Highest Efficiency Supply for 2 to 6 Series WLEDs in a 2mm x 2mm TDFN Package

Absolute Maximum Ratings

IN to GND	0.3V to +7V
CS, COMP, ON to GND	0.3V to $(V_{IN} + 0.3V)$
OVP, LX to GND	0.3V to +28V
PGND to GND	0.3V to +0.3V
LX Current	770mA _{RMS}
Continuous Power Dissipation (multilay	er board at +70°C)
8-pin, 2mm x 2mm TDFN	
(derate above +70°C by 11 9mW/°C	953mW

Operating Temperature Range	40°C to +85°C
Junction Temperature	+150°C
Storage Temperature Range	65°C to +150°C
Lead Temperature (soldering, 10s)	+300°C
Soldering Temperature (reflow)	

Stresses beyond those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

Electrical Characteristics

 $(V_{IN} = V_{ON} = V_{OVP} = 3.6V, V_{PGND} = V_{GND} = 0V, COMP, CS, and LX are unconnected, T_A = -40°C to +85°C, unless otherwise noted. Typical values are at T_A = +25°C.) (Note 1)$

PARAMETER	SYMBOL	CONDITIONS		MIN	TYP	MAX	UNITS
Input Voltage	V _{IN}					5.5	V
Input Undervoltage Lockout Threshold	V _{IN_UVLO}	V _{IN} rising		2.25		2.55	V
THESHOLD		V _{IN} falling				2.53	
Input Overvoltage Lockout	VINI 01/10	V _{IN} rising		6.2		6.35	V
Threshold	V _{IN} _OVLO	V _{IN} falling		6.0		6.3	
Shutdown Input Current	ISHDN	V _{ON} = 0V	$T_A = +25^{\circ}C$		0.01	1	μΑ
Shataown input Current	ISHUN	VOW - OV	T _A = +85°C		0.1		
Quiescent Current	IQ	$V_{CS} = 0.55V$, no load	MAX8901A		70	135	μА
Quiescent Current 	iQ	(not switching)	MAX8901B		115	185	
Output Voltage Range	V _{OUT}	V _{DIODE} = external boost diode voltage drop		V _{IN} - V _{DIODE}		V_{OV}	V
OVP Overvoltage Protection Threshold	Vov	6 LEDs with 25V OVP option		24	25	26	V
OVP Input Current	lov	V _{OVP} = 20V			20		μΑ
CS Regulation Voltage	Vcs	No dimming		0.475	0.50	0.525	V
ON Shutdown Delay	tshdn	Time V _{ON} is below low threshold until shutdown (Figure 1)		1.18	1.33	1.50	ms
ON High Voltage	Von_hi	2.6V < V _{IN} < 5.5V		1.3			V
ON Low Voltage	V _{ON_LO}	2.6V < V _{IN} < 5.5V				0.4	V
ON Input Current	lov	ON = IN	$T_A = +25^{\circ}C$		0.01	1	μΑ
	I _{ON}		$T_A = +85^{\circ}C$		0.1		
Initial ON High Pulse Width	thi_init	First ON high pulse to enable IC (MAX8901B) (Figure 1)		40			μs
ON High Pulse Width	tHI	MAX8901B (Figure 1)		0.5			μs
ON Low Pulse Width	tLO	MAX8901B (Figure 1)		0.5		500.0	μs

Electrical Characteristics (continued)

 $(V_{IN} = V_{ON} = V_{OVP} = 3.6V, V_{PGND} = V_{GND} = 0V, COMP, CS, and LX are unconnected, TA = -40°C to +85°C, unless otherwise noted. Typical values are at TA = +25°C.) (Note 1)$

PARAMETER	SYMBOL	CONDITIONS		MIN	TYP	MAX	UNITS
Serial Dimming		I _{LED} = 24.75mA	$T_A = +25^{\circ}C$	-1.3		+1.3	%
Full-Scale LED Current Accuracy		(MAX8901B)	$T_A = -40^{\circ}C \text{ to } +85^{\circ}C$	-3		+3	%
PWM Frequency for PWM Dimming Control	fpwm	MAX8901A			100		kHz
CS to COMP Transconductance				40	60	80	μS
Soft-Start Interval	tsoft- Start	C _{COMP} = 0.022μF (Figure 1)			10		ms
Thermal Shutdown					160		°C
OSCILLATOR							
Operating Frequency	fsw			700	750	800	kHz
Maximum Duty Cycle	DMAX	V _{CS} = 0.4V		90	92		%
n-CHANNEL SWITCH							
LX Leakage Current		V _L X = 27V	T _A = +25°C		0.1	5	
	ILXLKG		$T_A = +85^{\circ}C$		0.1	•	μA
n-Channel Switch On-Resistance	R _L X			·	0.7	1.4	Ω
n-Channel Current Limit	I _{LIM}		_	0.63	0.70	0.77	Α

Note 1: Specifications to -40°C are guaranteed by design and characterization and are not production tested.

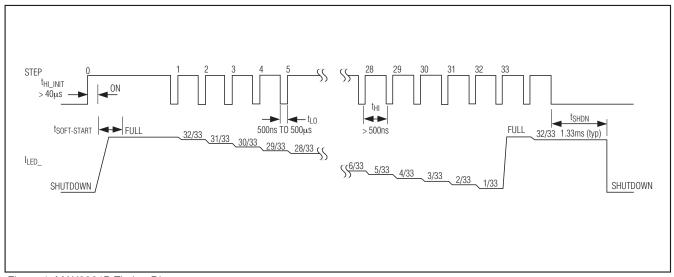
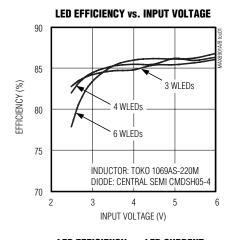
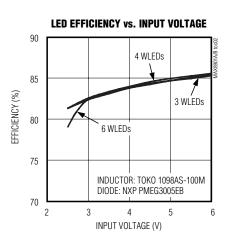


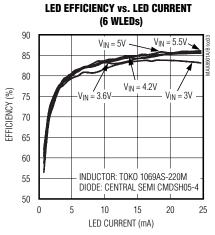
Figure 1. MAX8901B Timing Diagram

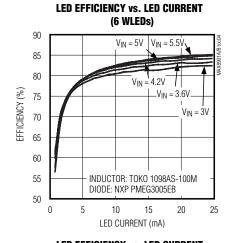
Typical Operating Characteristics

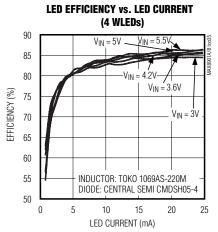
(Circuit of Figure 5, $I_{LED} = 24.75$ mA, $V_{IN} = V_{ON} = 3.6$ V, C1 = 1µF, C2 = 1µF, $T_A = +25$ °C, unless otherwise noted.)

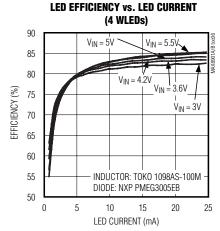






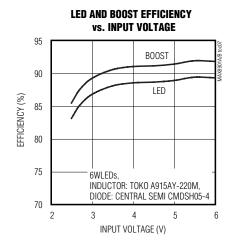


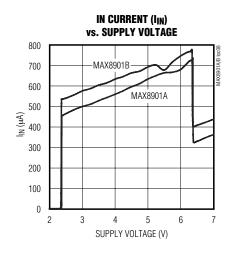


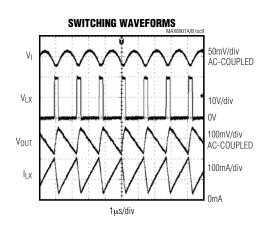


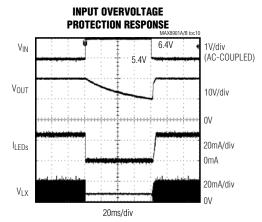
Typical Operating Characteristics (continued)

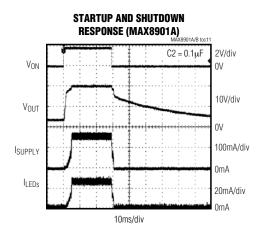
(Circuit of Figure 5, I_{LED} = 24.75mA, V_{IN} = V_{ON} = 3.6V, C1 = 1 μ F, C2 = 1 μ F, T_A = +25°C, unless otherwise noted.)

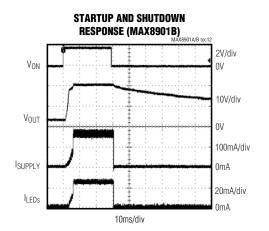








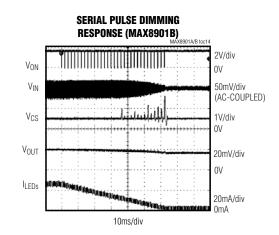


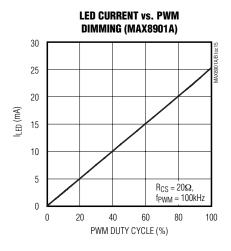


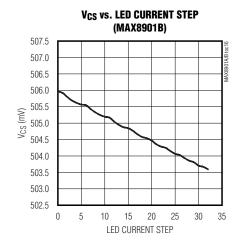
Typical Operating Characteristics (continued)

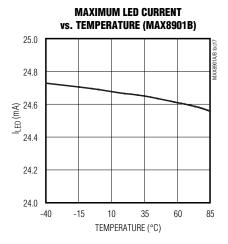
(Circuit of Figure 5, I_{LED} = 24.75mA, V_{IN} = V_{ON} = 3.6V, C1 = 1 μ F, C2 = 1 μ F, TA = +25°C, unless otherwise noted.)

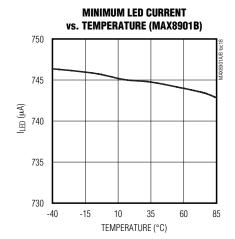
PWM DIMMING RESPONSE (MAX8901A) 2V/div V_{ON} D = 10% 50mV/div V_{IN} (AC-COUPLED) 500mV/div V_{CS} 0V V_{OUT} 20mV/div fon PWM = 100kHz, DUTY 0V CYCLE CHANGE: 90% TO 10%, $C2 = 0.1 \mu F$ 20mA/div I_{LEDs} 4ms/div











Highest Efficiency Supply for 2 to 6 Series WLEDs in a 2mm x 2mm TDFN Package

Pin Description

PIN	NAME	FUNCTION
1	OVP	WLED Overvoltage Protection Input. OVP monitors voltage at the WLEDs. Connect OVP to the positive terminal of the output capacitor. If an OVP condition is detected, the MAX8901_ latch off. Cycle V _{IN} or toggle V _{ON} to restart the IC.
2	ON	Enable and Dimming Control Input. Drive ON high to enable the IC. Drive ON low for greater than 1.33ms (typ) to shutdown the WLED current regulator. For the MAX8901A: After V _{IN} is above V _{UVLO} and ON is driven high, the MAX8901A enters soft-start and increases the WLED current to full brightness. Apply a minimum 30kHz (500kHz max) PWM signal to ON to adjust the WLED brightness from 100% to off, proportional to the duty cycle of the PWM signal. See the <i>PWM Dimming Control (MAX8901A)</i> section for more details. For the MAX8901B: After ON is driven high for 40µs (min), the MAX8901B enters soft-start and increases the WLED current to full brightness. Subsequent pulses on ON cause the MAX8901B WLED current to decrease in 32 equal steps. See the <i>Serial-Pulse Dimming Control (MAX8901B)</i> section for more details.
3	IN	Power Supply Input. Bypass IN to GND with a 1µF ceramic capacitor placed as close as possible to the IC. If V _{IN} exceeds the input overvoltage lockout threshold (6.5V max, V _{IN} rising), the IC stops switching and no WLED current flows (if the forward voltage of the WLED string is greater than V _{IN}). When V _{IN} falls below the overvoltage lockout hysteresis level (6.0V min, V _{IN} falling), soft-start is initiated and normal operation resumes.
4	LX	Boost Inductor Node. Connect an inductor between IN and LX. LX is high impedance in shutdown.
5	PGND	Power Ground. Connect to GND and the exposed paddle (EP) with a short, wide trace.
6	GND	Analog Ground. Connect GND to the exposed paddle with a short, wide trace.
7	COMP	WLED Boost Compensation Node. Connect a 0.022µF ceramic capacitor from COMP to GND. CCOMP stabilizes the converter and sets the soft-start time. COMP discharges to GND when in shutdown.
8	CS	WLED Current Sense Input. For the MAX8901A, connect a current-sense resistor from CS to GND. Voltage sensed at CS regulates the WLED current. For the MAX8901B, do not connect a sense resistor from CS to GND. The MAX8901B provides an internal current source from CS to GND to program the WLED current. The MAX8901B regulates V _{CS} to 0.5V (typ) for all I _{LED} . The MAX8901A regulates V _{CS} to 0.5V (typ) for maximum duty cycle only.
_	EP	Exposed Metal Paddle. Connect EP to GND. For good thermal dissipation, the exposed paddle must be connected to a large ground plane.

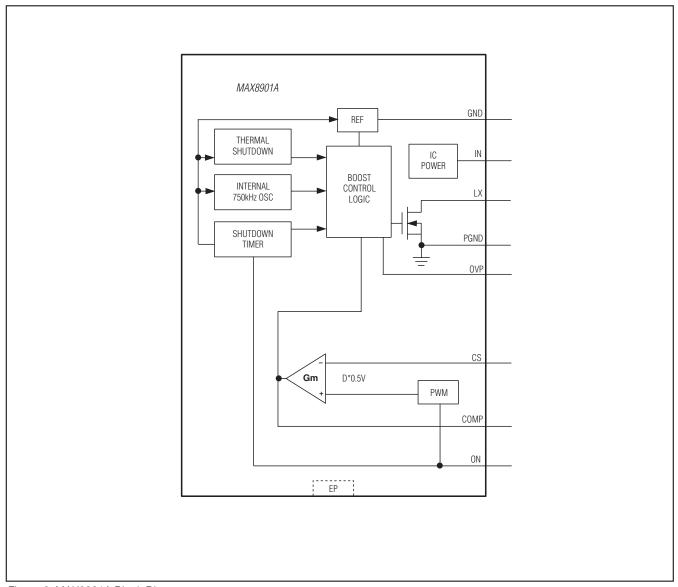


Figure 2. MAX8901A Block Diagram

Detailed Description

The high efficiency and tiny size of the MAX8901A/ MAX8901B WLED step-up converters make them ideally suited for driving LCD backlighting in cell phones, PDAs, and other portable and handheld devices. The MAX8901_ drive from 2 to 6 series-connected WLEDs with constant current for uniform WLED intensity and are optimized for the highest possible efficiency (up to 91%) over the full 1-cell Li+/Li-Poly battery range. These tiny

2mm x 2mm devices operate at a constant 750kHz switching frequency, and use small external components to achieve minimal input and output ripple while occupying the smallest possible footprint.

These converters use a single input (ON) to enable the IC and to control WLED intensity. The MAX8901A requires a direct PWM input to modulate WLED intensity, with WLED current proportional to the PWM duty cycle. The MAX8901B uses single-wire, serial-pulse dimming that

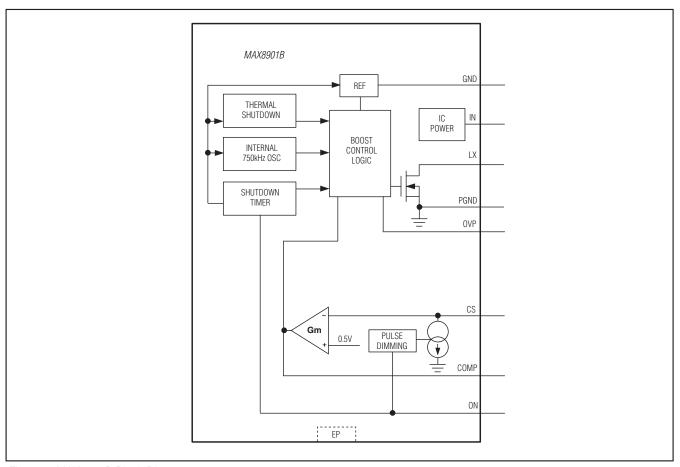


Figure 3. MAX8901B Block Diagram

reduces the WLED intensity in 32 equal steps. Full-scale WLED current for serial-pulse dimming is 24.75mA.

The MAX8901 feature:

- Internal soft-start that gradually illuminates the WLEDs to eliminate inrush current during startup.
- Input overvoltage protection (6.5V max) that prevents IC switching when exceeded.
- WLED overvoltage protection (25V typ) that latches off the IC.
- Shutdown mode that reduces current to 0.01µA (typ).

No WLED current is present in shutdown or during an overvoltage condition if the WLED forward voltage is greater than the input supply voltage. Figures 2 and 3 show the block diagrams for the MAX8901A and MAX8901B, respectively.

Input Overvoltage Lockout

When V_{IN} exceeds 6.2V (min), input overvoltage lockout (OVLO) is engaged to protect the MAX8901_from high-input voltage conditions. Once input OVLO occurs, the MAX8901_ stop switching and no WLED current flows, provided the forward voltage of the WLED string is greater than V_{IN}. When V_{IN} falls below 6V (min), the input OVLO condition is cleared, and the IC is enabled and enters soft-start.

WLED Overvoltage Protection

WLED overvoltage protection (OVP) occurs when the WLED output voltage rises above the WLED OVP threshold. The WLED OVP protection circuitry latches off the IC and the IC enters shutdown when the WLED OVP threshold is exceeded. After OVP is engaged, cycle $V_{\rm IN}$ or toggle ON to reenable the IC and enter soft-start.

Highest Efficiency Supply for 2 to 6 Series WLEDs in a 2mm x 2mm TDFN Package

Startup and Soft-Start

The MAX8901_ use a single input (ON) to enable the IC and to control WLED intensity. After ON is driven high, the MAX8901_ enter soft-start and gradually increase the WLED current to full brightness.

When the MAX8901_ are first turned on, C_{COMP} is charged to 1.25V with a 60 μ A current source. When V_{COMP} rises above 1.25V, C_{COMP} is charged with a 4 μ A current source, and the internal MOSFET begins switching at a reduced duty cycle. When V_{COMP} rises above 2.25V, the duty cycle is at its maximum. The V_{COMP} where the IC exits soft-start depends on the final duty cycle required by the load. Typical startup timing characteristics are shown in the *Typical Operating Characteristics* section.

A PWM signal (MAX8901A) or serial pulses (MAX8901B) are applied to dim the WLEDs. See the *PWM Dimming Control (MAX8901A)* and *Serial-Pulse Dimming Control (MAX8901B)* sections for more details.

Shutdown

The MAX8901_ enter shutdown when VON is held low for more than 1.33ms (typ), or when an output overvoltage condition is engaged. In shutdown, the supply current is reduced to 0.01µA (typ) by powering down the entire IC except for the ON voltage-detection circuitry. CCOMP is discharged to GND during shutdown, allowing the device to reinitiate soft-start when it is re-enabled. Although the internal n-channel MOSFET does not switch in shutdown, there is still a DC current path between the input and the WLEDs through the inductor and Schottky diode. In shutdown, the minimum forward voltage of the WLED string must exceed the maximum input voltage to ensure that the WLEDs remain off. Typically, with two or more WLEDs, the forward voltage is large enough to keep leakage current low, less than 1µA (typ). Shutdown timing characteristics are shown in the Typical Operating Characteristics.

Thermal Shutdown

The MAX8901_ include a thermal shutdown feature that protects the IC by turning it off when the die temperature reaches +160°C (typ). After thermal shutdown occurs, the MAX8901_ must be restarted by toggling the ON pin low, then high, or by cycling the input voltage.

PWM Dimming Control (MAX8901A)

After V_{IN} is above UVLO, apply a PWM signal to ON for a WLED current that is proportional to the PWM signal's duty cycle (0% duty cycle corresponds to zero LED current and 100% duty cycle corresponds to full LED current). Restrict PWM frequency to between 30kHz to 500kHz for a maximum I_{LED} current accuracy. If dimming control is not required, ON works as a simple on/off control. Drive ON high to enable the IC, or drive ON low for shutdown. A resistor connected from CS to GND programs maximum I_{LED}. See the *Current-Sense Resistor* (MAX8901A Only) section for more details.

Serial-Pulse Dimming Control (MAX8901B)

After the MAX8901B is enabled by driving ON high for the minimum initial ON high pulse width (40µs, min), soft-start is engaged and brings WLED current to maximum brightness. After soft-start is completed, dim the MAX8901B by pulsing ON low (500ns to 500µs pulse width). Each pulse reduces the WLED current by 0.75mA. The maximum WLED current is 24.75mA for the MAX8901B (0.75mA/step). The overall dimming range for serial-pulse dimming control is from maximum WLED current down to 1/33 of maximum WLED current. The WLEDs are turned off by holding ON low for at least 1.33ms (typ). See Figure 1 for the serial-pulse dimming control timing diagram.

Applications Information

Inductor Selection

Recommended inductor values range from $10\mu H$ to $47\mu H$. A $22\mu H$ inductor optimizes the efficiency for most applications while maintaining low input voltage ripple. With input voltages near 5V, a larger value of inductance may be more efficient. To prevent core saturation, ensure that the inductor saturation current rating exceeds the peak inductor current for the application. Calculate the peak inductor current with the following formula:

$$I_{PEAK} = \frac{V_{OUT(MAX)} \times I_{LED(MAX)}}{900 \times V_{IN(MIN)}} + \frac{V_{IN(MIN)} \times 1.2}{2 \times L}$$

where I_{LED(MAX)} is in mA and L is in μH.

Highest Efficiency Supply for 2 to 6 Series WLEDs in a 2mm x 2mm TDFN Package

Schottky Diode Selection

The high switching frequency of the MAX8901_demands a high-speed rectification diode for optimum efficiency. A Schottky diode is recommended due to its fast recovery time and low forward voltage drop. Ensure that the diode's average and peak current rating exceeds the average output current and peak inductor current. In addition, the diode's reverse breakdown voltage must exceed VOUT. The RMS diode current can be approximated from the following equation:

IDIODE (RMS) =
$$\sqrt{\text{LED} \times \text{PEAK}}$$

Capacitor Selection

Ceramic capacitors with X5R, X7R, or better dielectric are recommended for stable operation over the entire operating temperature range. The exact values of input (C1) and output (C2) capacitors are not critical (see Figures 4 and 5). The typical value for the input capacitor is 1µF, and the typical value for the output capacitor is 0.1µF (MAX8901A) or 1µF (MAX8901B). Higher value capacitors can reduce input and output ripple, but at the expense of size and higher cost. The compensation capacitor (C3) stabilizes the converter and controls soft-start. The compensation capacitor is typically chosen to be 0.022µF for most applications.

Current-Sense Resistor (MAX8901A Only)

The MAX8901A uses a sense resistor (R_{CS}) connected from CS to GND to program the maximum WLED current for 100% PWM duty cycle. The MAX8901A regulates V_{CS} to 0.5V (typ) for 100% duty cycle. Calculate R_{CS} (in ohms) using the following equation:

$$R_{CS} = \frac{500}{I_{LED(MAX)}}$$

where $I_{\text{LED}(\text{MAX})}$ is the maximum WLED current in milliamps. Maximum WLED current is programmed to 25mA using a 20Ω resistor.

PCB Layout

Due to fast switching waveforms and high-current paths, careful attention is required when the PCB layout is designed. Minimize trace lengths between the IC and the inductor, the diode, the input capacitor, and the output capacitor. For the MAX8901A, minimize the trace length between the IC and RCs. Keep traces short, direct, and wide. Keep noisy traces, such as the LX node trace, away from CS. The input bypass capacitor (CIN) should be placed as close as possible to the IC. The ground connections of CIN and COUT should be as close together as possible. PGND and GND should be connected together at the input capacitor ground terminal. Refer to the MAX8901 EV Kit for an example layout.

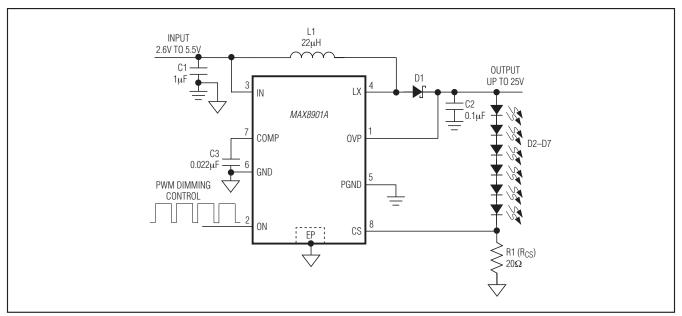


Figure 4. MAX8901A Application Circuit

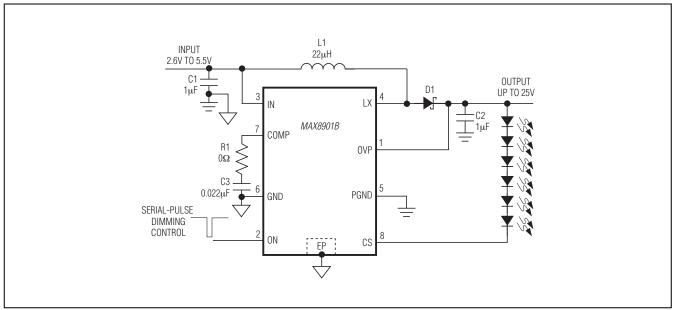
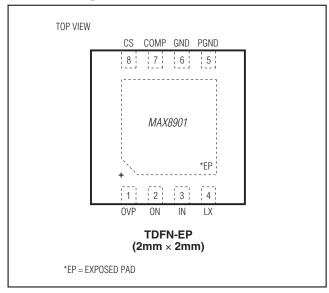


Figure 5. MAX8901B Application Circuit

Pin Configuration



Chip Information

PROCESS: BiCMOS

Package Information

For the latest package outline information and land patterns (footprints), go to www.maximintegrated.com/packages. Note that a "+", "#", or "-" in the package code indicates RoHS status only. Package drawings may show a different suffix character, but the drawing pertains to the package regardless of RoHS status.

PACKAGE	PACKAGE	OUTLINE	LAND
TYPE	CODE	NO.	PATTERN NO.
8 TDFN-EP	T822-1	<u>21-0168</u>	<u>90-0064</u>

Highest Efficiency Supply for 2 to 6 Series WLEDs in a 2mm x 2mm TDFN Package

Revision History

REVISION NUMBER	REVISION DATE	DESCRIPTION	PAGES CHANGED
0	8/07	Initial release	_
1	5/08	Replaced TOC 16	6
2	6/08	Updated LED current accuracy, input undervoltage lockout threshold, and input overvoltage lockout threshold.	1, 2, 3
3	8/10	Removed Input Undervoltage Lockout Threshold (V _{IN} falling) minimum value and added soldering temperature	2
4	4/15	Updated Benefits and Features section	1

For pricing, delivery, and ordering information, please contact Maxim Direct at 1-888-629-4642, or visit Maxim Integrated's website at www.maximintegrated.com.

Maxim Integrated cannot assume responsibility for use of any circuitry other than circuitry entirely embodied in a Maxim Integrated product. No circuit patent licenses are implied. Maxim Integrated reserves the right to change the circuitry and specifications without notice at any time. The parametric values (min and max limits) shown in the Electrical Characteristics table are guaranteed. Other parametric values quoted in this data sheet are provided for guidance.