. 8. DRAIN STYLE 16: PIN 1. EMITTER, DIE #1 2. BASE, DIE #1 EMITTER, DIE #2 3 BASE, DIE #2 4. 5. COLLECTOR, DIE #2 6. COLLECTOR, DIE #2 COLLECTOR, DIE #1 7. COLLECTOR, DIE #1 8. STYLE 20: PIN 1. SOURCE (N) GATE (N) SOURCE (P) 2. 3. 4. GATE (P) 5. DRAIN 6. DRAIN DRAIN 7. 8. DRAIN STYLE 24: PIN 1. BASE 2. EMITTER 3 COLLECTOR/ANODE COLLECTOR/ANODE 4. 5. CATHODE 6. CATHODE COLLECTOR/ANODE 7. 8. COLLECTOR/ANODE STYLE 28: PIN 1. SW_TO_GND 2. DASIC OFF DASIC_SW_DET З. 4. GND 5. 6. V MON VBULK 7. VBULK 8 VIN

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SOURCE 1/DRAIN 2

7.

8. GATE 1

7.

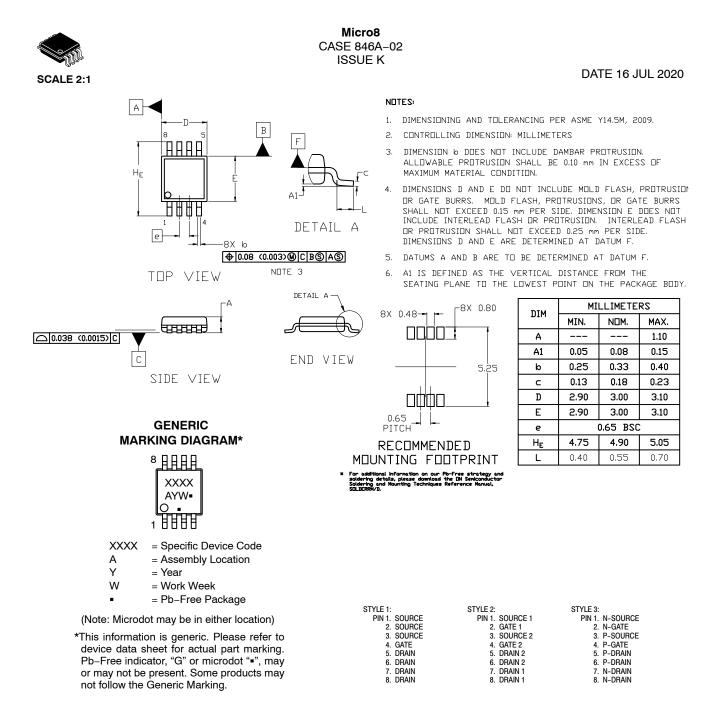
8

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COLLECTOR, #1

COLLECTOR, #1





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